

UNCLASSIFIED

AD **256 554**

*Reproduced
by the*

ARMED SERVICES TECHNICAL INFORMATION AGENCY
ARLINGTON HALL STATION
ARLINGTON 12, VIRGINIA



UNCLASSIFIED

NOTICE: When government or other drawings, specifications or other data are used for any purpose other than in connection with a definitely related government procurement operation, the U. S. Government thereby incurs no responsibility, nor any obligation whatsoever; and the fact that the Government may have formulated, furnished, or in any way supplied the said drawings, specifications, or other data is not to be regarded by implication or otherwise as in any manner licensing the holder or any other person or corporation, or conveying any rights or permission to manufacture, use or sell any patented invention that may in any way be related thereto.

Technical Report No. AF-87

734 500

256554

CATALOGED BY ASTIA
AS AD No. _____

THE CUT-OFF WAVELENGTHS AND POWER-VOLTAGE IMPEDANCES
OF COMPOSITE WAVEGUIDES FOR THE FUNDAMENTAL MODE

by

G. R. Valenzuela



THE JOHNS HOPKINS UNIVERSITY
RADIATION LABORATORY
BALTIMORE 2, MD.

CONTRACT NO. AF 33(616)-6753

MAY 1961



61-3-2
XEROX

Copy No. 47

THE JOHNS HOPKINS UNIVERSITY
RADIATION LABORATORY
BALTIMORE, MARYLAND

Technical Report No. AF-87

The Cut-Off Wavelengths and Power-Voltage
Impedances of Composite Waveguides for the Fundamental Mode

by

G. R. Valenzuela

Contract AF 33(616)-6753

May, 1961

ACKNOWLEDGEMENT

Acknowledgement is due to Miss Marguerite Velten for performing the numerical calculation and to Mr. Tomos L. apRhys for reviewing the manuscript.

CONTENTS

	<u>Page</u>
I. INTRODUCTION	1
II. THEORY	2
III. APPLICATION	6
IV. EXPERIMENT	12
V. CONCLUSION	15
REFERENCES	16
DISTRIBUTION	

ABSTRACT

The Rayleigh-Ritz method is used to solve the eigenvalue problem for electromagnetic waves propagating through a waveguide of a non-simple cross section. The method is applied to obtain the cut-off wavelength and the power-voltage impedance for the fundamental mode (that equivalent to the H_{10} mode in rectangular waveguide and the H_{11} mode in circular waveguide) for a waveguide with semicircular side walls and flat top and bottom walls, and for a truncated-circular waveguide.

Numerical results are given for both waveguides, and supporting experimental evidence is adduced for the waveguide with semicircular sides and flat top and bottom walls, for aspect ratios of 0.302 and 0.702. The cut-off wavelengths obtained are within 3 per cent while the power-voltage impedances are within 6 per cent of their respective exact values.

I. INTRODUCTION

Interest in these composite waveguides arises from their possible applications in transition from rectangular to circular waveguides, where they provide convenient intermediate sections.

When electromagnetic waves propagate in waveguides of non-simple cross section no exact solution can be obtained by the conventional method of solving the wave equation by separation of variables. In such cases approximations are available with the aid of perturbation methods and variational methods¹.

The Rayleigh-Ritz method will be used to obtain an approximate solution for the cut-off wavelength and the power-voltage impedance in waveguides with semicircular side walls and flat top and bottom walls, and for truncated-circular waveguides.

The results obtained contradict the results published in Principles of Microwave Circuits² on the cut-off wavelength for a waveguide of semicircular side walls and flat top and bottom walls for aspect ratios between 0.3 and 1.0.

II. THEORY

The boundary value problem of electromagnetic waves propagating in a cylindrical waveguide can always be reduced to a simple two dimensional problem. If the longitudinal waveguide axis is in the z direction,

$$\nabla^2 \psi_i(x, y) + k_i^2 \psi_i(x, y) = 0, \quad (1)$$

where $\psi_i(x, y)$ and k_i are the eigenfunction and the propagation constant for the i^{th} mode, respectively.

Kornhauser³ showed that in any cylindrical waveguide the lowest order mode with non-zero eigenvalue is an H-mode. Now let ψ represent the eigenfunction for the fundamental mode and drop any subscript. The eigenfunction ψ is proportional to the magnetic field component in the direction of propagation.

Since the modes in a waveguide are orthogonal, that is,

$$\int_S \psi_m \psi_n d\sigma = \delta_{mn} \quad (\text{kroncker delta}), \quad (2)$$

where S is the waveguide cross section, the eigenfunction ψ for the fundamental mode can be constructed from³

$$\int_S \psi d\sigma = 0. \quad (3)$$

Then a legitimate choice for the eigenfunction for the fundamental mode is,

$$\psi(x) = \sum_{n=0}^N C_{2n+1} x^{2n+1}. \quad (4)$$

For simplicity choose $N = 1$, then

$$\psi(x) = C_1 x + C_3 x^3 \quad (5)$$

In general it can be easily shown that if ψ satisfies either the Dirichlet or Neumann boundary conditions, its eigenvalue is minimized by the expression.

$$k^2 = \frac{\int |\text{grad } \psi|^2 d\sigma}{\int \psi^2 d\sigma} \quad (6)$$

the integrals being taken over the waveguide cross section.

By using the trial function (5) for the fundamental mode, Equation (6) takes the following general form,

$$k^2 = \frac{\alpha C_1^2 + \beta C_1 C_3 + \gamma C_3^2}{\delta C_1^2 + \epsilon C_1 C_3 + \zeta C_3^2} \quad (7)$$

where α , β , γ , δ , ϵ , and ζ are constants dependent on the dimensions of the waveguide.

Then the parameters C_1 and C_3 are adjusted independently to minimize the eigenvalue k^2 .

$$\frac{\partial [k^2]}{\partial C_1} = 0 \quad \frac{\partial [k^2]}{\partial C_3} = 0.$$

After performing the minimization, we obtain the following system of equations:

$$\begin{aligned} (2\delta k^2 - 2\alpha) C_1 + (\epsilon k^2 - \beta) C_3 &= 0 \\ (\epsilon k^2 - \beta) C_1 + (2\zeta k^2 - 2\gamma) C_3 &= 0 \end{aligned} \quad (8)$$

The non-trivial solution is found in the conventional way by equating the determinant of the coefficients to zero, which yields

$$k^4 - 2(H)k^2 + \Phi = 0, \quad (9)$$

where

$$(H) = \frac{2\alpha\zeta + 2\delta\gamma - \epsilon\beta}{4\delta\zeta - \epsilon^2}$$

and

$$\Phi = \frac{4\alpha\gamma - \beta^2}{4\delta\zeta - \epsilon^2}$$

An expression for the power-voltage impedance is obtained next. The power-voltage impedance is defined as the ratio of the maximum mean voltage $V/\sqrt{2}$ across the waveguide squared divided by the total average power W flowing through the waveguide.

$$Z_{WV} = \frac{VV^*}{2W}$$

The maximum voltage across the waveguide can be found from

$$V = 2 \int_0^b E_y |_{x=0} dy$$

where the electric field $E_y = \frac{\partial \psi}{\partial x} = C_1 + 3C_3x^2$. The maximum voltage then is

$$V = 2bC_1 \quad (10)$$

The power flowing down the waveguide can be obtained for an H mode from

$$W = \frac{k^2}{2Z_0} \int \psi^2 d\sigma \quad (11)$$

where

$$Z_0 = 120\pi \lambda_g / \lambda \quad \text{is the field impedance, and}$$

as usual,

$$\lambda_g = \text{is the waveguide wavelength}$$

and

$$\lambda = \text{is the free space wavelength.}$$

The expression for the power-voltage impedance for composite waveguides is,

$$Z_{WV} = \frac{b^2 Z_0}{\alpha + \beta \left(\frac{C_3}{C_1} \right) + \gamma \left(\frac{C_3}{C_1} \right)^2} \quad (12)$$

The ratio $\frac{C_3}{C_1}$ can be obtained from Equation (8).

III. APPLICATION

1. The approximate method was used to obtain the eigenvalue for the fundamental mode of waveguide with semicircular side walls and flat top and bottom walls. See Figure 1.

The coefficients obtained for this composite waveguide are the following:

$$\begin{aligned} \alpha &= ab \left[p + \frac{\pi}{4} r \right] \\ \beta &= a^3 b \left[2p^3 + \frac{3\pi}{2} p^2 r + 4pr^2 + \frac{3\pi}{8} r^3 \right] \\ \gamma &= a^5 b \left[\frac{9}{5} p^5 + \frac{9\pi}{4} p^4 r + 12p^3 r^2 + \frac{27\pi}{8} p^2 r^3 + \frac{24}{5} pr^4 + \frac{9\pi}{32} r^5 \right] \\ \delta &= a^3 b \left[\frac{1}{3} p^3 + \frac{\pi}{4} p^2 r + \frac{2}{3} pr^2 + \frac{\pi}{16} r^3 \right] \\ \epsilon &= a^5 b \left[\frac{2}{5} p^5 + \frac{\pi}{2} p^4 r + \frac{8}{3} p^3 r^2 + \frac{3\pi}{4} p^2 r^3 + \frac{16}{15} pr^4 + \frac{\pi}{16} r^5 \right] \\ \zeta &= a^7 b \left[\frac{1}{7} p^7 + \frac{\pi}{4} p^6 r + 2p^5 r^2 + \frac{15\pi}{16} p^4 r^3 + \frac{8}{3} p^3 r^4 + \frac{15\pi}{32} p^2 r^5 \right. \\ &\quad \left. + \frac{48}{105} pr^6 + \frac{5\pi}{256} r^7 \right] \end{aligned}$$

where

$$r = \frac{b}{a} \text{ and } p = 1 - r.$$

The numerical results are shown in Figure 2.

2. The method was also applied to a truncated-circular waveguide. See Figure 3.

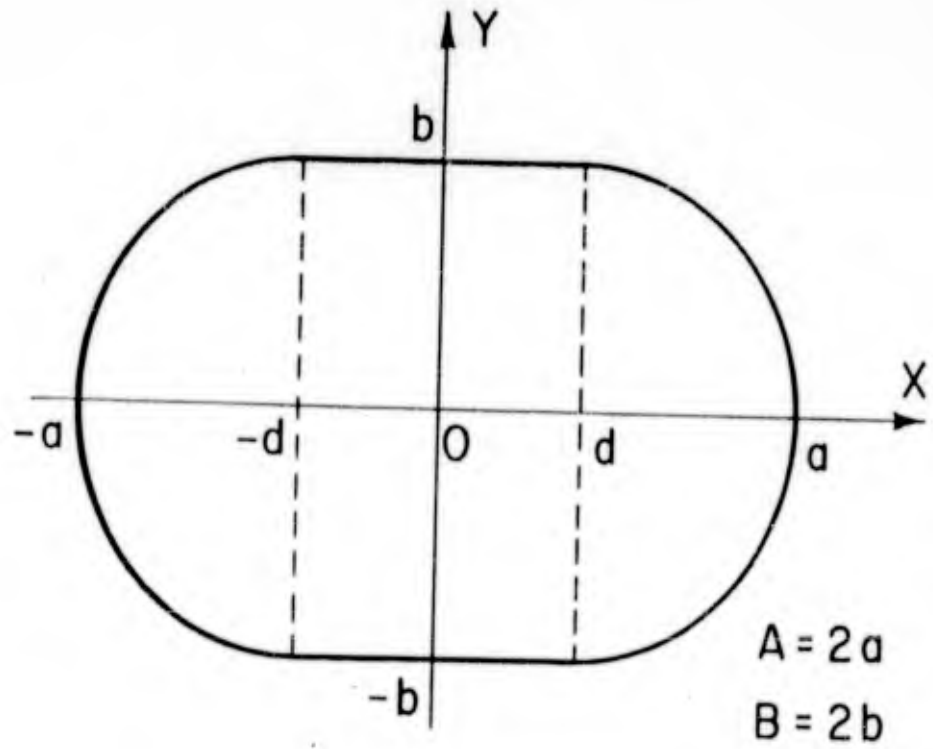


FIGURE 1: COMPOSITE CIRCULAR WAVEGUIDE

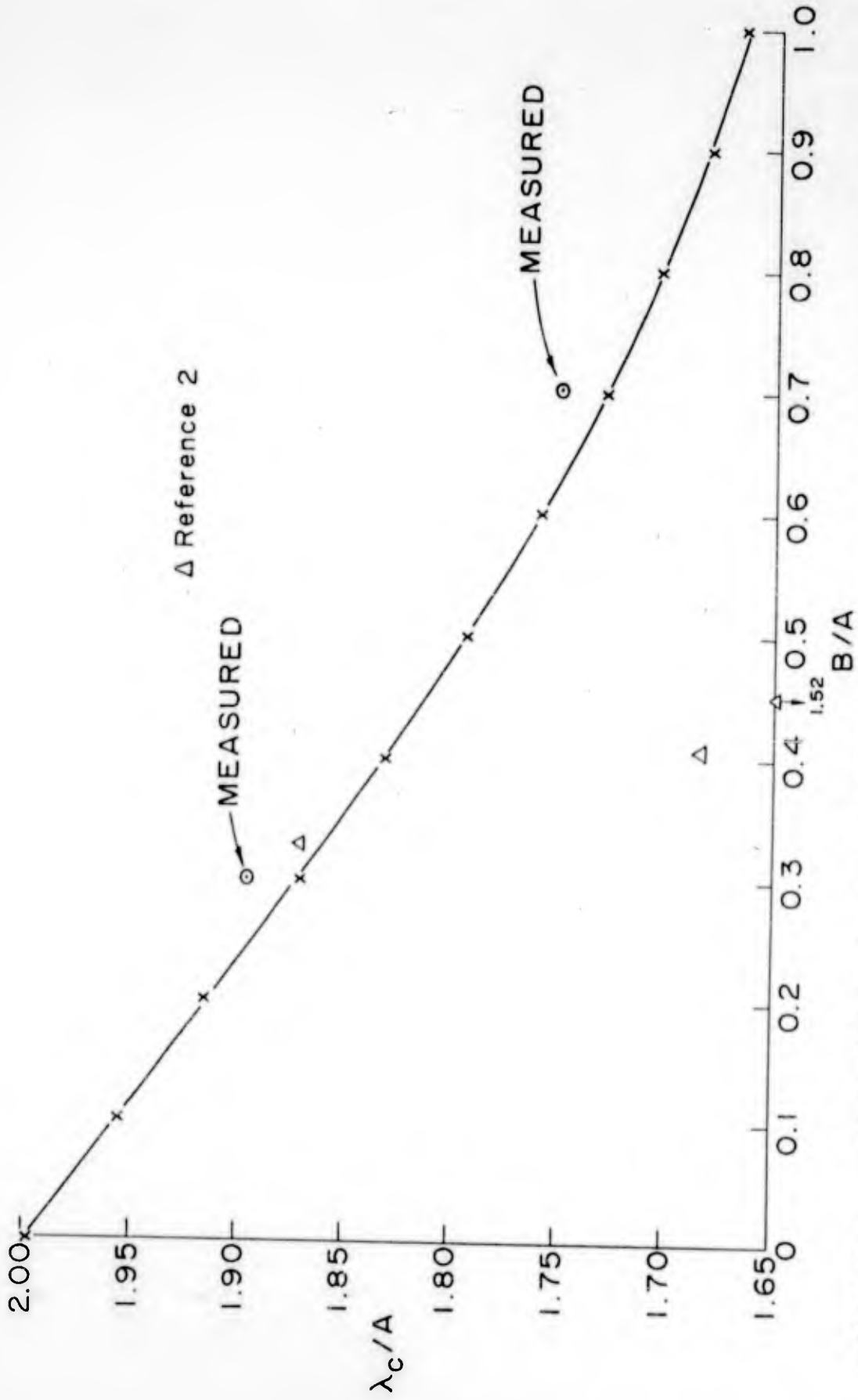


FIGURE 2: SEMI-CIRCULAR SIDE WALLS AND UPPER AND BOTTOM FLAT WALLS WAVEGUIDE CUT-OFF WAVELENGTH PER WIDE DIMENSION AS A FUNCTION OF NARROW TO WIDE DIMENSION RATIO.

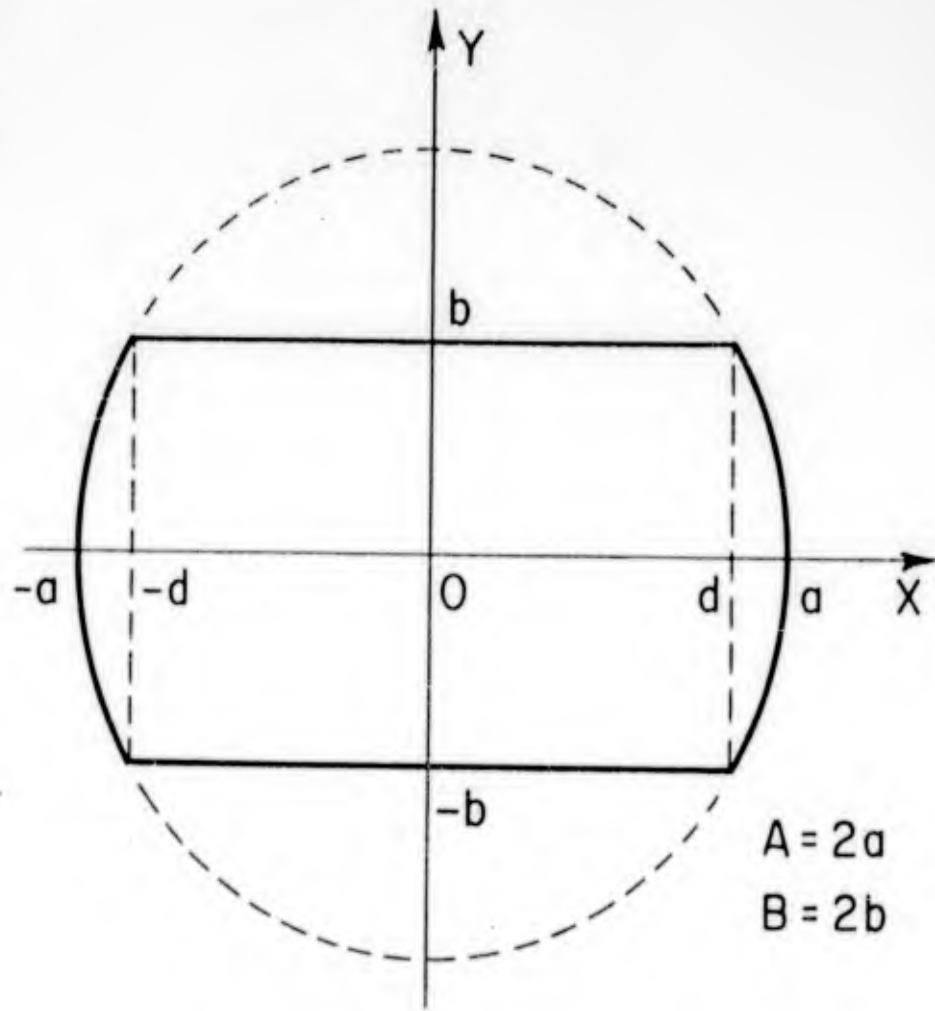


FIGURE 3: TRUNCATED CIRCULAR WAVEGUIDE

For this waveguide the coefficients obtained are the following:

$$\begin{aligned} \alpha &= a^2 \left[qr + \frac{1}{2} \Delta - \frac{1}{4} \sin 2\tau \right] \\ \beta &= a^4 \left[2q^3 r + \frac{3}{4} \Delta + \frac{3}{16} \sin 4\tau \right] \\ \gamma &= a^6 \left[\frac{9}{5} q^5 r + \frac{3}{2} \sin^3 \tau \cos^3 \tau + \frac{9}{16} \Delta + \frac{9}{64} \sin 4\tau \right] \\ \delta &= a^4 \left[\frac{1}{3} q^3 r + \frac{1}{8} \Delta + \frac{1}{32} \sin 4\tau \right] \\ \epsilon &= a^6 \left[\frac{2}{5} q^5 + \frac{\sin^3 \tau \cos^3 \tau}{3} + \frac{1}{8} \Delta + \frac{1}{32} \sin 4\tau \right] \\ \zeta &= a^8 \left[\frac{1}{7} q^7 r + \frac{1}{8} \left\{ \sin^5 \tau \cos^3 \tau + \frac{5}{6} \sin^3 \tau \cos^3 \tau + \frac{5}{16} \Delta + \frac{5}{64} \sin 4\tau \right\} \right] \end{aligned}$$

where

$$r = \frac{b}{a}; \quad q = (1-r^2)^{1/2}; \quad \tau = \sin^{-1} q; \quad \Delta = \left(\frac{\pi}{2} - \tau \right).$$

The numerical results are plotted in Figure 4.

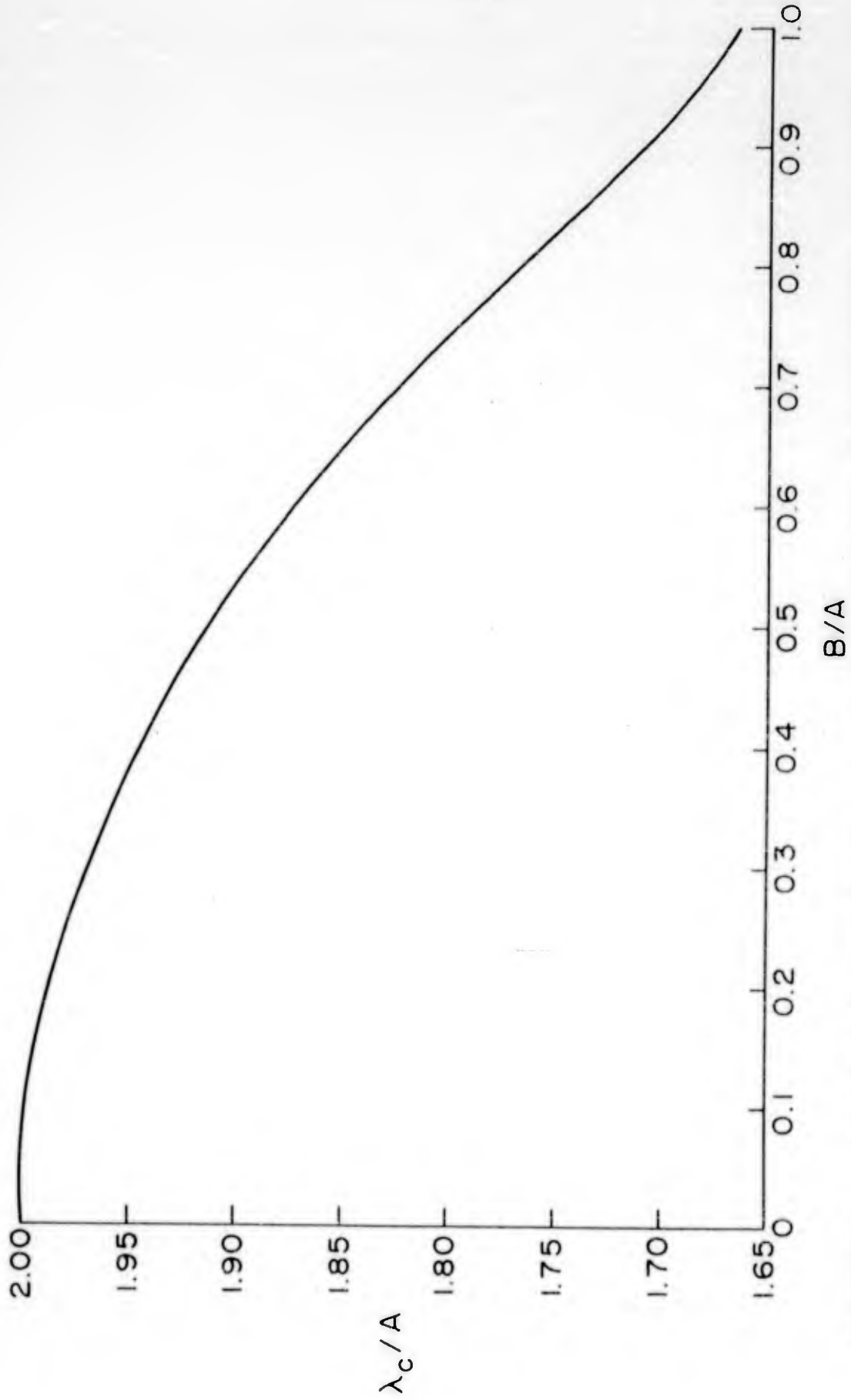


FIGURE 4: TRUNCATED-CIRCULAR WAVEGUIDE CUT-OFF WAVELENGTH PER WIDE DIMENSION AS A FUNCTION OF NARROW TO WIDE DIMENSION RATIO.

IV. EXPERIMENTAL

To assess the accuracy of the results, an experiment was performed. Two 3" waveguide sections with semicircular side walls and fiat top and bottom walls of aspect ratios of 0.302 and 0.702 were electroformed. Their wide dimension was 0.622 inch the width of the standard P-band rectangular waveguide. See Figure 5.

The cut-off wavelength was obtained by two independent methods: firstly, by measuring the phase change when the composite waveguide was shortened by about 1 inch; and secondly, by measuring the transmission-versus-frequency characteristic using an electrically-swept backward-wave oscillator (source).

The measured points and the points obtained from reference 2 are also included in Figure 2.

The power-voltage impedance was also measured. For this purpose a termination was developed for the composite waveguide sections, and by using a standard P-band slotted line the impedance was measured at the junction between the rectangular and composite waveguide. The real part accounts for the impedance mismatch while the imaginary part accounts for the junction susceptance.

The measured points are shown in Figure 6.

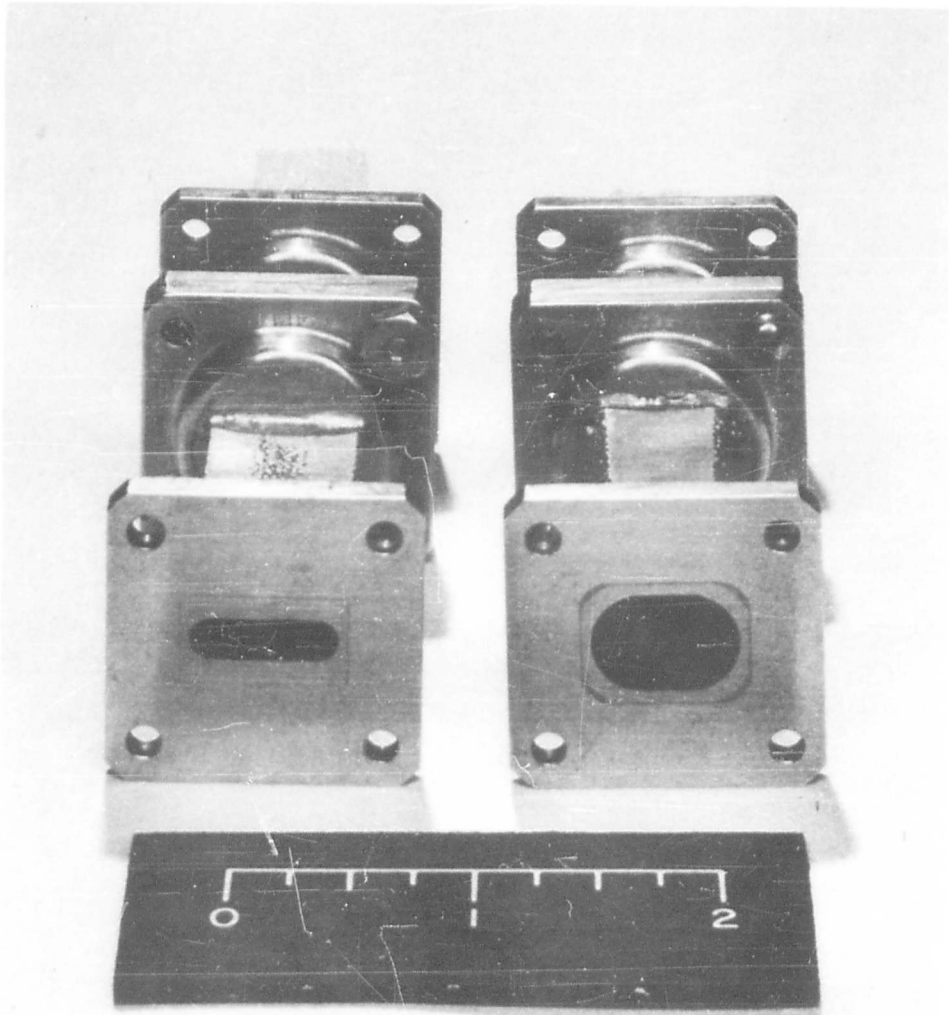


FIGURE 5: WAVEGUIDE WITH FLAT TOPS AND BOTTOMS AND SEMICIRCULAR SIDE WALLS EXPERIMENTAL SECTIONS

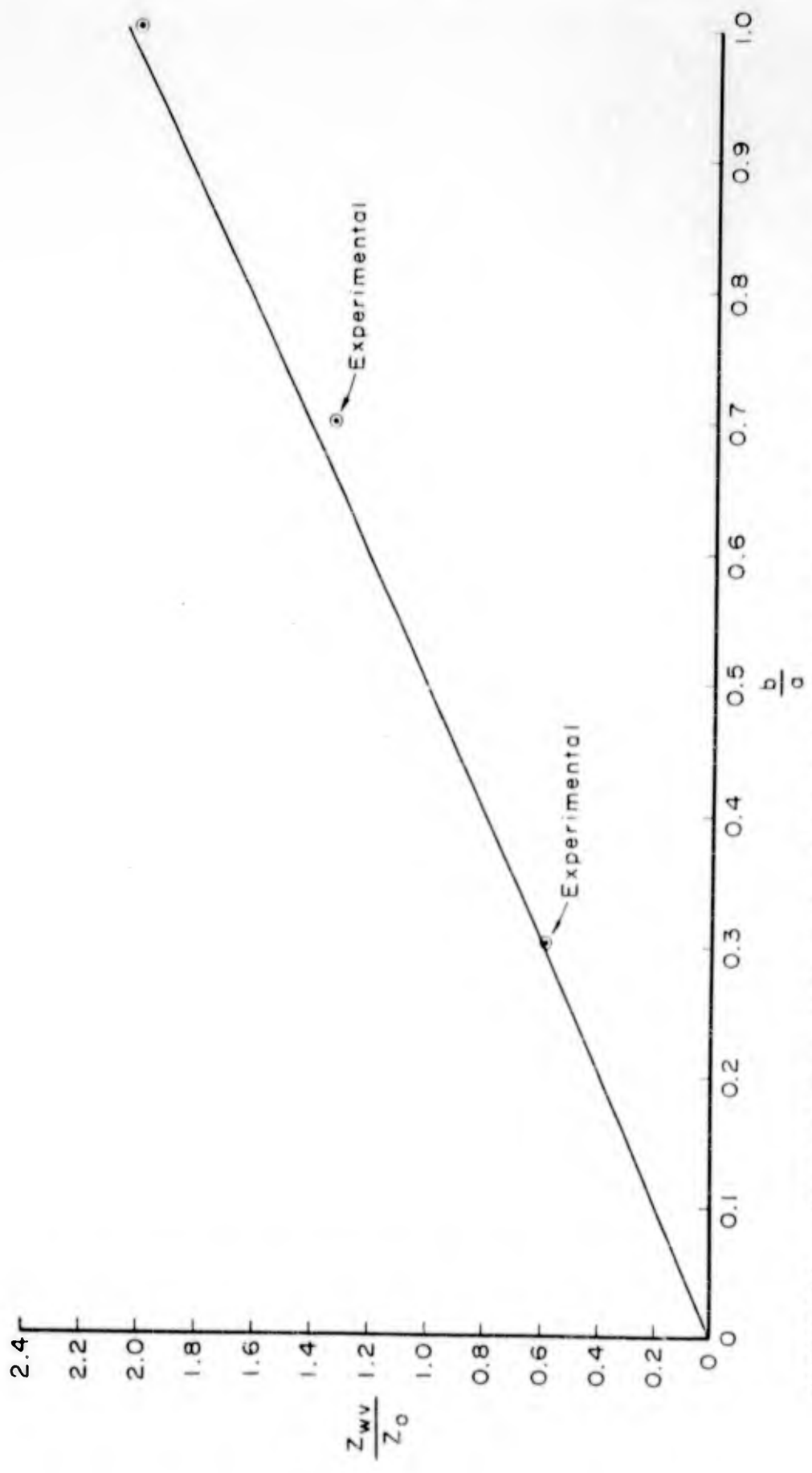


FIGURE 6: THE POWER-VOLTAGE IMPEDANCE AS A FUNCTION OF ASPECT RATIO FOR COMPOSITE WAVEGUIDES.

V. CONCLUSION

The Rayleigh-Ritz method was applied to the eigenvalue problem for the fundamental mode in composite waveguides. Supporting experimental evidence was obtained for the waveguide of semicircular side walls and flat top and bottom walls, which indicates the cut-off wavelengths are within 3 per cent and the power-voltage impedances are within 6 per cent of their exact value.

The cut-off wavelengths obtained by this approximate method are smaller than the exact value, since this is a minimization problem of the eigenvalue k^2 .

The cut-off wavelength for the waveguide of semicircular side walls and flat top and bottom walls indicated in Reference 2 for aspect ratios of 0.406 and 0.450 are even smaller than the cut-off wavelengths obtained by this method; therefore their approximation must have been inferior to the one used in this paper.

Surprisingly enough the calculated values for the power-voltage impedance for both composite waveguides came out to be identical for all practical purposes.

The accuracy could be improved by selecting a trial function containing terms in Y^2 .

REFERENCES

1. Morse, P. R. and Feshbach, H., Methods in Theoretical Physics II, McGraw-Hill Book Co., Inc., New York, New York, 1953.
2. Montgomery, C. G., Dicke, R. H. and Purcell, E. M., Principles of Microwave Circuits, MIT Rad. Lab. Ser., McGraw-Hill Book, Co., Inc., New York, New York, Vol. 8, p. 45, 1948.
3. Kornhauser, E. T. and Stakgold, Application of Variational Methods to the Equation $\nabla^2 u + \lambda u = 0$, Harvard Univ., Cruft Laboratory, Cambridge, Mass., T.R. No. 117, October 5, 1950.

ELECTRONICS DISTRIBUTION

1	Wright Air Development Division (WWACC) Wright-Patterson AFB, Ohio Attn: WWAD	1	Director Weapons Systems Evaluation Group Room 1 E 880, The Pentagon Washington 25, D. C.	1	Bjorksten Research Labs., Inc. P. O. Box 265 Madison, Wisconsin Attn: Security Office
2	WWDPVD-1	1	Office of Secretary of Defense Office of Asst. Secretary (R. and D.) W. Pentagon Washington 25, D. C.	1	Boeing Airplane Company Pilotless Aircraft Division P. O. Box 3925 Seattle 24, Washington Attn: Library Supervisor
1	WWDPVD-2	1	Commander, Fort Monmouth Red Bank, New Jersey Attn: Countermeasures Branch	1	Boeing Airplane Company Mallstop 18-47 Seattle 14, Washington Attn: Cont. AF 33(038)-19589
1	WWRDM	1	Director, Coles Signal Laboratory Fort Monmouth, New Jersey	1	Laboratories for Applied Sciences, University of Chicago Museum of Science and Industry Chicago 37, Illinois Attn: Library-W-305
1	WWRDS	1	Commanding General Countermeasures Division, DST Fort Monmouth, New Jersey Attn: USASGS, Mr. H. Allen	1	Convair Fort Worth, Texas Attn: Chief Librarian
1	WWRNA	1	Chief, Bureau of Naval Weapons Department of the Navy Washington 25, D. C.	1	Cook Research Laboratories 6401 Oakton Street Morton Grove, Illinois Attn: Cont. NOrd-16369
1	WWRNE	1	Attn: EL-45	1	Cornell Aeronautical Laboratory 4455 Genesee Street P. O. Box 235 Buffalo 21, New York Attn: Dr. Carl Miller Cont. AF 33(616)-6307
1	WWRNEA	1	EL-411	1	Electronic Communications, Inc. 1830 York Road Timonium, Maryland
3	WWRNGW-1	1	Chief, Bureau of Ships, Code 335 Room 4532, Main Navy Building 18th and Constitution Ave., N. W. Washington 25, D. C.	1	Fairchild Astrionics Division Fairchild Engine and Airplane Corporation Wyandanch, Long Island, New York
1	WWRNGW-2	1	Commander U. S. Naval Missile and Astronautics Center Point Mugu, California Attn: Technical Library	1	General Electric Advanced Electronics Center Cornell University Ithaca, New York Attn: Librarian
1	WWRNGW-3	1	New York Naval Shipyard Brooklyn 1, New York Attn: Material Laboratory	1	General Electric Company B-278, 901 Broad Street Utica, New York Attn: Cont. AF 33(600)-30632
2	WWRNR	1	Commanding Officer U. S. Naval Air Development Center Engineering Development Department Johnsville, Pennsylvania Attn: J. M. McGlone	1	Grumman Aircraft Engineering Corp. Bethpage, Long Island, New York Attn: Engineering Library Plant 5
1	WWRNRS	1	Commanding Officer (Code 74) Naval Ordnance Laboratory Corona, California	1	HRB-Singer, Inc. P. O. Box 60 Science Park State College, Pennsylvania
1	WWZB	2	Director Naval Research Laboratory Washington 25, D. C.	1	Hallcrafters Company 4401 W. Fifth Avenue Chicago 24, Illinois Attn: Cont. AF 33(600)-26117
1	AFCIN-4B1a	1	Director National Security Agency Ft. George G. Meade, Md. Attn: CREF-332 (Room 2C087)	1	Hughes Aircraft Company c/o Plant Representative Florence and Teal Streets Culver City, California Attn: Company Technical Document Center, Bldg. 6, Room C2048
1	AFCIN-4C3	1	U. S. Army Signal Electronic Research Unit P. O. Box 205 Mountain View, California	1	University of Illinois Coordinated Science Laboratory 264 Engineering Research Laboratory Urbana, Illinois
	Commander Rome Air Development Center Griffiss Air Force Base Rome, New York	1	Commanding General Army Rocket and Guided Missile Agency Redstone Arsenal, Alabama Attn: ORDXR-RF	1	Jansky and Bailey, Inc. Systems Engineering Dept. 1339 Wisconsin Ave., N. W. Washington 7, D. C. Attn: Cont. AF 33(616)-6740
1	Attn: RCEMSA	1	Commanding General White Sands Missile Range New Mexico Attn: ORDBS-OM-Tech. Library-403	1	Radio Corporation of America Delaware and Cooper Camden, New Jersey Attn: Mr. Jack J. Rudnick Cont. AF 33(600)-39608
1	RCERHS	1	Missile Electronics Laboratories Ordnance Missile Laboratory Army Rocket and Guided Missile Agency Redstone Arsenal, Alabama Attn: J. P. Hallows, Jr. Bldg. 5429 P. S. Birman		
1	RGSSST	1	Office, Chief Signal Officer Room 2# - 258 The Pentagon Washington 25, D. C.		
1	RCU	10	ASTIA Arlington Hall Station Arlington 12, Virginia Attn: TIPDR		
1	RCWE	1	Airborne Instruments Laboratory Walt Whitman Road Melville, Long Island, New York Attn: Cont. AF 33(616)-2143		
1	RCWEA				
2	RCWEC				
1	RCWED				
1	RCWEE				
1	RCWES				
1	Air Force Cambridge Research Labs. AFRD (ARDC) CRREL-R Scientific Library Branch L. G. Hanscom Field Bedford, Massachusetts				
1	Commander, HQ, ARDC Air Force Ballistic Missile Division 5760 Arbor Vitae Avenue Inglewood, California Attn: WDSOT				
1	Director, Development Planning HQ, USAF Washington 25, D. C. Attn: AFDAP-T				
1	Director, Research and Technology HQ, USAF Washington 25, D. C. Attn: AFDRT-ER				
1	Commander, HQ Air Defense Command Ent AFB, Colorado Attn: ADOOA				
1	Commander-in-Chief Strategic Air Command Offutt AFB, Nebraska Attn: Operations Analysis				
1	Tactical Air Command Directorate of Operations Analysis Langley AFB, Virginia				
1	HQ, Ninth Air Force Shaw AFB, South Carolina				
1	HQ, Air Force Missile Development Center Holloman AFB, New Mexico Attn: MDTXT				
1	Director Air University Library Maxwell AFB, Alabama Attn: AUL-6234				
1	HQ, USAF Security Service Kelly AFB San Antonio, Texas Attn: CLR, Russell Watson				
5	National Aeronautics and Space Administration 1520 H Street, N. W. Washington 25, D. C. Attn: Bertram A. Mulcahy Assistant Director for Technical Information				

ELECTRONICS DISTRIBUTION (CONTD)

- 1 Radio Corporation of America
Talos Equipment Projects
Moorestown, New Jersey
Attn: H. D. Albrecht
Coordinator, Bldg, 108-2
- 1 Radio Corporation of America
Missile Electronics and Controls Division
Box 588
Dartington, Massachusetts
Attn: W. Ramsey
- 1 Radio Corporation of America
WCM + SRD Eng. Library 306/2
Attn: L. R. Hund, Librarian
8500 Balboa Blvd.
Van Nuys, California
- 1 Ramo Wooldridge, a Division of
Thompson Ramo Wooldridge, Inc.
8433 Fallbrook Avenue
Canoga Park, California
- 1 RAND Corporation
1700 Main Street
Santa Monica, California
Attn: Cont. AF 18(600)-1600
- 1 Raytheon Company
Santa Barbara Operation
P. O. Box 636
Santa Barbara, California
Attn: Cont. AF 33(600)-28016
- 1 Raytheon Company Equipment Division
Surface Radar and Navigation Operations
State Road West
Wayland, Massachusetts
Attn: Librarian
- 1 Revere Copper and Brass, Inc.
196 Diamond Street
Brooklyn 22, New York
Attn: Cont. AF 33(616)-5672
V. Lane, Security Officer
- 1 Space Technology Laboratories, Inc.
P. O. Box 95001
Los Angeles 45, California
Attn: Technical Information Center,
Document Procurement
- 1 Sperry Gyroscope Company
Great Neck, Long Island
New York
Attn: Cont. AF 33(038)-14524
- 1 Stanford Electronics Laboratories
Stanford University
Stanford, California
Attn: Applied Electronics Laboratory
Document Library
- 1 Sylvania Electric Products, Inc.
100 First Avenue
Waltham 54, Massachusetts
Attn: Dr. L. S. Sheingold
Cont. AF 33(600)-24208
- 1 Sylvania Electronic Systems
Sylvania Electric Products, Inc.
P. O. Box 188
Mountain View, California
Attn: Roger Battie
Supervisor, Technical Liaison
- 1 University of Illinois
Ultramicrowave Group
Electrical Engineering Research
Urbana, Illinois
- 1 International Telephone and Telegraph
Corp.
500 Washington Avenue
Nutley 10, New Jersey
Attn: Cont. W33-038-ac-15012
- 1 The Johns Hopkins University
Applied Physics Laboratory
8621 Georgia Avenue
Silver Spring, Maryland
Attn: Cont. NOrd-7386
- 1 Litton Industries, Inc.
4900 Calvert Road
College Park, Maryland
Attn: Jack Giles
- 1 Lockheed Aircraft Corporation
Missile and Space Division
3251 Hanover Street
Palo Alto, California
Attn: Technical Information Center
- 1 Lockheed Aircraft Corporation
Georgia Division
Marietta, Georgia
Attn: Dept. 72-15
Cont. AF 34(601)-4852
- 1 The Martin Company
Baltimore 3, Maryland
Attn: Engineering Librarian
Cont. AF 33(600)-21703
- 2 Lincoln Laboratory
Mass. Inst. of Technology
P. O. Box 73
Lexington 73, Massachusetts
Attn: Library D-224
- 1 Electronic Systems Laboratory
Mass. Inst. of Technology
Cambridge, Massachusetts
Attn: Cont. AF 33(616)-5477
- 1 Melpar, Inc.
3000 Arlington Blvd.
Falls Church, Virginia
Attn: Cont. AF 33(038)-21250
- 1 University of Michigan
Willow Run Research Center
Ypsilanti, Michigan
Attn: Cont. AF 33(616)-3153
- 1 Radiation Laboratory
The University of Michigan
201 Catherine Street
Ann Arbor, Michigan
Attn: Keeve M. Slegel
- 1 Electronics Defense Group
Dept. of Electrical Engineering
Eng. Research Institute
University of Michigan
Ann Arbor, Michigan
Attn: Cont. DA-36-038-sc-15358
- 1 The Mitre Corporation
P. O. Box 208
Bedford, Massachusetts
- 1 Motorola, Inc.
Riverside, California
- 1 North American Aviation, Inc.
International Airport
Los Angeles 45, California
Attn: Cont. AF 33(600)-36599
J. W. Schwartz
- 1 Northrup Corporation
Norair Division
1001 E. Broadway
Hawthorne, California
Attn: Technical Information
- 1 The Ohio State University
Research Foundation
1314 Kinnear Road
Columbus 12, Ohio
Attn: Dr. T. E. Tice
- 1 University of Texas
Defense Research Laboratory
Box 8029 University Station
Austin, Texas
Attn: Cont. AF 33(616)-313
- 1 Westinghouse Electric Corp.
Air Arm Division
Engineering Library
P. O. Box 746
Baltimore 3, Maryland
Attn: A. E. Battaglia
Librarian

UNCLASSIFIED

UNCLASSIFIED