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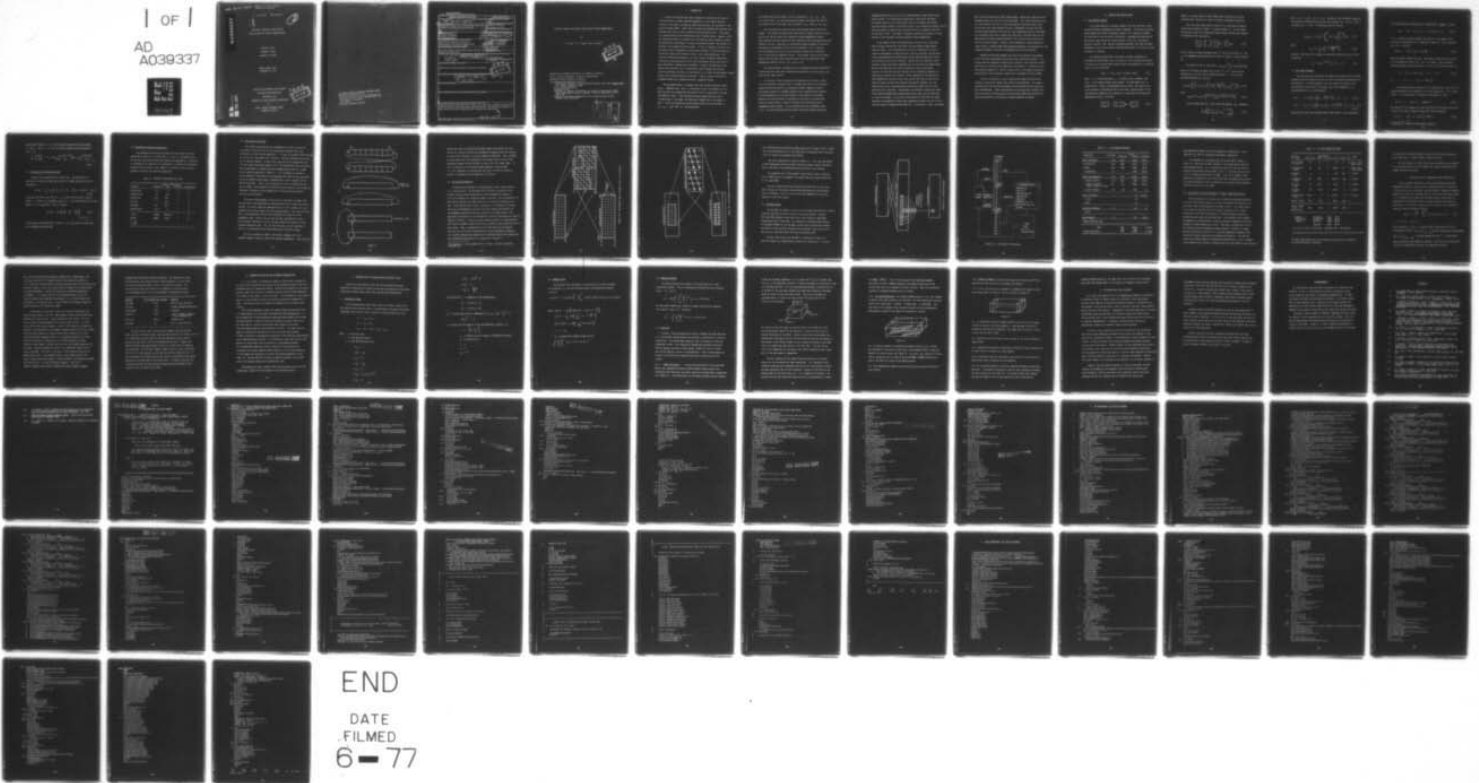
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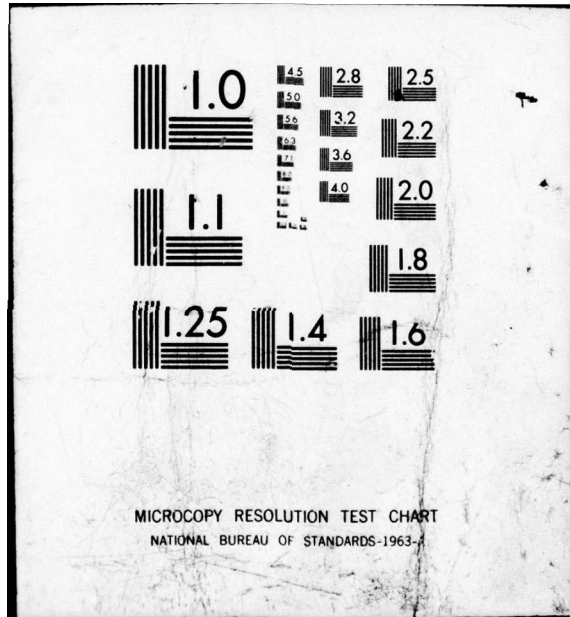
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PIPELINE, PARALLEL AND SERIAL
REALIZATION OF PHASE DEMODULATORS

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1. INTRODUCTION

A theory of designing the "best" processor for estimating the state of a Markov process or signal observed indirectly as a zero memory nonlinear function of the state, corrupted by additive white noise, was developed in the early and middle 1960s. "Best" here is to mean the estimate that minimizes the expected loss, with the loss a function of the estimation error. This theory is a generalization of the Kalman-Bucy linear filtering theory. However, unlike the Kalman-Bucy theory, no blueprint for the black box which accepts as inputs the observations and produces as outputs the optimal estimate, the minimum loss estimate, can be deduced from the theory. In fact, in general, the black box, nonlinear filter corresponding to the optimal estimate is infinite dimensional in that the state of the nonlinear filter is not finite dimensional. For important technological problems, the signal process estimators generated by linearization and application of the linear Kalman-Bucy design often exhibit poor performance, sometimes even divergence. Further, even when the linear design seems effective, one is interested in what estimator has the best possible performance and what this performance is, in order to know whether further effort on this optimal design is justified. Finally, study of the optimal estimator allows one to generate effective suboptimal designs.

For the above reasons, it became clear that building nonlinear filters was an important task. Now it is quite easy to see, for example see [10], that the black box, determining the optimal filter, is specified by the maps with domain at time n , the observation process sample path up to and including time n and range, the conditional probability of the signal at time n given the observation process sample path up to and including n , $J_n(\cdot, \underline{z}_0, \dots, \underline{z}_n)$, with \underline{z}_i the observations.

Our problem then has two phases, one the replacement of $J_n(\cdot, z_0, \dots, z_n)$ by a finite vector \underline{J}_n , the representation problem, and second, the realization problem which is the means used to generate \underline{J}_{n+1} from \underline{J}_n and z_{n+1} .

The realization tool which has been used has been digital and hybrid systems. The problems considered have been the cubic sensor, passive receiver and the one, two and three state dimensional phase demodulation problem. These problems have been chosen for their importance, but also because the state dimension was low, a necessity if sufficiently precise estimates of the error performance of the optimal filter are to be generated. This is because error performance can so far only be evaluated by Monte Carlo simulation. This is also true for suboptimal filters. The problems which we have studied the most are the cubic sensor problem and the two-state dimensional phase demodulation problem; see [4], [11], [12], & [13].

The representations considered for the two-dimensional phase demodulation problem have been: point mass, Gauss-Hermite polynomials, Fourier Series, and Cubic Splines under tension.

On the basis of the structure of the equations to be solved in order to build an optimal nonlinear filter, it became clear early in our synthesis research effort that parallel or associative digital computers offered considerable speed-up in estimate production over standard serial machines of third generation; see [1] and [4] where these effects are discussed. Because of our lack of access to machines of the parallel or associative type in the early 1970's, a parallel processor was constructed using a contemporary hybrid system. In [2] the feasibility was considered, while in [3], [5] the construction of the actual system was reported along with

subsequent Monte Carlo runs, all for a one-dimensional signal process cubic sensor problem. The results were encouraging; in particular the hybrid realization proved to be 16 times faster than an all-digital realization, using the digital portion of the hybrid system. Our ultimate aim, however, was to study the effect of using CDC Star 100 and Illiac IV as synthesis tools for the nonlinear filter. This paper is devoted to detailing the impact of these machines on the nonlinear filtering problem of phase demodulation.

In the summer of 1975, we obtained access to the Illiac IV and made plans to access the CDC Star 100, when it was delivered to NASA Langley Research Center. It was decided that in order to effectively be able to judge the improvement, we should initially realize a problem, the two-dimensional phase demodulation problem where we had extensive numerical results and experience on the serial CDC 6600. Further for this problem we could generate estimate and signal sequences with an existing serial 6600 program. We were convinced that this serial program was close to the fastest possible, and that that was not the case, was an interesting byproduct of producing an effective Star program. While studying this two-dimensional phase demodulation problem on the Illiac and the Star gave us good data on speed-up factors possible with pipeline and parallel machines, our real objective was to do a problem which was beyond the capabilities of third generation serial machines. Consequently, we developed a generalized three-state dimension demodulation problem where phase, phase rate and amplitude all must be estimates. In the Fall and the Spring of 1975-1976, Ken Senne developed a two-dimension optimal phase demodulator program for the Illiac IV using the Glypnir language and the Arpa net. In the month of June, the authors

were visiting scientists at ICASE, NASA Langley. During this time, the corresponding program for the Star 100 was developed. This latter program not only was very effective for Star (a large percentage of operations were of long vector type, over 4,000 component vectors, consequently assuring that Star was doing long streaming), but also this Star program was used as a prototype of an extremely fast serial program which was over 2-1/3 times faster on the 6600 than our original program reported in [4]; also see the listing of the cyclic point mass program. Finally, a Star program for a three-dimensional signal process, combining amplitude and phase estimation, was developed and time comparison between it and the corresponding serial version obtained.

Methods of obtaining extremely fast serial realization of the optimal filter are being investigated using optical and surface wave devices. It is curious that the mathematical mapping which permits a time correlator surface wave device to calculate multi-dimensional convolutions is intrinsic in the Star program--compare [9]. We will discuss later in this paper the possible application of array processors coupled with minicomputers to this problem.

In [10], a discussion is given comparing the speed-up possibilities inherent in realization method versus density representation. This paper is concerned only with realization methods and uses essentially only the point mass representation. Other representations can be used, but in general they must be accuracy calibrated against the point mass method, and the effect of pipeline and parallel realization on estimate speed-up is unclear.

2. COMPUTER REALIZATION SURVEY

2.1 The Physical Problem

It has been observed in previous studies ([4]) that nonlinear filters can be efficiently implemented on parallel computers. Since every so-called vector machine has unique limitations, however, it is necessary to adapt the algorithm to each particular architecture. Candidate architectures for the present study include the array processor (Illiac IV), the pipeline processor (CDC-Star 100), and the look-ahead processors (CDC 6600 and 7600). Benchmarks on fast serial machines (IBM 370-168 and PDP 11-70) have also been included for reference.

An interesting application for optimal nonlinear estimation was introduced by Mallinckrodt, Bucy, and Cheng [7], who considered the problem of tracking a first-order phase process based on measurements of a modulated signal in noise of the form

$$ds(t) = A \cos [\omega_0 t + x_1(t)] dt + dv(t) \quad (2.1)$$

where A is a known amplitude, ω_0 is a known carrier frequency, and $x_1(t)$ is the message process being tracked. The measurement noise is assumed white. Using a voltage-controlled oscillator, the known carrier may be removed by heterodyning down to base band, producing both in-line and quadrature components and resulting in an equivalent two-dimensional measurement process of the form

$$\begin{bmatrix} dz_1(t) \\ dz_2(t) \end{bmatrix} = \begin{bmatrix} \cos x_1(t) \\ \sin x_1(t) \end{bmatrix} dt + \begin{bmatrix} dv_1(t) \\ dv_2(t) \end{bmatrix}, \quad (2.2)$$

where A has been taken as unity without loss of generality, and the noise has been replaced by a vector of mutually independent quantities.

The first-order phase process studied in [7] consisted of Brownian motion with increment of length h having variance qh . In this paper, we describe a study of a second order phase process involving the integral of Brownian motion, expressed as

$$\begin{bmatrix} dx_1 \\ dx_2 \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} dt + \begin{bmatrix} 0 \\ 1 \end{bmatrix} dB_t \quad (2.3)$$

We will retain the same measurement model (1) and let the noises v_1 and v_2 be independent Brownian motions with paths of length h having variance rh .

The optimal filter is specified by $J_{n+1|n+1} \begin{pmatrix} y_1 \\ y_2 \end{pmatrix}$, the conditional density of the phase and phase rate at time $(n+1)\Delta$, given discrete observations up to this time at sampling interval Δ . The following equations determine this density:

$$J_{n+1|n+1} \begin{pmatrix} y_1 \\ y_2 \end{pmatrix} = D_{n+1}(y_1) \int_{-\infty}^{\infty} \exp \left\{ -\frac{(y_2 - \mu)^2}{2q\Delta} \right\} J_{n|n} \begin{pmatrix} y_1 - \mu\Delta \\ \mu \end{pmatrix} d\mu, \quad (2.4)$$

where

$$D_{n+1}(y_1) \triangleq C_0 \exp \left\{ \frac{z_1(n+1) \cos y_1 + z_2(n+1) \sin y_1}{r/\Delta} \right\}. \quad (2.5)$$

It can be shown--see [4]-- that a modulated density $\tilde{J}_{n|n}$ defined as

$$\tilde{J}_{n|n} \begin{pmatrix} \sigma \\ \tau \end{pmatrix} \triangleq \sum_{k=-\infty}^{\infty} \sum_{l=-\infty}^{\infty} J_{n|n} \begin{pmatrix} \sigma + 2\pi k \\ \tau + \frac{2\pi l}{\Delta} \end{pmatrix}. \quad (2.6)$$

with $-\pi \leq \sigma < \pi$, and $-\pi/\Delta \leq \tau < \pi/\Delta$, carries all the information necessary for nonlinear filtering when the cyclic loss function $\frac{1}{2}(1 - \cos \varepsilon_1)$ with ε_1 the phase error, is used. The modulated density satisfies

$$\tilde{J}_{n+1|n+1} = D_{n+1}(\sigma) \int_{-\pi/\Delta}^{\pi/\Delta} a(\tau-\xi) \tilde{J}_{n|n} \left(\begin{matrix} \sigma-\xi\Delta \\ \xi \end{matrix} \right) d\xi, \quad (2.7)$$

where

$$a(\tau-\xi) \triangleq \sum_{i=-\infty}^{\infty} \exp \left\{ - \frac{(\tau-\xi+2\pi i/\Delta)^2}{2q\Delta} \right\} \quad (2.8)$$

We recall that the cyclic estimate, the one that minimizes the cyclic loss, is given by

$$x_n^* = \arg E e^{ix_1(n)} \Big|_{z_i^1, z_i^2, i \leq n} \quad (2.9)$$

2.2 Point Mass Placement

The most attractive formulation for numerical solution of the filtering problem on parallel machines involves the use of point-mass representation of the densities. A fixed rectangular grid will consist of m points in the phase variable and n points in the phase rate, with coordinates (i, j) corresponding to (σ, ξ) by the formulae

$$\sigma(i) = -\pi + 2\pi \left(\frac{i}{m} \right) + \pi \left(\frac{1}{m} \right) = \pi \left(\frac{1+2i}{m} - 1 \right), \quad i = 0, \dots, m-1 \quad (2.10)$$

$$\xi(j) = -\pi/\Delta + 2\pi/\Delta \left(\frac{j}{n} \right) + \pi/\Delta \left(\frac{1}{n} \right) = \pi/\Delta \left(\frac{1+2j}{n} - 1 \right), \quad j = 0, \dots, n-1 \quad (2.11)$$

Note that the phase rate variables may be used directly in the convolution

(2.7) but the phase variables must be interpolated in general, so that

$$n(k) = \sigma(i) - \Delta\xi(j) \Rightarrow k = i + \frac{m}{2} - \frac{m}{n} \left(\frac{1}{2} + j \right) \quad (2.12)$$

If m is taken as an even integer, such that n/m is an integer, then (2.12) may be decomposed into a subdominant integer $[k]$ and a fractional part Δk , as follows:*

$$[k(i)] = i + \frac{m}{2} - \left(\frac{1}{2} + j \right) \text{DIV} \left(\frac{n}{m} \right) - 1 \quad (2.13)$$

where DIV denotes integer division. Note that the addition of $\frac{1}{2}$ in (2.13) will not change the result since n/m is an integer. Next, we observe that the remainder after the division in (2.13) is the fractional part Δk :

$$\Delta k = \left(\frac{1}{2} + j \right) \text{MOD} \left(\frac{n}{m} \right), \quad j = 0, \dots, n-1 \quad (2.14)$$

or
$$1 - \Delta k = 1 - \left(\frac{1}{2} + j \right) \text{MOD} \left(\frac{n}{m} \right).$$

The interpolated result desired will be between $[k(i)]$ and $[k(i)] + 1$, (evaluated modulo m) with weighting Δk on the former and $1 - \Delta k$ on the latter. If the convolution is formed for fixed phase (i), then

$$[k(i+1)] = [k(i)] + 1 \quad (\text{modulo } m) \quad (2.15)$$

The relation (2.15) suggests the recursion which will be used on the Illiac. (2.15) will be used to iterate, based on an initial condition

$$[k(\)] = \left(\frac{m}{2} - 1 - j \text{DIV} \frac{n}{m} \right) \text{MOD } m \quad (2.16)$$

* Note that $[k]$ should be interpreted as modulo m .

During such iteration, $i, \Delta k$ will be used to weight the term evaluated at $[k(i)]$, while $1 - \Delta k$ will be used to weight the term evaluated at $[k(1 + i)]$.

$$J_{n|n}^i \begin{pmatrix} n(k) \\ \xi(j) \end{pmatrix} \approx (1 - \Delta k) J_{n|n} \begin{pmatrix} n([k(i+1)]) \\ \xi(j) \end{pmatrix} + \Delta k J_{n|n} \begin{pmatrix} n([k(i)]) \\ \xi(j) \end{pmatrix} \quad (2.17)$$

2.3 Evaluating the Filtered Data Term

The term (2.8) consists of an infinite sum. The values which Γ may take on in the grid coordinates consist of integer multiples of $2\pi/n\Delta$, resulting in

$$a(\Gamma(p)) = \sum_{\ell=-\infty}^{\infty} \exp \left[-\frac{2}{q\Delta} \left(\frac{\pi}{\Delta} \right)^2 \left(\frac{p}{n} + \ell \right)^2 \right] \quad p = -2(n-1), \dots, 2(n-1) \quad (2.18)$$

It may be seen from (2.18) that $a(\cdot)$ is an even function which is cyclic, modulo n . Further, for reasonable values of q the contribution of all terms except $\ell = 0$, is negligible. Thus, we compute

$$a(\Gamma(p)) = \exp \left[-\frac{2}{q\Delta} \left(\frac{\pi}{\Delta} \right)^2 \left(\frac{|p|}{n} \right)^2 \right] \quad (2.19)$$

$$0 \leq |p| \leq n - 1.$$

In the examples described below, 5 terms of $a(\cdot)$ were taken to be non-zero for the computing convolutions.

2.4 Assessment of Required Computations

The computations required to implement the point mass filter are summarized in Table 2.1, as a function of m and n . The sensor terms are the only ones which require math functions (exponentials). Since only m exponentials are required, this computation is generally insignificant compared with the overall filter update, so no special effort has been expended to optimize the required computations.

TABLE 2.1 ARITHMETIC OPERATIONS FOR FILTER

Function	Number of Operations			
	Multiplies	Adds	Divisions	Exponentials
Sensor Terms	$2m$	m	0	m
Interpolation	mn	$2mn$	0	0
Convolution	$5mn$	$10mn$	0	0
Row Sums	0	$mn-m$	0	0
Estimates	$3m$	$3m-3$	1	0
Normalization	$mn+m$	0	0	0
Total	$7mn+6m$	$13mn+3m-3$	1	m
Example	28864	53341	1	32
$m = 32$				
$n = 128$				

2.5 The Illiac IV Algorithm

The primary considerations for programming the Illiac IV array are the proper utilization of all of the 64 Processor Elements (PE's) and minimization of data routing between PE's. In order to accomplish the efficient use of the PE's, the values of $m = 32$ and $n = 128$ were selected and utilized for all machine examples. On the Illiac, two rows of PE storage were used for each value of the phase samples in JN. The interpolation and convolution were accomplished in a totally parallel fashion, using all 64 PE's, except for the operation depicted in Figure 2.1. It is necessary to perform a cyclic rotation of each phase row to accumulate the terms for the convolution (a cyclic rotation to the right is combined with a cyclic rotation to the left at each step). Since a cyclic routing on Illiac IV involves only one row at a time, it was necessary to form the two-row rotation by two single-row rotations, followed by an end element switch (involving three transfers with only one PE enabled).

The overall effectiveness of the Illiac IV algorithm is evident from the fact that so few operations are required which involve fewer than 64 PE's enabled. The sensor terms are computed with only 32 PE's enabled, but this operation only involves about 5% of the estimate update. The single PE transfers in the convolution are also only responsible for about 5% of the computation time. Finally, the row sums are done logarithmically with a PE utilization efficiency of 16.7%, and they account for another 5% of the estimate computation time. Thus, the net efficiency of this algorithm is 87.5%, or the equivalent of 56 times faster than a single PE program.

The Illiac program was coded in the Glypnir language, utilizing assembly language listings to reduce the overhead computations. The resulting

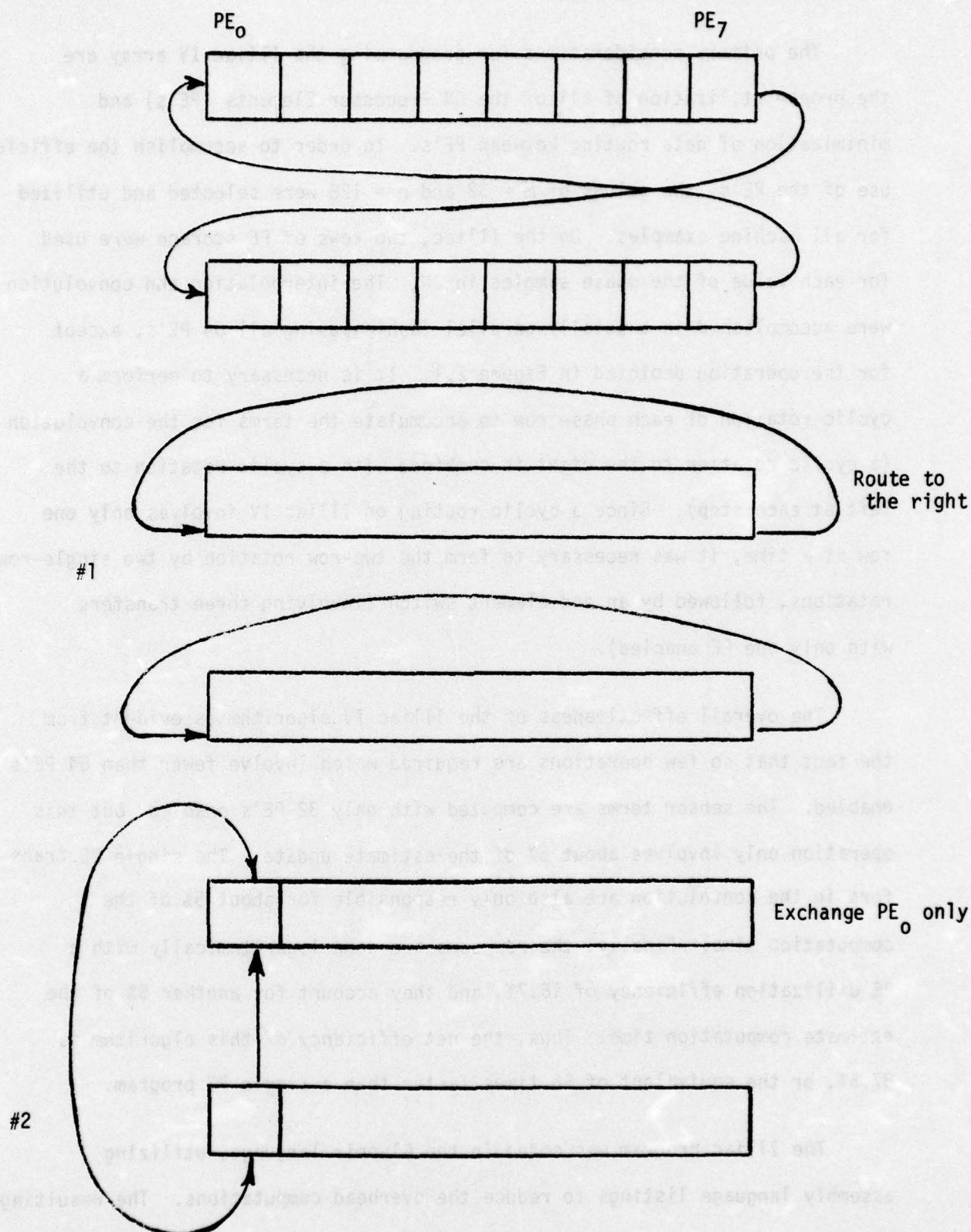


FIGURE 2.1

program was timed in the non-overlap mode, wherein the Control Unit (CU) and the PE's operate serially and considerable overhead cycles are required to allow error conditions to ring out between instructions. Thus, although we would expect the Illiac program to run at least twice as fast as the equivalent program for the CDC-STAR, in fact it ran five times slower. It will be reported in [17] how the program operates in overlap mode. It is also important to indicate that the Illiac IV Clock is running at 80 nsec. as compared with the design goal of 50 nsec.

2.6 The CDC-STAR Algorithm

The pipeline architecture is unconstrained by small fixed resources (i.e., 64 processors). On the other hand, efficient utilization of the pipeline requires detailed attention to pre-arrangement of vectors to allow for streaming from consecutive memory locations. This consideration is particularly important for STAR, which has a relatively slow memory cycle time. Since the nonlinear filter is recursive, it is necessary to include the vector re-arrangement as part of the filter update and therefore the re-arrangement constitutes the major overhead of the STAR program. The operations on the matrix JN to precondition it for the convolution are shown in Figures 2.2 and 2.3. First, the column-ordered JN matrix is column-shuffled with itself to produce a matrix which has two copies of every phase variable in each column. Then, a scrambled JN can be formed which has the property that each row in the final convolved matrix can be generated by operating on a suitable interpolation between the two adjacent rows of the scrambled JN. The interpolation which does this is depicted in Figure 2.3.

*The Figures 2.3 - 2.4 are shown with $m = 4$ and $n = 16$ for illustrative purposes only.

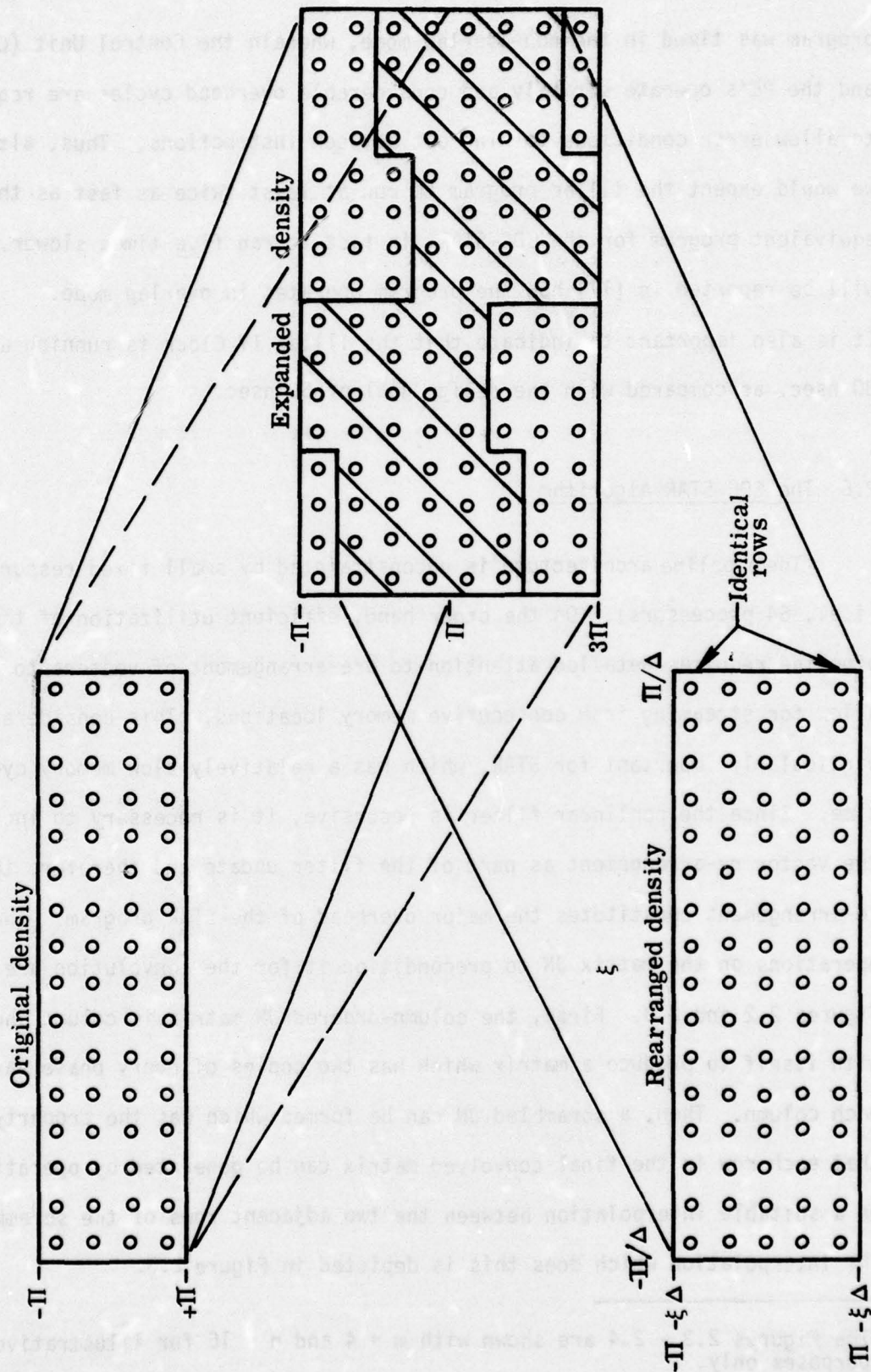


Figure 2.2.- Scrambling phase variable prior to convolution.

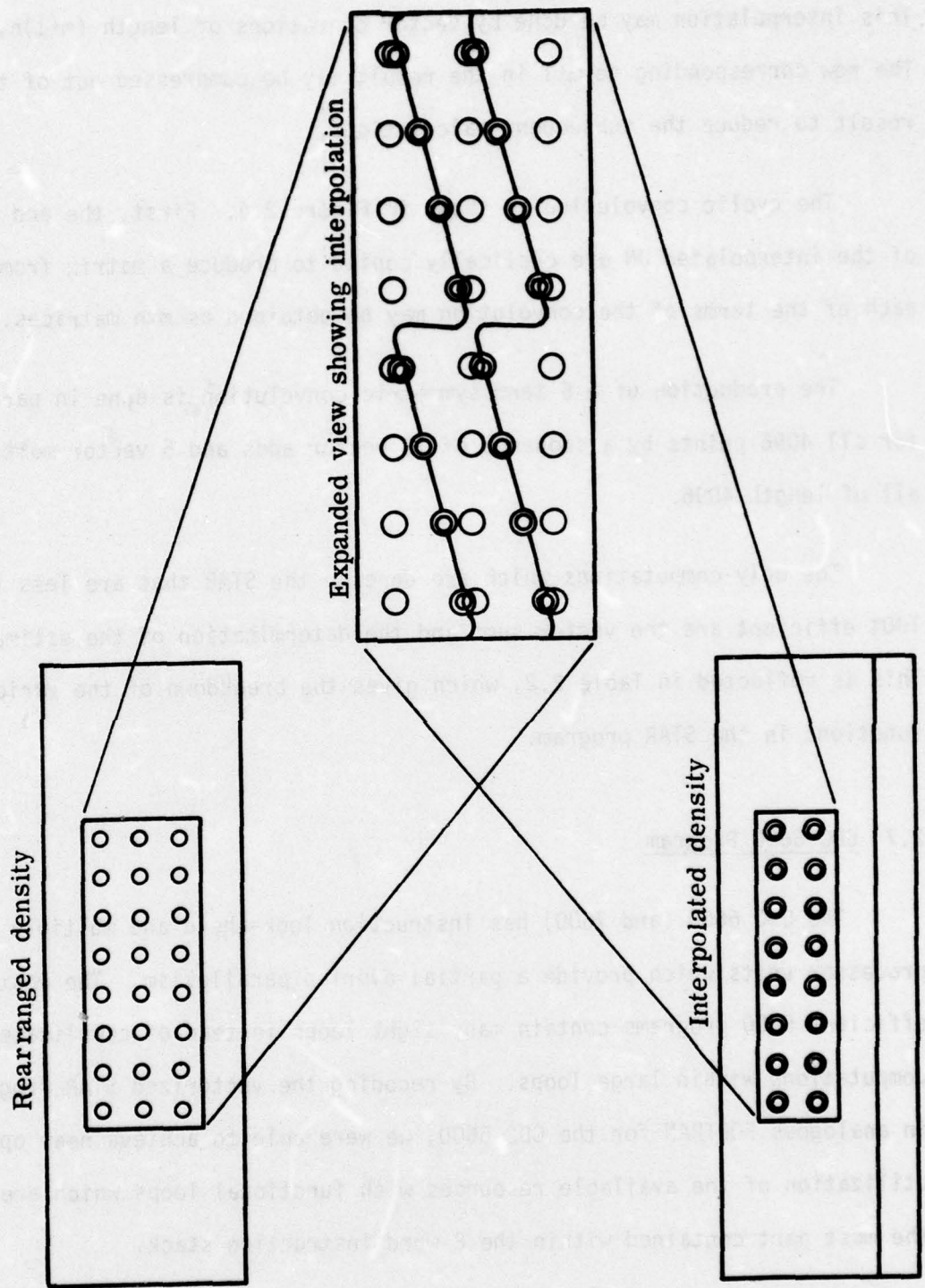


Figure 2.3.- Interpolation of scrambled matrix.

This interpolation may be done by vector operations of length $(m+1)n$, or 4224. The row corresponding to $m+1$ in the result may be compressed out of the final result to reduce the subsequent calculations.

The cyclic convolution is shown in Figure 2.4. First, the end columns of the interpolated JN are cyclically copied to produce a matrix from which each of the terms of the convolution may be obtained as $m \times n$ matrices.

The production of a 5-term symmetric convolution is done in parallel for all 4096 points by a sequence of 10 vector adds and 5 vector multiplies, all of length 4096.

The only computations which are done on the STAR that are less than 100% efficient are the vector sums and the determination of the estimates. This is reflected in Table 2.2, which gives the breakdown of the various functions in the STAR program.

2.7 CDC 6600 Program

The CDC 6600 (and 7600) has instruction look-ahead and multiple arithmetic processor units which provide a partial overlap parallelism. The most efficient 6600 programs contain many tight loops instead of complicated computations within large loops. By recoding the vectorized STAR program in analogous FORTRAN for the CDC 6600, we were able to achieve near optimal utilization of the available resources with functional loops which are for the most part contained within the 8-word instruction stack.

The basic data flow of the CDC 6600 is illustrated in Figure 2.5. Reads from memory are accomplished by setting the A Registers A1 - A5 with

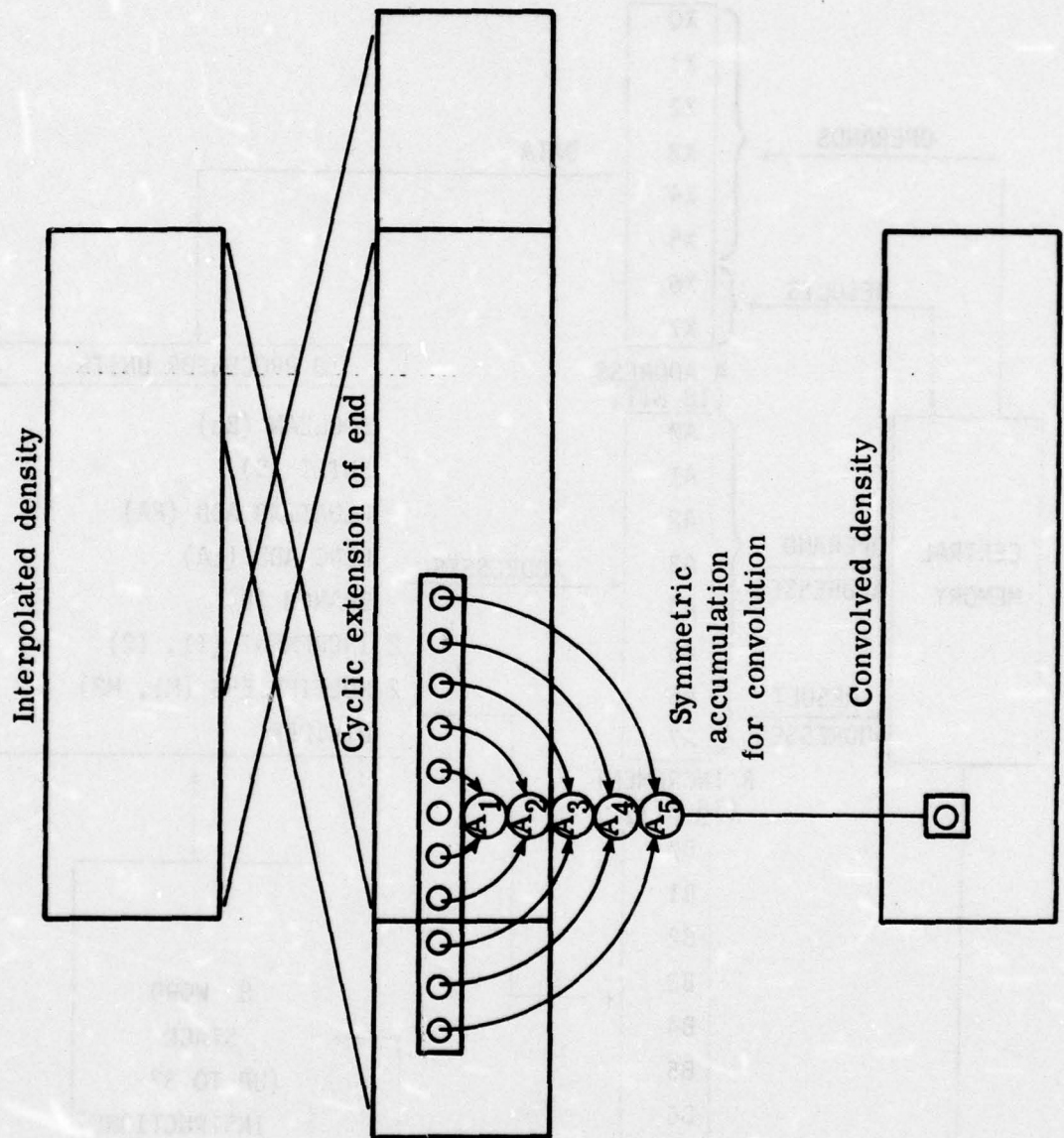


Figure 2.4.- Convolved density.

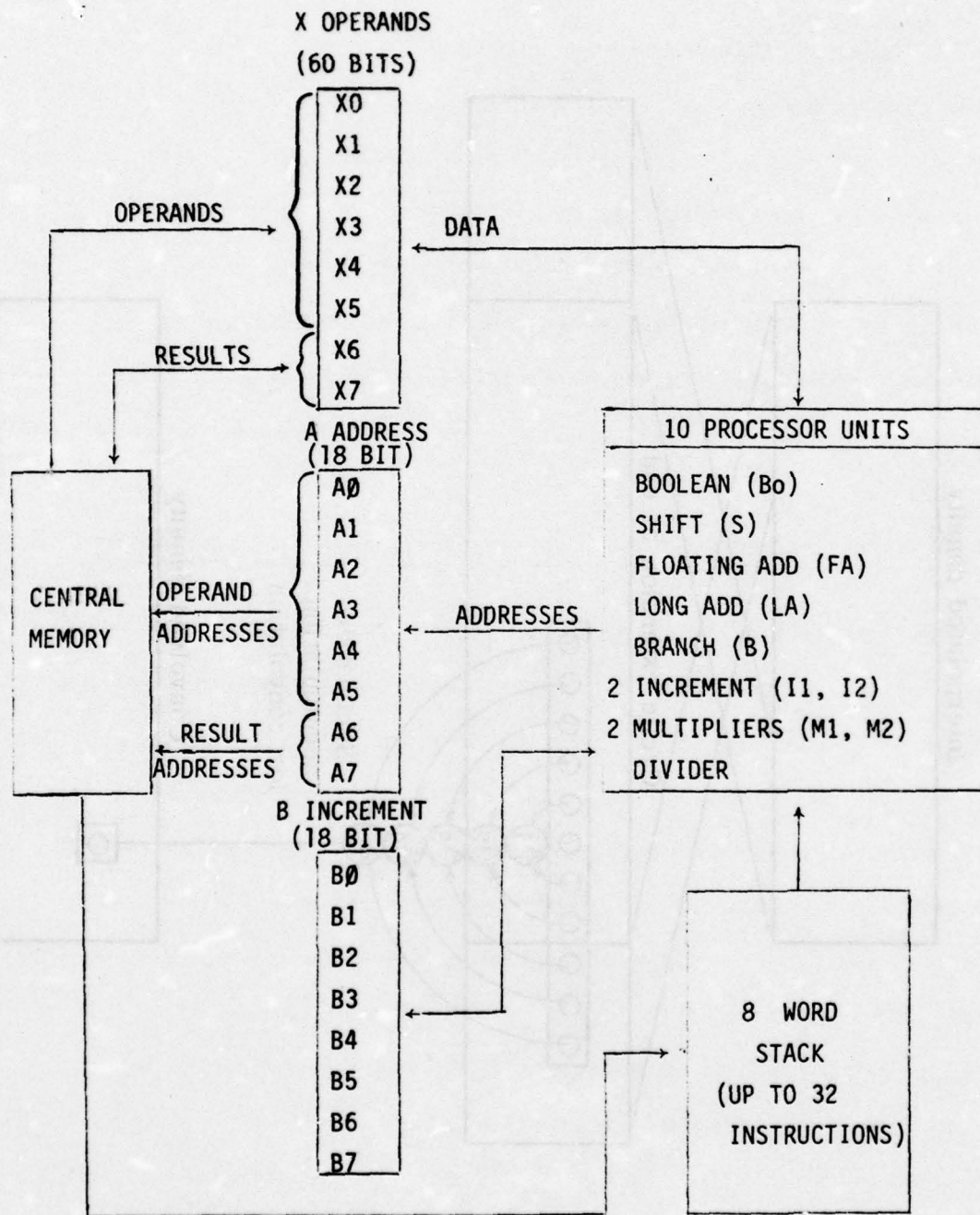


Figure 2.5. CDC 6600 CPU Architecture

TABLE 2.2 -- STAR PROGRAM BREAKDOWN

Operations	% of Total	Start-ups	Cycles Compute	Required
Vector Arithmetic	63.4	5174	76793	78.5%
12 Multiplies	33.3	1908	41152	70.1%
46 Adds	26.4	3266	30861	86.4%
1 Exponential	3.7	-	4780	100.0%

Vector Rearrangement	31.3	1233	39292	83.7%
1 Vector Transfer	5.8	1001	6464	68.0%
2 Indexed Transfers (Block Lengths 32 & 33)	22.1	144	28480	100.0%
1 Vector Compress	3.4	88	4348	0%

Scalar Arithmetic	-	-	79	100.0%
1 Divide			46	
3 Adds			33	

Subroutine Overhead	4.1	-	5142	0%
3 Calls				

Miscellaneous	1.2	-	1537	100.0%
Memory Conflicts, etc.				

TOTAL		6407	122843	=> 5.17msec
		5%	95%	
Minimum Achievable		6224	111296	=> 4.70msec

the appropriate address; writes are obtained by loading A6-A7. The B Registers are used for incrementing and address computation.

The breakdown of the computations for the CDC 6600 is shown in Table 2.3. Note that the major overhead is for reading and writing and miscellaneous waiting. It is interesting to note also that the ratio of multiply rates between STAR and 6600 is 25 to 1, while the add rate ratio is 17.5 (including normalization on 6600). Thus, for arithmetic alone, STAR would be expected to be 21 times faster than the 6600 on this problem. The achieved speed-up of 29 is explained by the slightly lower overhead on STAR (36.6% versus 52.9% for 6600).

2.8 Implications for Future Synthesis of Higher Dimensional Filters

The main purpose of this study was to measure the speed-up achievable by parallel and pipeline machines, in order to determine the feasibility of realizing 3 and 4 state dimensional nonlinear filters. In a later section of this paper we describe the fairly natural extension of our 2-state dimensional filter STAR 100 program to a 3-state dimensional problem of estimating phase, phase rate and amplitude. One outgrowth of the software development was a determination of machine structure limitations for natural synthesis of higher dimensional filters. Basically, we found that Illiac IV was limited to the two-dimensional problem because of limited P.E. memory and the availability of only 64 PE's, and most importantly, because of the increase of program complexity and overhead with dimension. The STAR 100 program also becomes complex to program and overhead prone, but at a somewhat higher dimension with the basic limitation that the total number of grid points be

TABLE 2.3-- CDC 6600 PROGRAM BREAKDOWN

Function (% Cycles)	Operations					Notes
	Multiplies	Adds	Divisions	Reads	Writes, Exp.	
Sensor Update (~0)	2m	m	0	3m	m	m Not in stack (extr. refs.)
Interpolation (~16%)	mn	2mn	0	4mn+3n	mn	0 Outer Loop not in stack
Expansion (~0)	0	0	0	10m	10m	0 Instack
Convolution (~74%)	5mn	10mn	0	16mn+15	6mn	0 Instack
Row Sums (~4%)	0	mn-m	0	mn+m	m	0 Instack
Estimates (~0)	3m	3m-3	1	4m+11	3	0 Instack
Normalization (~6%)	mn+m	0	0	mn+2m	mn	0 Instack
Approx. Totals	7mn	13mn	0	22mn	8mn	0
Approx. Number of Minor Cycles	70mn	91mn	0	66mn*	24mn*	0

Summary:	Arithmetic:	161mn	47.1%
Minor Cycles	Read/Write:	90mn	26.3%
(Approx.)	Waiting:	91mn	26.6%

Measured: 342mn 100.0%

(m = 32, n = 128 for test case) Measured Time = 140 msec/est.

*To some extent these cycles are concurrent, but the total of read/write and waiting of 52.9% is accurate.

less than 65K, the maximum length of vector for Star vector operations, as we show later, a 3-state problem is feasible on Star.

The implications for future machine design suitable for our problem are a pipeline machine which can compute with indexed vectors and with reduced operation cycle time.

3. TIME FACTORS FOR 2-DIMENSIONAL PHASE DEMODULATION

In [4] a complete description of the two-state dimensional phase demodulation problem is given as well as numerous numerical results and a listing of the cyclic nonlinear filtering program. Because we wished to compare the Illiac, Star and 6600 directly, the grid of 32×128 was chosen, i.e., 32 subdivisions in phase and 128 subdivisions in phase rate, to represent J_N the conditional density of phase and phase rate given the observations. This grid was natural for the Illiac as it has 64 parallel channels so that the integral equation

$$J_n(x,y) = \frac{D_n(x)}{k} \int_{-\pi/\Delta}^{\pi/\Delta} S(y-\xi) J_{n-1}(x-\xi\Delta, \xi) d\xi \quad (3.1)$$

$$y \in (-\pi/\Delta, \pi/\Delta) \quad x \in (-\pi, \pi)$$

with the values of $J_n(x_i, \cdot)$ could be found in two passes; the first simultaneously giving $\{J_n(x_i, y_j)\}_{j=1 \dots 64}$ and the second giving simultaneously

$\{J_n(x_i, y_j)\}_{j=65 \dots 128}$ and then repeated for each i . We knew before

that this mesh was fine enough for accuracy. The serial cyclic point mass filter with interpolation required .700 seconds per estimate.

The Illiac on the other hand produced estimates every .0308 second. The Illiac was not run at full capacity for our runs and we were unable to run it under the fastest possible conditions because other jobs had tied up the machine in August, so that our timing should be looked at as an upper bound to actual performance, clearly far away from the predicted theoretical performance. It is unclear whether in the production mode Illiac can achieve anywhere near theoretical performance, i.e., 64 times faster than CDC 6600 on a full parallel job. It may be that Illiac hardware cannot be made to operate at its theoretical speeds.

As described in the previous section for the Star 100 program, the density was carried as a $32 \times 128 = 4096$ component vector and the convolution developed as a sum of scalar times translates of this vector. This was done in order to take advantage of the pipeline speed of Star. At first with a direct Star Fortran program, a rate of .0118 second per estimate was obtained. By examination of the assembly language version of the Star program and timing each element, we replaced certain Star Fortran calls by more efficient assembly language routines, e.g. VGATHER and VSUM were replaced. After this was done, we achieved a rate of .0049 second per estimate. From this latter number it became clear that the serial 6600 program was not efficient. We were exceeding the factor of 25 theoretical speed ratio of Star to the 6600. This ratio is the ratio of Star multiplication time to 6600 multiply time. A new program, Starrun, was written which was the serial version of the Star program. Starrun achieved .178 second per estimate with the FTN, OPT = 2, Level 410 Fortran compiler. The FTN 4.5 compiler produces significantly different and faster assembly language

programs than the earlier optimizing compilers. By elimination of some multiplications and arranging things so that everything is in the stack, the Starrun program was made to run at .14 second per estimate. The following table gives the timing of our programs on various machines. With considerable effort in each case devoted to generating fast code:

<u>Machines</u>	<u>Milliseconds Per Estimate</u>	<u>Compiler</u>
CDC 6600	137.0	FTN 4.5 Level 410 OPT=2
STAR 100	4.7	Assembly Language Optimization
IBM 370-168	130.0	H Compiler
ILLIAC IV	30.8	Glypnir Language, Assembly Listing Optimization
PDP 11-70	870.0	FORTRAN 4 plus
CDC 7600	25	FTN 4.5 Level 410 OPT=2

In order to get a feel for the relative speeds of the Star and the 6600, the ratio of multiply times, when the pipe on the Star is filled is 25 to 1, i.e., 6600 takes 10 minor cycles at 100 nano seconds per cycle, while a multiply on Star takes 1 cycle at 40 nanoseconds per cycle. However, the 6600 has several look-ahead features as well as a stack which can speed up the processing. The FTN, OPT=2 compiler seeks to put all loops in the stack and to fill all the registers so that the ratio we have found 28.57 to 1 is very reasonable for a problem such as ours which is structured so that vector operations can be done most of the time on Star. Of course, using Star for a problem which does not have parallel structure would be wasteful as one would do as well or better on the 7600.

4. BENCHMARK APPLICATION FOR DIFFERENT APPROXIMATIONS

It is very popular in engineering studies to come up with a compromise realization which approaches the accuracy of the optimum nonlinear filter with tremendous amount of computation savings. Without the existence of an accurate optimum scheme, which serves as a benchmark against which any other approximation schemes can be judged, little can be done. The development of such approximations is possible when the exact performance and amount of computations is known. Then and only then, the engineering compromise can be studied intelligently.

One of these compromise studies has recently been published--see [14]--as an approximation for the optimum two-dimensional phase demodulation. It can be generalized to three-dimensional phase-amplitude demodulation problems. This approximation is based on fitting the probability density function at each time step to a Gaussian distribution which produces the optimum first and second moments for the same error criterion. This means that the update is accurate and required only for the first and second moments. In order to be able to investigate the extension of this Gaussian approximation to the three-dimensional problem, it is essential to study very carefully the behavior of the optimum filter and establish a performance benchmark. This 3-dimensional problem was neither analyzed in the optimum way nor simulated on any digital computer before because of its size and complexity. A first attempt was carried out to realize the optimum demodulator on the CYBER 175 and CDC STAR-100 in order to establish boundary limits on the size of the program storage and the computation time.

The mathematical model, optimum filter and preliminary results for the 3-dimensional problem will be discussed in the following sections.

5. INTRODUCTION TO 3D-PROBLEM AND PRELIMINARY RESULTS

Achieving a good speed-up factor for the two-dimensional phase demodulation problem was the signal to move one step ahead and do the three-dimensional problem (phase, phase rate, and amplitude demodulation).

5.1 Mathematical Model

The three-dimensional state vector consists of phase x , phase rate y with additive white Gaussian noise u and amplitude A corrupted with another independent white Gaussian noise v additive to the amplitude mean value.

$$x_n = x_{n-1} + y_{n-1} \Delta$$

$$y_n = y_{n-1} + u_{n-1}$$

$$A_n = \alpha(a - A_{n-1})\Delta + A_{n-1} + v_{n-1}$$

where Δ is the time step

α is the damping constant

a is the amplitude mean value.

$$E u_n = 0$$

$$E u_n^2 = q_1 \Delta$$

$$E v_n = 0$$

$$E v_n^2 = q_2 \Delta$$

$$E x_0 = 0$$

$$E x_0^2 = c_1 \sqrt{2} q_1^{1/4} r^{3/4}$$

$$E y_0 = 0$$

$$E y_0^2 = \sqrt{2} q_1^{3/4} r^{1/4}$$

$$E A_0 = a$$

$$E A_0^2 = \frac{c_2 q_2}{2\alpha}$$

The observation z is modelled in the following way;

$$z_n^1 = A_n \cos x_n + w_n^1$$

$$z_n^2 = A_n \sin x_n + w_n^2$$

$w_n^{1,2}$ is white Gaussian noise independent of u_n, v_n with $E w_n^{1,2} = 0$

$$E(w_n^{1,2})^2 = \frac{r}{\Delta}$$

Δ is chosen to be the same as in the two-dimensional problem, i.e.,

$$\Delta = .1 \left[\sqrt{2} \left(\frac{r}{q_1} \right)^{1/4} \right]$$

c_1 is chosen to be 2.19 (same as 2-dimensional problem)

c_2 is chosen to be 4

$$\alpha = 1$$

$$a = 1$$

$$q_2 = .01$$

5.2 Optimum Filter

The optimum filter realization is carried out by a recursive update of the density J_n using the discrete form of the presentation theorem --see [4].

$$J_n(x, y, A) = K_n S_n(x, A) \int_{-\infty}^{\infty} \int_{-\pi/\Delta}^{\pi/\Delta} a_1(y-z) a_2(A - \beta\eta - \alpha\alpha\Delta) J_{n-1}(x-z\Delta, z, \eta) dz d\eta$$

$$\text{where } S_n(x, A) = \text{Exp} \left[\frac{\Delta}{r} \left[A(z_n^1 \cos x + z_n^2 \sin x) - \frac{A^2}{2} \right] \right]$$

$$a_1(y-z) = \sum_{k=-\infty}^{\infty} \text{Exp} \left[-\frac{1}{2q_1\Delta} \left(y-z + \frac{2\pi k}{\Delta} \right)^2 \right]$$

$$a_2(A - \beta\eta - \alpha\alpha\Delta) = \text{Exp} \left[-\frac{1}{2q_2\Delta} (A - \beta\eta - \alpha\alpha\Delta)^2 \right]$$

$$\beta = 1 - \alpha\alpha\Delta$$

K_n is normalizing constant chosen so that

$$\int_{-\infty}^{\infty} \int_{-\pi/\Delta}^{\pi/\Delta} \int_{-\pi}^{\pi} J_n(x, y, A) dx dy dA = 1$$

5.3 Optimum Estimates

The phase estimates are produced in the same manner as in two-dimensional problems. They are obtained by minimizing the error function $L_1(e) = 2(1 - \cos e)$

$$x_n^* = \arg \int_{-\infty}^{\infty} \int_{-\pi/\Delta}^{\pi/\Delta} \int_{-\pi}^{\pi} e^{ix} J_n(x, y, A) dx dy dA .$$

The amplitude estimates are produced by simply minimizing the quadratic loss function $L_2(e) = e^2$, leading to

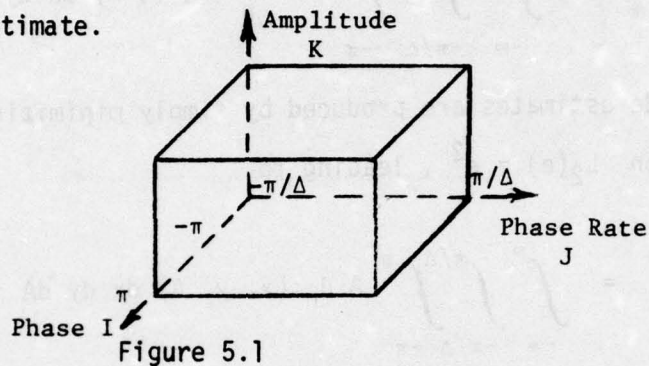
$$A_n^* = \int_{-\infty}^{\infty} \int_{-\pi/\Delta}^{\pi/\Delta} \int_{-\pi}^{\pi} A J_n(x, y, A) dx dy dA .$$

5.4 Simulation

A standard Fortran program was written to simulate the model described in 5.1 and produce optimum density and estimates as described in 5.2 and 5.3 respectively. The random number generator used is the same as in the two-dimensional problem. This program was debugged and executed on the next fastest computer available at NASA Langley --CYBER 175, which is faster than the CDC 6600 by a factor of approximately 3. Then a final program was written for the STAR which produced output which checked with the C175 results.

5.4.1 CYBER 175 Program -- A point mass realization of the conditional density was simulated by choosing $16 \times 96 \times 16$ masses located inside a box representing the three axes, the phase, phase rate and amplitude, respectively (see Figure 5.1). The 16×96 masses for the phase and phase rate are located

inside the rectangle bounded by $(-\pi, \pi]$ in phase and $(-\frac{\pi}{\Delta}, \frac{\pi}{\Delta}]$ in phase rate. The grid in the amplitude direction is floating according to the location of the prior estimate position. For every new amplitude estimate, the whole box will be centered on the amplitude estimate with eight masses equally distributed above the center and the same number below. The mesh size between the amplitude masses is chosen to be equal to half the square root of the variance of the estimate.



The reason we chose the number of discrete masses to be $16 \times 96 \times 16$ is that the maximum vector length for STAR operation is limited to 65K as well as storage limitations on the CYBER. Because we know from the two-dimensional problem that 16×96 masses were not accurate enough for density representation the purpose of the $16 \times 96 \times 16$ scheme is just to debug the program on the CYBER and try to produce the same results on the STAR later. This way, we can have a valid comparison for the speed of the STAR as compared to other computers for the same amount of computation.

The main structure of the program follows essentially the original program for the two-dimensional phase demodulator. The interpolative and scrambling matrices were precomputed once and for all, and called out inside the mass convolution loop for each iteration. Because of the choice of the dynamic model for the phase rate, $a_1(y - z)$ is symmetric and constant. The cut-off limit for the significant terms of $a_1(y - z)$ is determined by a radius

of length $= \sqrt{50 q_1 \Delta}$. This is not the case for the amplitude dynamics which does not have a symmetry because of the damping factor and all the sixteen terms have to be computed for every iteration. When this program was executed, it produced estimates every 12 seconds.

5.4.2 CDC STAR-100 Program-- The standard FORTRAN program in 5.4.1 was changed to adapt to the vector notations for the STAR. The three-dimensional array was swept in the right order I (phase) - J (phase rate) - K (amplitude). The main steps of the program follow the same structure of the two-dimensional STAR program, except for the added third dimension as follows:

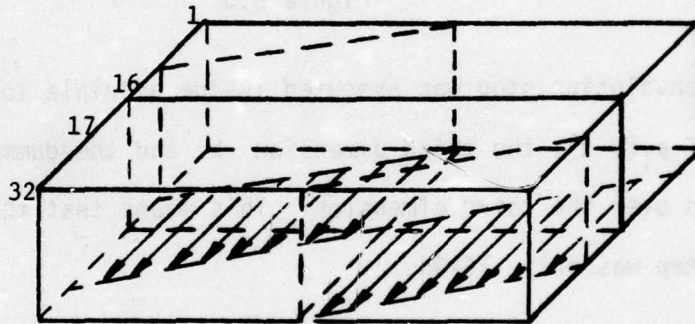


Figure 5.2

(i) In order to perform the scrambling according to the rule (I-J) inside the convolution in the most efficient way, a block mapping VXTOD is used with length 17 for 96×16 blocks--see Figure 5.2. Of course, that required an initial step of copying the box on itself by using the 'MERGE' COMMAND and carrying a copy of the three axis array for the density masses.

(ii) The interpolation between two successive rows was carried on the box of size $17 \times 96 \times 16$.

(iii) 'COMPRESS' COMMAND was finally used to get rid of the extra 17th element and to bring the box into its normal size 16×96×16.

(iv) At this step the box was expanded in J direction far enough to cover the maximum number of terms of $a_1(y-z)$ as shown in Figure 5.3.

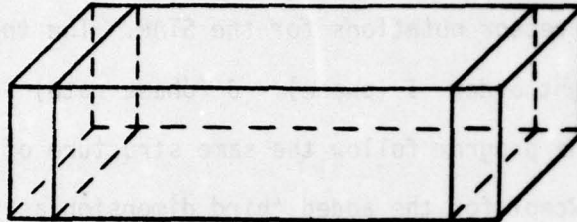


Figure 5.3

(v) The convolution step was executed inside a triple loop for the number of terms of $a_1(y-z)$, the third dimension k and the dummy variable for integration over the third dimension. This means that the length of vector for each step was only 16×96.

(vi) Multiplication by the sensor terms was done in one vector operation of length 24,576.

(vii) Normalization constant was obtained by folding the density successively on itself until it collapsed to single element.

(viii) The density then was multiplied by the inverse of the constant and transferred to the old density for new iteration.

(ix) This normalized density is used for producing estimates for phase and amplitude. It took 89 milliseconds cpu time for every iteration to produce the same results as on the CYBER 175. This present speed factor between the STAR and CYBER will be cut down drastically after converting the

standard FORTRAN program for the CYBER into a serial version in an analogous way to the STAR program where all the arrays are treated as single vectors.

6. EXTENSIONS AND FUTURE RESEARCH

As is clear from equation (3.1), the computational problem of realization of nonlinear filters in general consists of two tasks; first, the passage from the filter density to the one-step predictor density, a convolution task; and secondly, using the new piece of data to construct from the 1-step predictor density the next filter density. Independently of how the densities are finitely represented, these two partitions of the density update are always present. The convolution task is time consuming when done serially and calls for parallel or pipeline implementation. In general, strictly parallel devices are exorbitantly expensive as dimension and/or gridding increase.

Currently, we are investigating optical realization of the convolution task for two state dimensional problems with Drs. Steier and Sawchuck of the University of Southern California. The problem here is one of accuracy and loop closure. We hope to achieve with the optical methods the accuracy obtained with the hybrid realization. Another approach for the convolution task would be to map space on the line--see [9]--and achieve multi-dimension convolutions by using mega-hertz surface wave devices to realize single dimension convolutions; again, here the problem of loop closure is important and unsolved. However, this latter method is not confined to two-dimensions as the optical approach is.

Probably the most promising methods in terms of high-speed, low-cost systems are contemporary mini-computers used with special floating point array processors, with the convolution task essentially done in the array processor and the mini-computer used to accomplish the second task.

An example of such an array processor with theoretical floating point arithmetic speeds only 2.78 slower than the STAR 100 is the AP-120B of Floating Point Systems Inc. We plan to program and evaluate such a system in the course of the next year. Most probably, special purpose digital pipeline devices offer the most promise when accuracy is important, while optical and other analog devices would be useful for filtering problems when accuracy is not that important.

Another problem which merits attention is what are the interactions of other methods of representing the density besides the point mass used here, and the use of pipeline machines. It is not at all clear that either the Spline or Fourier filters can be made to run 30 times faster on the STAR than on the 6600; essentially because the vectors involved are a factor of 10 smaller and start-up times are no longer negligible.

Finally, there are two other contemporary machines which may offer interesting possibilities for producing fast nonlinear filters; they are the Cray machine and a vector machine produced by Texas Instruments. We hope to be able to report on these in the course of the year.

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REFERENCES

- [1] R. S. Bucy and K. D. Senne, "Digital Synthesis of Nonlinear Filters," Automática, 7 (1971), 287-298.
- [2] R. S. Bucy, M. J. Merritt, and D. S. Miller, "Hybrid Synthesis of the Optimal Discrete Nonlinear Filter," Stochastics, 1 (1974), 151-211.
- [3] L. Basañez, P. Brunet, R. S. Bucy, R. Huber, D. S. Miller, and J. Pagés, "A Hybrid Computer Optimal Filter," Proceedings of the Sixth Symposium on Nonlinear Estimation Theory and Its Applications, (1975), San Diego, California.
- [4] R. S. Bucy, C. Hecht, K. D. Senne, "An Engineer's Guide to Building Nonlinear Filters," Final Report SRL-TR-72-0004, Project 7904, Frank J. Seller Research Laboratory, USAF Academy, Colorado 80840 (1972).
- [5] L. Basañez, P. Brunet, R. S. Bucy, R. Huber, D. S. Miller and J. Pagés, "Simulation and Implementation of a Hybrid Computer Algorithm for an Optimal Nonlinear Filtering," Proceedings of the 9th Symposium on System Science, Honolulu, Hawaii (1976).
- [6] R. S. Bucy, A. J. Mallinckrodt, H. Youssef, "High Speed Convolution of Periodic Functions," to appear J.S.I.A.M. Applied Math.
- [7] R. S. Bucy, H. Youssef, "Optimal phase demodulation," I.E.E.E. Trans. Autom. Contr., AC-21, 5, 1976, 732-736.
- [8] H. Youssef, "Interpolative Spline Filters," PhD Thesis, Aerospace Eng. Dept., University of Southern California, 1975.
- [9] R. S. Bucy, H. Youssef, "Fourier Realization of the Optimal Phase Demodulator," Proc. 4th Symp. on Nonlinear Estimation Theory and Its Applications, San Diego, Western Periodicals, 1973, 34-38.
- [10] R. S. Bucy, "Linear and Nonlinear Filtering," Proc. I.E.E.E., 58, 1970, 854-864.
- [11] R. S. Bucy, J. Pages, "A Priori Bounds for the Cubic Sensor Problem," to appear.
- [12] R. S. Bucy, H. Youssef, "Dependence of the Optimal Phase Demodulator on Statistical Parameters," I.E.E.E. Trans. Autom. Contr., AC-20, 2, 1975, 259-260.
- [13] R. S. Bucy, C. Hecht, K. D. Senne, "New Methods for Nonlinear Filtering," Rev. Francais d'Automatique, J-1, 1973, 3-54.
- [14] H. M. Youssef, "Suboptimal Phase Demodulation," Proc. of 6th Symp. on Nonlinear Estimation and Its Applications, San Diego, 1975.

- [15] D. Casseres, "Illiic IV Machine Reference Manual for the Programmer," Inst. for Advanced Computation Doc. #PG-11700-0000-A, June 1975.
- [16] STAR-100 Computer System Reference Manual. Control Data Corporation Publication No. 60256000, 1975.
- [17] R. S. Bucy, K. D. Senne, H. M. Youssef, "Pipeline Software for Filtering, to appear.


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5  CONTINUE
  WRITE(6,651) Y1EST,Y2EST,ALP110,DELF,Q22C,NUM1,NUM2,NO2
651 FORMAT(* *,* CYCLIC INPUT*,4X,5P10.5,3I5)
  P110=10.**((ALP110/10.))
  QQ=Q22C**(.25)
  RX=(P110/(SQRT(2.0)*QQ))**(4.0/3.0)
  FTC=SQRT(2.0)*RX**(.25)/QQ
  DELT=DELF*FTC
  Q22=Q22C*DELT
  R11=RX/DELT
  P220=P110*SQRT(Q22C/RX)
  ISAMP=1
  NSAMP=0
  SUMP1=0.0
  SUMP2=0.0
  CONTINUE
  A11=10.**((ALP110+1.4)/10.)
  A22= P220
  KOUNT=1
  DELSQ=DELT**2
  PI=3.1415926536
  PI2=2.0*PI
  PIDLT=PI/DELT
  CONST=-2.0*PIDLT*PIDLT/Q22
  PINV=1.0/PI
  PI2DLT=2.0*PIDLT
  U1=NUM1
  U2=NUM2
  XLIM=.75*PI
  LIMNL=0
  LIMKB=0
  Q(1,1)=0.0
  Q(2,2)=Q22
  A=DELT*PINV*SQRT(10.0*Q22)
  IA=A+0.5
  IY2=U2/PI2DLT*SQRT(50.0*Q22) + .5
  CALL MLF(0,0,Z1,Z2,SHAT,CHAT,TNLF)
11 CALL MLF(1,0,Z1,Z2,SHAT,CHAT,TNLF)
  XHAT(1)=Y1EST
  XHAT(2)=Y2EST
  XHAT1=Y1EST
  XHAT2=Y2EST
  PN(1,1)=A11
  PN(2,2)=A22
  PN(1,2)=0.
  PN(2,1)=0.
  R=R11
  DEV1= SQRT(A11)
  DEN=1.0/(PN(1,1)+R)
  F(1,1)=1.0
  F(1,2)=DELT
  F(2,1)=0.
  F(2,2)=1.0
  DEV2= SQRT(A22)

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DEV3= SQRT(R11)
CALL GAUSS (JSEPD,DEV1,Y1EST,X1)
KOUNT=1
XDAT (KOUNT,1)=X1
CALL GAUSS (JSEED,DEV2,Y2EST,X2)
CALL GAUSS (JSEED,DEV3,COS (X1),Z1)
CALL GAUSS (JSEED,DEV3,SIN (X1),Z2)
DEVQ2= SQRT (Q22)
R=R11
WRITE (6,1509)
205 FORMAT (*0*,8X,*POSIT.*,5X,*POSIT. MOD 2 PI*,2X,*EST. POSIT.*,9X,
**Z1 AND Z2*,19X,*CYCLIC LOSS*,5X,* K-B EST. AND P11*)
GO TO 470
C ***** END BLOCK 1 *****
C ***** START BLOCK 2 *****
450 CONTINUE
X1=X1 + X2*DELT
XDAT (KOUNT,1)=X1
CALL GAUSS (JSEED,DEVQ2,X2,X2)
CALL GAUSS (JSEED,DEV3,COS (X1),Z1)
CALL GAUSS (JSEED,DEV3,SIN (X1),Z2)
C ***** RICCATI EQUATION UPDATE *****
PDUMY (1,1)=(R*(PN (1,1)+2.0*PN (1,2)*DELT)-PN (1,2)**2*DELSQ)*DEN
* + PN (2,2)*DELSQ
PDUMY (1,2)=PN (1,2)*(R-PN (1,2)*DELT)*DEN + PN (2,2)*DELT
PDUMY (2,2)=-PN (1,2)**2*DEN + PN (2,2) + Q (2,2)
PN (1,1)=PDUMY (1,1)
PN (1,2)=PDUMY (1,2)
PN (2,2)=PDUMY (2,2)
PN (2,1)=PN (1,2)
DEN = 1.0/(PN (1,1) + R)
C ***** END BLOCK 2 *****
C ***** START BLOCK 3 *****
470 CONTINUE
CALL NLF (1,1,Z1,Z2,SHAT,CHAT,TNLF)
WRITE (6,5687) TNLF
5687 FORMAT (F10.5)
CXHAT = ATAN2 (SHAT,CHAT)
367 PLOSS=2.0*(1.0-SQRT (SHAT**2+CHAT**2))
XDAT (KOUNT,2)=CXHAT
XDAT (KOUNT,3)=PLOSS
PNF (1,1)=PN (1,1)*R*DEN
PNF (1,2)=PN (1,2)*R*DEN
PNF (2,1)=PNF (1,2)
PNF (2,2)=PN (2,2) - PN (1,2)**2*DEN
C ***** FILTER UPDATE *****
SINF1=SIN (XHAT1)
COSF1=COS (XHAT1)
XHAT (1)=XHAT1+DEN*(-PN (1,1)*SINF1*Z1+PN (1,1)*COSF1*Z2)
XHAT (2)=XHAT2+DEN*(-PN (1,2)*SINF1*Z1+PN (1,2)*COSF1*Z2)
X1FF=XHAT (1)
34 CONTINUE
IF (X1FF.GT.PI) GO TO 88
IF (Y1FF.LT.-PI) GO TO 89
GO TO 90

```

```

88 X1FF=X1FF-PI2
GO TO 84
89 X1FF=X1FF+PI2
GO TO 84
90 CONTINUE
IF (ABS(CXHAT) .GT. XLIM) LIMNL=LIMNL+1
IF (ABS(X1FF) .GT. XLIM) LIMKB=LIMKB+1
***** PREDICTOR UPDATE *****
XHAT1=XHAT(1) + DELT*XHAT(2)
XHAT2=XHAT(2)
XDAT(KOUNT,4)=XHAT(1)
XDAT(KOUNT,5)=PNF(1,1)
X1MOD=X1
184 CONTINUE
IF(X1MOD.GT.PI) GO TO 188
IF(X1MOD.LT.-PI) GO TO 189
GO TO 190
188 X1MOD=X1MOD-PI2
GO TO 184
189 X1MOD=X1MOD+PI2
GO TO 184
190 CONTINUE
IKBSLP=0
X1F2=ABS(XHAT(1)-X1)
339 IF(X1F2.GT.PI) GO TO 340
GO TO 341
340 CONTINUE
X1F2=X1F2-PI2
IKBSLP = IKBSLP+1
GO TO 339
341 CONTINUE
ERRLF=ABS(X1FF-X1MOD)
ERRNL=ABS(CXHAT-X1MOD)
IF(ERRLF.GT.PI) ERRLF=ABS(ERRLF -PI2)
IF(ERRNL.GT.PI) ERRNL=ABS(ERRNL-PI2)
ERRDF=ABS(X1MOD-CXHAT)
WRITE(6,201) KOUNT,XDAT(KOUNT,1),X1MOD,XDAT(KOUNT,2),Z1,Z2,(XDAT
* (KOUNT,I),I=3,5)
201 FORMAT(*0*,I3,1X,1P3E14.6,4X,1P2E14.6,4X,1P3E14.6 /)
IF(KOUNT.EQ.NO2) GO TO 505
KOUNT=KOUNT + 1
GO TO 450
505 CONTINUE
SUMP=0.0
SUMC=0.0
DO 1501 I=31,NO2
XD=ABS(XDAT(I,1)-XDAT(I,2))
1498 CONTINUE
IF(XD.GT.PI) GO TO 1499
GO TO 1500
1499 XD=XD-PI2
GO TO 1498
1500 SUMP=(XD)*2+SUMP
SUMC=XDAT(I,3)+SUMC
1501 CONTINUE

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H=NO2-30
SUMP=SUMP/H
SUMC=SUMC/H
  XNSAMP=NSAMP
  XAA=XNSAMP+1.0
SUMP1=(SUMP+XNSAMP*SUMP1)/XAA
DSUMP1=ALOG10(SUMP1)*10.
WRITE(6,1508)
1508 FORMAT(*0*,5X,*NONLINEAR CYCLIC ESTIMATOR*)
WRITE(6,1511)SUMP1,DSUMP1
1511 FORMAT(*0*,*AVERAGE STATISTICAL VARIANCE =*,1PE13.6,10X,
* *AVERAGE COMPUTED VARIANCE =*,1PE13.6//)
SUMP=0.0
SUMC=0.0
DO 1601 I=31,NO2
  XD=ABS(XDAT(I,1)-XDAT(I,4))
1698 CONTINUE
  IF(XD.GT.PI) GO TO 1699
  GO TO 1700
699 XD=XD-PI2
  GO TO 1698
1700 SUMP=(XD)**2+SUMP
  SUMC=XDAT(I,5)+SUMC
1601 CONTINUE
SUMP=SUMP/H
SUMC=SUMC/H
SUMP2=(SUMP+XNSAMP*SUMP2)/XAA
DSUMP2=ALOG10(SUMP2)*10.
WRITE(6,1509)
1509 FORMAT(*0*,5X,*RE-LINEARIZED K-B FILTER*)
WRITE(6,1511)SUMP2,DSUMP2
NSAMP=NSAMP+1
IF(ISAMP.EQ.NO3)GO TO 2200
ISAMP=ISAMP+1
GO TO 11
***** END BLOCK 3 *****
1200 WRITE(6,2201)
1201 FORMAT(*0*,40X,*NORMAL COMPLETION*)
STOP
END
```

```

SUBROUTINE GAUSS(JS,SD,XM,X)
DIMENSION NST(2)
COMMON /RN/ N1, N2, MC, T1, T2
COMMON /GN/ TWOPI, J, XR(2)
IF (J) 10, 10, 20
10 J=2
TWOPI=8.*ATAN(1.)
NST(1)=102943
NST(2)=195617
XR(1)=BANF(NST,1)
GO TO 35
20 GO TO (30,40), J
30 J=2
XR(1)=BANF(NST,0)
35 XR(2)=BANF(NST,0)
X1=SQRT(ABS(-2.*ALOG(XR(1))))
XR(2)=TWOPI*XR(2)
XR(1)=X1*SIN(XR(2))
XR(2)=X1*COS(XR(2))
X=XR(1)*SD+XM
RETURN
40 J=1
X=XR(2)*SD+XM
RETURN
END

```

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```

FUNCTION BANF(NS,MODE)
DIMENSION NS(2), NC(2)
COMMON /RN/ N1, N2, MP, T1, T2
DATA M1, M2/244734, 159551/
C      MODE=0 TO CONTINUE, OTHERWISE RESTART WITH
C      INTEGER NUMBER NS(1)*2**13+NS(2)
IF (MODE) 10, 100, 10
10 N1=NS(1)
N2=NS(2)
T1=2.**(-18)
T2=2.**(-36)
MP=2**18
100 DO 200 I=1,2
GO TO (110,120), I
110 K=M2*N2
GO TO 190
120 K=M1*N2+M2*N1+KD
190 KD=K/MP
200 NC(I)=K-KD*MP
N1=NC(2)
N2=NC(1)
XN1=N1
YN2=N2
BANF=XN1*T1+YN2*T2
RETURN
END

```

```

SUBROUTINE NLF (MC, SAMP, Z1, Z2, SHAT, CHAT, TNLF)
INTEGER MC, SAMP
REAL Z1, Z2, SHAT, CHAT, TNLF
INTEGER I, I1, JC, J1, J2, K1, K2, KL, KH, NJ1, NJ2, NK1, NK2, NTERM,
1 NTERM32, NSIZE, NSIZE32
REAL A11, A22, CL, CNORM, CONST, CB, PI, PIDEI, Q22, SI, T, TT,
1 Y1EST, Y2EST, TEMP, TEMP1, TEMP2
REAL IN (8)
REAL TROW (32)
REAL COSY (32), SINY (32), SN1 (32), S1 (32), S2 (32), SIGMA (32)
REAL PSI (128), A (10), DELJ (128)
INTEGER JNS (4224)
REAL JN (4096), JN1 (4096), JO (4096), JNA (4756)
COMMON /PROB/ TWOPI, PI, ALP110, DELF, Q22C, Y1EST, Y2EST,
1 A11, A22, CONST, DEL, FTC, PIDEI, P110, RDEL, RK, QQ, Q22
COMMON /NLFC/ NC, NT, NTERM, NTERM32, S1, S2, SIGMA, PSI, A, COSY,
1 DELJ, JO, JNA, JNS, SINY
EQUIVALENC (JN1 (1), JNA (321))
IF (SAMP.LE.0) GO TO 100
SET CLOCK
T=SECOND (T)
EVALUATE SENSOR TERMS
DO 10 I=1, 32
10 SN1 (I)=EXP (Z1*S1 (I)+Z2*S2 (I))
FORM THE INTERPOLATED JN AND PUT IN JN1
J1=0
J2=0
DO 30 I=1, 128
TEMP=DELJ (I)
DO 20 J=1, 32
K1=J1+J
KL=JNS (K1)
KH=JNS (K1+1)
TEMP1=JN (KL)
K2=J2+I
20 JN1 (K2)=TEMP1+TEMP*(JN (KH)-TEMP1)
J1=J1+32
30 J2=J2+32
EXPAND INTERPOLATED MATRIX ON BOTH SIDES
J1=NJ1
J2=NJ2
K1=NK1
K2=NK2
DO 40 I=1, NTERM32
JNA (J1)=JNA (J2)
JNA (K1)=JNA (K2)
J1=J1+1
J2=J2+1
K1=K1+1
40 K2=K2+1
CONVOLUTION
DO 50 I=1, 4096
J=I+NSIZE32
50 JN (I)=JNA (J)

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```

J1=NSIZE32
J2=J1
DO 60 I=1,NTERM
J1=J1+32
J2=J2+32
TEMP=A(I)
DO 60 J=1,4096
K1=J1+J
K2=J2+J
60 JN(J)=JN(J)+TEMP*(JNA(K1)+JNA(K2))
C CUMULATE ROW SUMS
DO 80 I=1,32
I1=I
TEMP2=JN(I1)
DO 70 J=1,127
I1=I1+32
70 TEMP2=TEMP2+JN(I1)
80 TROW(I)=TEMP2
C ACCUMULATE ESTIMATES AND NORMALIZATION CONSTANT
CNORM=TROW(I)*SN1(I)
SHAT=SINY(I)*CNORM
CHAT=COSY(I)*CNORM
DO 85 I=2,32
TEMP2=TROW(I)*SN1(I)
SHAT=SHAT+SINY(I)*TEMP2
CHAT=CHAT+COSY(I)*TEMP2
85 CNORM=CNORM+TEMP2
CNORM=1.0/CNORM
SHAT=SHAT*CNORM
CHAT=CHAT*CNORM
C TRANSFER NORMALIZED DENSITY
DO 90 I=1,32
I1=I
TEMP2=SN1(I)*CNORM
DO 90 J=1,127
JN(I1)=TEMP2*JN(I1)
90 I1=I1+32
C TIMEOUT
TNLF=SECOND(TT)-T
RETURN
C INITIALIZE SAMPLE PATH BY TRANSFERING J0 TO JN
100 IF (MC.LE.0) GO TO 200
DO 110 I=1,4096
110 JN(I)=J0(I)
RETURN
C GLOBAL INITIALIZATIONS FOR NONLINEAR FILTER
200 NSIZE=10
NTERM=64.0*SQRT(50.*Q22)/PIDEL+0.5
IF (NTERM.GT.NSIZE) NTERM=NSIZE
NSIZE32=NSIZE*32
NTERM32=NTERM*32
NK2=NSIZE32+1
NJ1=NK2-NTERM32
NJ2=NSIZE32+4097-NTERM32
NK1=NSIZE32+4097

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```

PHASE VARIABLES
DO 210 I=1,32
SIGMA (I)=PI*((2.*I-1.)/32.-1.)
COSY (I)=COS (SIGMA (I))
SINY (I)=SIN (SIGMA (I))
S1 (I)=COSY (I)/RDEL
210 S2 (I)=SINY (I)/RDEL
PHASE RATE VARIABLES
DO 220 I=1,128
220 PSI (I)=PIDEL*((2.*I-1.)/128.-1.)
SETUP THE TRANSFER MATRIX
DO 240 J=1,128
J1=(J-1)*32
J2=(J-1)*33
DO 230 I=1,32
I1=J1+1+MOD (46-(J-1)/4+I,32)
I2=J2+I
230 JNS (I2)=I1
240 JNS (J2+33)=JNS (J2+1)
SETUP THE INTERPOLATION VECTOR
IN (1)=0.875
IN (2)=0.625
IN (3)=0.375
IN (4)=0.125
IN (5)=IN (1)
IN (6)=IN (2)
IN (7)=IN (3)
IN (8)=IN (4)
J=MOD (NTERM,4)
DO 245 I=1,4
245 DELJ (I)=IN (I)
DO 250 I=5,125,4
DELJ (I)=DELJ (I-4)
DELJ (I+1)=DELJ (I-3)
DELJ (I+2)=DELJ (I-2)
250 DELJ (I+3)=DELJ (I-1)
EVALUATE CONVOLUTION TERMS A (I)
DO 280 I=1,NTERM
TEMP=I/128.
TEMP=CONST*TEMP*TEMP
A (I)=0.
IF (TEMP.GT.-47) A (I)=EXP (TEMP)
280 CONTINUE
CONSTRUCT THE A PRIORI DENSITY
CNORM=1.0/(TWOPI*SQRT (A11*A22))
CL=-0.5/A22
SI=-0.5/A11
DO 290 I=1,32
CB=SIGMA (I)-Y1EST
CS=CB*CB*SI
J1=0
DO 230 J=1,128
J2=J1+I
TEMP=PSI (J)-Y2EST
J0 (J2)=EXP (TEMP*TEMP*CL+CB)*CNORM
290 J1=J1+32
RETURN
END

```

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B. TWO-DIMENSIONAL CDC-STAR-100 PROGRAM

```

PROGRAM MAIN(INPUT,OUTPUT,TAPE5=INPUT,TAPE6=OUTPUT)
C   SAMPLE PATH VARIABLES
REAL SX1(130),SCHAT(130),SSHAT(130),SX1HATNL(130),SERRNL(130),
1  TNLF(130),SPLOSSNL(130),SX1HATPL(130),SERRPL(130),SP11PL(130)
C   CUMULATIVE SAMPLE PATH VARIABLES
REAL CERRNL(130),CESQNL(130),CEVARNL(130),CDBNL(130),CCSNL(130)
REAL CERRPL(130),CESQPL(130),CEVARPL(130),CDBPL(130),CCSPL(130)
C   MONTE CARLO SUMMARY STATISTICS
REAL XERRNL,XESQNL,XEVARNL,XDBNL,XCSNL,
1  XERRPL,XESQPL,XEVARPL,XDBPL,XCSPL
C   SINGLE SAMPLE VARIABLES
REAL CHAT,SHAT,X1HAT,P11,X1,Z1,Z2
C   CONSTANTS
REAL CS(130),DBEPS,EPS(130),TWOPI,ZERO(130),ONE(130),PI2(130)
C   WORKING VARIABLES
INTEGER I,J,K,L
REAL T,TEMP(130)
LOGICAL PATH,CUMPATH
BIT BT(130)
C   PROBLEM SETUP VARIABLES
REAL ALP110,DELF,Q22C,Y1EST,Y2EST
INTEGER NMC,NSAMP,MD,ND
C   DERIVED PROBLEM CONSTANTS
REAL A11,A22,CONST,DEL,FTC,PI,PIDEL,P110,RDEL,RX,QQ,Q22
C   PROBLEM COMMON
COMMON /PROB/ TWOPI,PI,ALP110,DELF,Q22C,Y1EST,Y2EST,A11,A22,
1  CONST,DEL,FTC,PIDEL,P110,RDEL,RX,QQ,Q22,MD,ND
C   SET PRINTOUT CONTROL
PATH=.TRUE.
CUMPATH=.TRUE.
C   READ INPUT PARAMETERS
10 READ (5,5000,END=500) Y1EST,Y2EST,ALP110,DELF,Q22C,NMC,NSAMP,MD
5000 FORMAT(5E10.4,3I5)
C   COMPUTE THE CONSTANTS
MD=(MD/2)*2
IF (MD.LT.20) MD=32
IF (MD.GT.64) MD=32
ND=4*MD
PI=4.0*ATAN(1.0)
TWOPI=2.0*PI
P110=10.**(ALP110/10.)
QQ=Q22C**(.25)
RX=(P110/(SQRT(2.0)*QQ))**(4.0/3.0)
FTC=SQRT(2.0)*RX**(.25)/QQ
DEL=DELF*FTC
Q22=Q22C*DEL
PIDEL=PI/DEL
RDEL=RX/DEL
A11=10.0**((ALP110+1.4)/10.)
A22=2.0*P110/(FTC*FTC)
CONST=-2.0*PIDEL*PIDEL/Q22
CS(1:130)=0.75*PI
PI2(1:130)=TWOPI
ZERO(1:130)=0.
EPS(1:130)=1.E-50

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DBEPS=ALOG10(EPS(1))
ONE(1;130)=1.
C INITIALIZE CUMULATIVE SAMPLE PATH VARIABLES
CERRNL(1;130)=0.
CESQNL(1;130)=0.
CCSNL(1;130)=0.
CERRPL(1;130)=0.
CESQPL(1;130)=0.
CCSPL(1;130)=0.
C PRINTOUT PROBLEM PARAMETERS
WRITE (6,6000) NMC,NSAMP,ALP110,DELF,Q22C,Y1EST,Y2EST,
1 P110,QQ,RX,FTC,DEL,Q22,PIDEL,RDEL,A11,A22,CONST,CS(1)
6000 FORMAT(1H1,31X,18HPROBLEM PARAMETERS/
1 1H0,11X,9HPARAMETER,6X,5HVALUE,11X,9HPARAMETER,
2 6X,5HVALUE/1H0,14X,3HNMC,9X,I4,14X,5HNSAMP,8X,I4/
3 13X,6HALP110,3X,E15.8,8X,4HDELF,4X,E15.8/
4 14X,4HQ22C,4X,E15.8,8X,5HY1EST,3X,E15.8/
5 14X,5HY2EST,3X,E15.8,8X,4HP110,4X,F15.8/
6 15X,2HQQ,5X,E15.8,9X,2HRX,5X,E15.8/
7 15X,3HFTC,4X,E15.8,9X,3HDEL,4X,E15.8/
8 15X,3HQ22,4X,E15.8,8X,5HPIDEL,3X,F15.8/
9 14X,4HRDEL,4X,E15.8,9X,3HA11,4X,E15.8/
A 15X,3HA22,4X,E15.8,8X,5HCONST,3X,F15.8/
B 15X,2HCS,5X,E15.8)
C BEGIN THE MONTE CARLO PROCESS
C INITIALIZATION OF THE SUBROUTINES
CALL STATE(0,0,X1,Z1,Z2)
CALL NLF(0,0,0.,0.,SHAT,CHAT,T)
C MONTE CARLO LOOP
DO 200 K=1,NMC
C INITIALIZATION OF SAMPLE PATH VARIABLES
CALL STATE(K,0,X1,Z1,Z2)
CALL NLF(K,0,Z1,Z2,SHAT,CHAT,T)
CALL PLL(K,0,Z1,Z2,X1HAT,P11)
C SAMPLE PATH LOOP
DO 100 J=1,NSAMP
CALL STATE(K,J,X1,Z1,Z2)
CALL NLF(K,J,Z1,Z2,SHAT,CHAT,T)
CALL PLL(K,J,Z1,Z2,X1HAT,P11)
C STORE THE SAMPLE VARIABLES
SX1(J)=X1
SSHAT(J)=SHAT
SCHAT(J)=CHAT
TNLF(J)=T
SX1HATPL(J)=X1HAT
100 SP11PL(J)=P11
C VECTOR ACCUMULATE THE SAMPLE PATH AVERAGES
SX1HATNL(1;130)=VATAN2(SSHAT(1;130),SCHAT(1;130);
1 SX1HATNL(1;130))
SERRNL(1;130)=SX1(1;130)-SX1HATNL(1;130)
CALL MOD2PI(SERRNL)
SPLOSSNL(1;130)=SSHAT(1;130)*SSHAT(1;130)+SCHAT(1;130)*
1 SCHAT(1;130)
SPLOSSNL(1;130)=2.0*(1.0-VSQRT(SPLOSSNL(1;130);SPLOSSNL(1;130)))
CERRNL(1;130)=VAVG(CERRNL(1;130),SERRNL(1;130),K;CERRNL(1;130))
TEMP(1;130)=SERRNL(1;130)*SERRNL(1;130)

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CESQNL(1;130)=VAVG(CESQNL(1;130),TEMP(1;130),K;CESQNL(1;130))
IF (K.LE.1) GO TO 110
CEVARNL(1;130)=CESQNL(1;130)-CERRNL(1;130)*CERRNL(1;130)
BT(1;130)=(CEVARNL(1;130).LE.EPS(1;130))
CFVARNL(1;130)=Q8VMASK(EPS(1;130),CEVARNL(1;130),BT(1;130):
1  CEVARNL(1;130))
CDBNL(1;130)=VALOG10(CEVARNL(1;130);CDBNL(1;130))
CDBNL(1;130)=10.0*CDBNL(1;130)
110 BT(1;130)=(SERRNL(1;130).GT.CS(1;130))
TEMP(1;130)=Q8VMASK(ONE(1;130),ZERO(1;130),BT(1;130):
1  TEMP(1;130))
CCSNL(1;130)=VAVG(CCSNL(1;130),TEMP(1;130),K;CCSNL(1;130))
SERRPL(1;130)=SX1(1;130)-SX1HATPL(1;130)
CALL MOD2PI(SERRPL)
CERRPL(1;130)=VAVG(CERRPL(1;130),SERRPL(1;130),K;CERRPL(1;130))
TEMP(1;130)=SERRPL(1;130)*SERRPL(1;130)
CESQPL(1;130)=VAVG(CESQPL(1;130),TEMP(1;130),K;CESQPL(1;130))
IF (K.LE.1) GO TO 120
CEVARPL(1;130)=CESQPL(1;130)-CERRPL(1;130)*CERRPL(1;130)
BT(1;130)=(CEVARPL(1;130).LE.FPS(1;130))
CEVARPL(1;130)=Q8VMASK(EPS(1;130),CEVARPL(1;130),BT(1;130):
1  CEVARPL(1;130))
CDBPL(1;130)=VALOG10(CEVARPL(1;130);CDBPL(1;130))
CDBPL(1;130)=10.0*CDBPL(1;130)
120 BT(1;130)=(SERRPL(1;130).GT.CS(1;130))
TEMP(1;130)=Q8VMASK(ONE(1;130),ZERO(1;130),BT(1;130):
1  TEMP(1;130))
CCSPL(1;130)=VAVG(CCSPL(1;130),TEMP(1;130),K;CCSPL(1;130))
IF (PATH.AND.(K.EQ.1 )) GO TO 130
GO TO 200
C PRINT SAMPLE PATH VARIABLES
130 WRITE (6,6010)
6010 FORMAT(1H1,36X,21HSAMPLE PATH VARIABLES/
1 38X,19H(FIRST SAMPLE PATH)/)
WRITE (6,6011)
6011 FORMAT(/18X,9HSX1- ,6X,
1 36H PHASE VARIABLES /
2 20X,1H1,13X,11HSX1(I) ,8X,13HSX1(I+1) /)
WRITE(6,6100)((I,SX1(I),SX1(I+1)),I=1,129,2)
WRITE (6,6012)
6012 FORMAT(/18X,9HSSHAT- ,6X,
1 36H SIN(SX1) ESTIMATES /
2 20X,1H1,13X,11HSSHAT(I) ,8X,13HSSHAT(I+1) /)
WRITE(6,6100)((I,SSHAT(I),SSHAT(I+1)),I=1,129,2)
WRITE (6,6013)
6013 FORMAT(/18X,9HSCHAT- ,6X,
1 36H COS(SX1) ESTIMATES /
2 20X,1H1,13X,11HSCHAT(I) ,8X,13HSCHAT(I+1) /)
WRITE(6,6100)((I,SCHAT(I),SCHAT(I+1)),I=1,129,2)
WRITE (6,6014)
6014 FORMAT(/18X,9HSX1HATNL- ,6X,
1 36H NONLINEAR ESTIMATES /
2 20X,1H1,13X,11HSX1HATNL(I) ,8X,13HSX1HATNL(I+1) /)
WRITE(6,6100)((I,SX1HATNL(I),SX1HATNL(I+1)),I=1,129,2)
WRITE (6,6015)
6015 FORMAT(/18X,9HSERRNL- ,6X,
1 36H NONLINEAR ERRORS /)

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2 20X,1HI,13X,11HSERRNL(I) ,8X,13HSERRNL(I+1) /)
WRITE(6,6100)((I,SERRNL(I),SERRNL(I+1)),I=1,129,2)
WRITE(6,6016)
6016 FORMAT(/18X,9HSPLOSSNL- ,6X,
1 36H NONLINEAR PLOSS FUNCTION /
2 20X,1HI,13X,11HSPLLOSSNL(I) ,8X,13HSPLLOSSNL(I+1) /)
WRITE(6,6100)((I,SPLLOSSNL(I),SPLLOSSNL(I+1)),I=1,129,2)
WRITE(6,6017)
6017 FORMAT(/18X,9HTNLF- ,6X,
1 36H NONLINEAR EXECUTION TIMES /
2 20X,1HI,13X,11HTNLF(I) ,8X,13HTNLF(I+1) /)
WRITE(6,6100)((I,TNLF(I),TNLF(I+1)),I=1,129,2)
WRITE(6,6018)
6018 FORMAT(/18X,9HSX1HATPL- ,6X,
1 36H PHASE LOCK LOOP ESTIMATES /
2 20X,1HI,13X,11HSX1HATPL(I) ,8X,13HSX1HATPL(I+1) /)
WRITE(6,6100)((I, SX1HATPL(I),SX1HATPL(I+1)),I=1,129,2)
WRITE(6,6019)
6019 FORMAT(/18X,9HSERRPL- ,6X,
1 36H PHASE LOCK LOOP ERRORS /
2 20X,1HI,13X,11HSERRPL(I) ,8X,13HSERRPL(I+1) /)
WRITE(6,6100)((I,SERRPL(I),SERRPL(I+1)),I=1,129,2)
WRITE(6,6020)
6020 FORMAT(/18X,9HSP11PL- ,6X,
1 36HP11 FROM PHASE LOCK RICATTI EQUATION /
2 20X,1HI,13X,11HSP11PL(I) ,8X,13HSP11PL(I+1) /)
WRITE(6,6100)((I,SP11PL(I),SP11PL(I+1)),I=1,129,2)
200 CONTINUE
C PRINT CUMULATIVE SAMPLE PATH VARIABLES
IF (CUMPATH.AND.(NMC.GT.1)) GO TO 310
GO TO 400
310 WRITE(6,6030)
6030 FORMAT(1HI,30X,32HCUMULATIVE SAMPLE PATH VARIABLES /)
WRITE(6,6031)
6031 FORMAT(/18X,9HCERRNL- ,6X,
1 36H CUMULATIVE NONLINEAR ERRORS /
2 20X,1HI,13X,11HCERRNL(I) ,8X,13HCERRNL(I+1) /)
WRITE(6,6100)((I,CERRNL(I),CERRNL(I+1)),I=1,129,2)
WRITE(6,6032)
6032 FORMAT(/18X,9HCERRPL- ,6X,
1 36HCUMULATIVE PHASE LOCK ERRORS /
2 20X,1HI,13X,11HCERRPL(I) ,8X,13HCERRPL(I+1) /)
WRITE(6,6100)((I,CERRPL(I),CERRPL(I+1)),I=1,129,2)
WRITE(6,6033)
6033 FORMAT(/18X,9HCESQNL- ,6X,
1 36HCUMULATIVE NONLINEAR SQUARED ERRORS /
2 20X,1HI,13X,11HCESQNL(I) ,8X,13HCESQNL(I+1) /)
WRITE(6,6100)((I,CESQNL(I),CESQNL(I+1)),I=1,129,2)
WRITE(6,6034)
6034 FORMAT(/18X,9HCESQPL- ,6X,
1 36HCUMULATIVE PHASE LOCK SQUARED ERRORS /
2 20X,1HI,13X,11HCESQPL(I) ,8X,13HCESQPL(I+1) /)
WRITE(6,6100)((I,CESQPL(I),CESQPL(I+1)),I=1,129,2)
WRITE(6,6035)

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```

6035 FORMAT(/18X,9HCEVARNL-      ,6X,
1  36HCUMULATIVE NONLINEAR ERROR VARIANCE /
2  20X,1HI,13X,11HCEVARNL(I) ,8X,13HCEVARNL(I+1) /)
WRITE (6,6100)((I,CEVARNL(I),CEVARNL(I+1)),I=1,129,2)
WRITE (6,6036)
6036 FORMAT(/18X,9HCEVARPL-      ,6X,
1  36HCUMULATIVE PHASE LOCK ERROR VARIANCE /
2  20X,1HI,13X,11HCEVARPL(I) ,8X,13HCEVARPL(I+1) /)
WRITE (6,6100)((I,CEVARPL(I),CEVARPL(I+1)),I=1,129,2)
WRITE (6,6037)
6037 FORMAT(/18X,9HCDBNL-      ,6X,
1  36HCUMULATIVE NONLINEAR ERROR DECIBELS /
2  20X,1HI,13X,11HCDBNL(I) ,8X,13HCDBNL(I+1) /)
WRITE (6,6100)((I,CDBNL(I),CDBNL(I+1)),I=1,129,2)
WRITE (6,6038)
6038 FORMAT(/18X,9HCDBPL-      ,6X,
1  36HCUMULATIVE PHASE LOCK ERROR DECIBELS /
2  20X,1HI,13X,11HCDBPL(I) ,8X,13HCDBPL(I+1) /)
WRITE (6,6100)((I,CDBPL(I),CDBPL(I+1)),I=1,129,2)
WRITE (6,6039)
6039 FORMAT(/18X,9HCCSNL-      ,6X,
1  36HCUMULATIVE NONLINEAR CYCLE SLIPS /
2  20X,1HI,13X,11HCCSNL(I) ,8X,13HCCSNL(I+1) / /)
WRITE (6,6100)((I,CCSNL(I),CCSNL(I+1)),I=1,129,2)
WRITE (6,6040)
6040 FORMAT(/18X,9HCCSPL-      ,6X,
1  36HCUMULATIVE PHASE LOCK CYCLE SLIPS /
2  20X,1HI,13X,11HCCSPL(I) ,8X,13HCCSPL(I+1) /)
WRITE (6,6100)((I,CCSPL(I),CCSPL(I+1)),I=1,129,2)
C  COMPUTE THE MONTE CARLO AVERGES
400 IF (NSAMP.LE.30) GO TO 10
I=NSAMP-30
T=I
XERRNL=Q8SSUM(CERRNL(31:I))/T
XESQNL=Q8SSUM(CESQNL(31:I))/T
XCSNL=Q8SSUM(CCSNL(31:I))/T
XERRPL=Q8SSUM(CERRPL(31:I))/T
XESQPL=Q8SSUM(CESQPL(31:I))/T
XCSPL=Q8SSUM(CCSPL(31:I))/T
XEVARNL=XESQNL-XERRNL*XERRNL
XDBNL=DBEPS
IF (XEVARNL.GT.EPS(1)) XDBNL=10.0*ALOG10(XEVARNL)
XEVARPL=XESQPL-XERRPL*XERRPL
XDBPL=DBEPS
IF (XEVARPL.GT.EPS(1)) XDBPL=10.0*ALOG10(XEVARPL)
C  PRINT THE MONTE CARLO AVERAGES
WRITE (6,6050) NMC,XERRNL,XERRPL,XESQNL,XESQPL,XEVARNL,
1  XEVARPL,XDBNL,XDBPL,XCSNL,XCSPL
6050 FORMAT(1H1,25X,30HMONTE CARLO SUMMARY STATISTICS//
1  32X,1H(,14,14H SAMPLE PATHS)//
2  36X,16HNONLINEAR FILTER,5X,15HPHASE LOCK LOOP//
3  14X,14H***VARIABLE***,7X,12H***VALUES***,9X,
4  12H***VALUES***//14X,13HAVERAGE ERROR,9X,
5  E15.8,6X,E15.8/
6  10X,21HAVERAGE SQUARED ERROR,5X,E15.8,6X,E15.8/
7  14X,14HERROR VARIANCE,8X,E15.8,6X,E15.8 /
8  14X,14HVARIANCE IN DB,8X,E15.8,6X,E15.8/
9  11X,19HAVERAGE CYCLE SLIPS,6X,E15.8,6X,E15.8)

```

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```

6100 FORMAT(20X,I4,8X,E15.8,4X,E15.8)
GO TO 10
500 STOP
END
SUBROUTINE VPROP(A,I)
REAL A(16640)
INTEGER I
INTEGER MD1,MD2,MD3,MD4,MD5,MD6,MD7,
1 MD8,MD9,MD10,MD11,MD12,MD13,MD14
COMMON /CPROP/ MD1,MD2,MD3,MD4,MD5,MD6,MD7,
1 MD8,MD9,MD10,MD11,MD12,MD13,MD14
IF (I.GT.0) GO TO 10
A(MD2;MD1)=A(1;MD1)
A(MD4;MD3)=A(1;MD3)
10 A(MD6;MD5)=A(1;MD5)
A(MD8;MD7)=A(1;MD7)
A(MD10;MD9)=A(1;MD9)
A(MD12;MD11)=A(1;MD11)
A(MD14;MD13)=A(1;MD13)
RETURN
END
SUBROUTINE MOD2PI(A)
REAL A(130),X(130),Y(130)
BIT BT(130)
COMMON /PROB/ TWOPI,PI
X(1:130)=PI
10 BT(1:130)=(A(1:130).GT.X(1:130))
IF (QBSCNT(BT(1:130)).EQ.0) GO TO 20
Y(1:130)=A(1:130)-TWOPI
A(1:130)=QBVMASK(Y(1:130),A(1:130),BT(1:130):A(1:130))
GO TO 10
20 X(1:130)=-PI
30 BT(1:130)=(A(1:130).LT.X(1:130))
IF (QBSCNT(BT(1:130)).EQ.0) RETURN
Y(1:130)=A(1:130)+TWOPI
A(1:130)=QBVMASK(Y(1:130),A(1:130),BT(1:130):A(1:130))
GO TO 30
END
REAL FUNCTION VAVG(AV,X,I;*)
DESCRIPTOR AV,X,VAVG
INTEGER I
REAL XI
XI=I
VAVG =((XI-1.)*AV+X)/XI
RETURN
END
REAL FUNCTION RNF(INIT)
INTEGER INIT,K,KD,M1,M2,N1,N2,NT,MP
REAL T1,T2
COMMON /RNUM/ K,KD,M1,M2,N1,N2,NT,MP,T1,T2
IF (INIT.EQ.0) GO TO 10
N1=102943
N2=185617
N1=244734
N2=158551
M1=244734

```

```

M2=158551
T1=2.**(-18)
T2=2.**(-36)
MP=2**18
10 K=M2*N2
KD=K/MP
NT=K-KD*MP
K=M1*N2+M2*N1+KD
KD=K/MP
N1=K-KD*MP
N2=NT
RNF=N1*T1+N2*T2
RETURN
END
REAL FUNCTION GAUSS(INIT,SD,XM)
INTEGER INIT,I,J
REAL SD,XM,TWOPI,X,XR1,XR2
COMMON /GNUM/ I,J,XR2
COMMON /PROB/ TWOPI
IF (INIT.EQ.0) GO TO 10
I=1
J=1
10 IF (J.NE.1) GO TO 20
J=2
XR1=RNF(I)
I=0
XR2=RNF(0)
X=SQRT(ABS(-2.*ALOG(XR1)))
XR2=TWOPI*XR2
XR1=X*SIN(XR2)
XR2=X*COS(XR2)
GAUSS=XR1*SD+XM
RETURN
20 J=1
GAUSS=XR2*SD+XM
RETURN
END
SUBROUTINE STATE(MC,SAMP,X1,Z1,Z2)
INTEGER MC,SAMP,INIT
REAL X1,Z1,Z2,DEV1,DEV2,DEV3,DEV4,X2
COMMON /STA/ DEV1,DEV2,DEV3,DEV4,INIT
COMMON /PROB/ TWOPI,PI,ALP110,DELTA,Q22C,Y1EST,Y2EST,A11,A22,
1 CONST,DEL,FTC,PIDEL,P110,RDEL,RX,QQ,Q22,MD,ND
IF (MC.NE.0) GO TO 10
DEV1=SQRT(A11)
DEV2=SQRT(A22)
DEV3=SQRT(RDEL)
DEV4=SQRT(Q22)
INIT=1
RETURN
10 IF (SAMP.GT.0) GO TO 20
X1=GAUSS(INIT,DEV1,Y1EST)
INIT=0
X2=GAUSS(0,DEV2,Y2EST)
RETURN

```


C CONVOLUTION LOOP
C

J1=NC
J2=NC
DO 10 I=1, NTERM
J1=J1+MD1
J2=J2-MD1
ASSIGN DJNAJ1, JNA(J1;MND)
ASSIGN DJNAJ2, JNA(J2;MND)
DJN=DJNAJ1+DJNAJ2
DJN=A(I)*DJN
10 DJN1=DJN1+DJN

C MULTIPLY BY SENSOR TERMS
C

DJN1=DSN1A*DJN1

C GET NORMALIZATION CONSTANT
C

CNORM=Q8SSUM(DJN1)
CNORM=1.0/CNORM

C TRANSFER THE NORMALIZED DENSITY
C

DJN=CNORM*DJN1

C CUMULATE ESTIMATES
C

DJNA1=DSINY*DJN
SHAT=Q8SSUM(DJNA1)
DJNA1=DCOSY*DJN
CHAT=Q8SSUM(DJNA1)

C TIMEOUT
C

CALL Q3CLOCKS(TNLF, TT)
RETURN

C
C-----
C
C-----

C SAMPLE PATH INITIALIZATION TAKES PLACE HERE
C

100 IF (MC.LE.0) GO TO 200

C TRANSFER THE INITIAL DENSITY FOR NEW SAMPLE PATH
C

JN(1;MND)=JO(1;MND)
RETURN

C
C-----

C
C
C
C
C
C
C
C
C

GLOBAL INITIALIZATIONS OCCUR HERE FOR THE ENTIRE RUN

DETERMINE THE NUMBER OF CONVOLUTION POINTS

```
200 NTERM=(ND/2.)*SQRT(50.*Q22)/PIDEL+0.5
    MD1=MD+1
    MD2=MD1+1
    MD3=MD1*2
    MD4=MD3+1
    MD5=MD3*2
    MD6=MD5+1
    MD7=MD5*2
    MD8=MD7+1
    MD9=MD7*2
    MD10=MD9+1
    MD11=MD9*2
    MD12=MD11+1
    MD13=MD11*2
    MD14=MD13+1
    MND=MD1*ND
    MND1=MND+1
    NTERM33=MD1*NTERM
    NT=2*NTERM33
    NC=NTERM33+1
```

C
C
C

SET UP THE VECTOR DESCRIPTORS FOR THE UPDATE FUNCTIONS

```
ASSIGN DS1,S1(1:MD)
ASSIGN DS2,S2(1:MD)
ASSIGN DSN1,SN1(1:MD)
ASSIGN DSM2,JNA(1:MD)
ASSIGN DJN,JN(1:MND)
ASSIGN DJNS,JNS(1:MND)
ASSIGN DJNA1,JNA(1:MND)
ASSIGN DJNA2,JNA(2:MND)
ASSIGN DDELJ,DELJ(1:MND)
ASSIGN DJN1,JN1(1:MND)
ASSIGN DJNAENT,JNA(MND1:NT)
ASSIGN DJNABNT,JNA(1:NT)
ASSIGN DJNANC,JNA(NC:MND)
ASSIGN DSN1A,SN1(1:MND)
ASSIGN DSINY,SINY(1:MND)
ASSIGN DCOSY,COSY(1:MND)
```

C
C
C

PHASE VARIABLES

```
DO 210 I=1,MD
SIGMA(I)=PI*((2.*I-1.)/MD -1.)
COSY(I)=COS(SIGMA(I))
SINY(I)=SIN(SIGMA(I))
S1(I)=COSY(I)/RDEL
```

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```
210 S2(I)=SINY(I)/RDEL
    SINY(MD1)=0.
    COSY(MD1)=0.
    CALL VPROP(SINY,0)
    CALL VPROP(COSY,0)
C
C   PHASE RATE VARIABLES
C
    DO 220 I=1,ND
220 PSI(I)=PIDEL*((2.*I-1.)/ND - 1.)
C
C   SETUP THE TRANSFER MATRIX
C
    DO 240 J=1,ND
    J1=MOD(ND-1-NTERM+J,ND)*MD1
    J2=(J-1)*MD1
    DO 230 I=1,MD
    I1=J1+1+MOD(MD+MD/2+32-(135-NTERM+J)/4+I,MD)
    I2=J2+I
230 JNS(I2)=I1
    J1=J2+MD1
240 JNS(J1)=JNS(J2+1)
C
C   SETUP INTERPOLATION MATRIX
C
    IN(1)=0.875
    IN(2)=0.625
    IN(3)=0.375
    IN(4)=0.125
    IN(5)=IN(1)
    IN(6)=IN(2)
    IN(7)=IN(3)
    IN(8)=IN(4)
    J=MOD(NTERM,4)
    DO 245 I=1,4
    I1=(I-1)*MD1+1
    J1=I+4-J
    T=IN(J1)
245 DELJ(I1:MD1)=T
    CALL VPROP(DELJ,1)
C
C   EVALUATE CONVOLUTION TERMS A(I)
C
    DO 280 I=1,NTERM
    T=I
    TT=ND
    TEMP=T/TT
    TEMP=CONST*TEMP*TEMP
    A(I)=0.
    IF (TEMP.GT.-47) A(I)=EXP(TEMP)
280 CONTINUE
C
C   CONSTRUCT THE A PRIORI DENSITY
C
```

```

CNORM=1.0/(TWOPI*SQRT(A11*A22))
CL=-0.5/A22
SI=-0.5/A11
DO 290 I=1,MD
  I1=I
  CR=SIGMA(I)-Y1EST
  CR=CR*CR*SI
  DO 290 J=1,ND
    TEMP=PSI(J)-Y2EST
    JO(I1 )=EXP(TEMP*TEMP*CL+CR)*CNORM
290 I1=I1+MD1
C
C   WRITE OUT PARAMS OF NLF
C
  WRITE (6,6000) MD,ND,MD1,ND
6000 FORMAT(1H0,26X,27HPPOINT MASS NONLINEAR FILTER//1H ,
1  28X,24HVERSION 2, CODED 6/27/76//1H ,
2  18X,I3,1HX,I3,25H DENSITIES REPRESENTED BY ,I3,1HX,I3)
  WRITE (6,6001) NTERM,(A(I),I=1,NTERM)
6001 FORMAT(1H ,33X,7HA(1)-A(,I2,2H) /(1X,5E15.8))
  RETURN
C
C-----
C
  END

0.0      0.0      -3.01      0.1      0.1      200  130  64
0.0      0.0      -1.79      0.1      0.01     200  130  64
END OF FILE

```

C. THREE-DIMENSIONAL CDC-STAR-100 PROGRAM

```

PROGRAM MAIN(INPUT,OUTPUT,PUNCH,TAPE6=OUTPUT,TAPE5=INPUT)
  DIMENSION XDAT(130,10),XHAT(4),Y1(35),Y2(135),
*EXPON(35),EXDON(17),Y3(17),PNF(3,3) ,YA(35),EXP33(16,16),
*Q(3,3),PBAR(3,3),PN(3,3),AN(3),F(3,3),PDUMMY(3,3),PDUMY2(3,3)
  COMMON/GN/JGAUSS,XZZZ(2)
  REAL COSY(24576),SINY(24576),SN1(24576),SN2(1536),JN(24576),
*JN1(26112),JNA(54272),DELJ(26112),S1(16),S2(16),YB(24576),D(2000)
  INTEGER JNS(1536),JNF(3072)
  BIT B3(26112)
  DESCRIPTOR DB,KJNF,DJNA,KJNS,DJN1,DC
  ASSIGN DB,JN(1;16)
  ASSIGN KJNF,JNF(1;3072)
  ASSIGN DJNA,JNA(1;49152)
  ASSIGN KJNS,JNS(1;1536)
  ASSIGN DJN1,JN1(1;26112)
  ASSIGN DC,JNA(1;17)
  JGAUSS=0
  READ(5,63)Y3EST,Q33C,ALF,GAM,NUM3
63  FORMAT(4F10.5,I5)
  READ(5,64)Y1EST,Y2EST,ALP110,DELF,Q22C,NUM1,NUM2,N02,N03
64  FORMAT(5F10.5,4I5)
  P110=10.**((ALP110/10.))
  QQ=Q22C**(.25)
  RX=(P110/(SQRT(2.0)*QQ))**((4.0/3.0))
  FTC=SQRT(2.0)*RX**(.25)/QQ
  DELT=DELF*FTC
  Q22=Q22C*DELT
  R11=RX/DELT
  P220=P110*SQRT(Q22C/RX)
  ALFD=ALF*DELT
  BET=1.0-ALFD
  A11=10.**((ALP110+GAM)/10.)
  A22=P220
  P330=0.5*Q33C/ALF
  A33=2.0*P330
  Q33=Q33C*DELT
  DEV1=SQRT(A11)
  DEV2=SQRT(A22)
  DEV3=SQRT(R11)
  DEV4=SQRT(A33)
  DEVQ2=SQRT(Q22)
  DEVQ3=SQRT(Q33)
  KOUNT=1
  ISAMP=1
  NSAMP=0
  SUMP1=0.0
  SUMP2=0.0
  SUMP3=0.0

```

```

DELSQ=DELT**2
PI=4.*ATAN(1.)
PI2=2.0*PI
PIDLT=PI/DELT
PINV=1.0/PI
PI2DLT=2.0*PIDLT
U1=NUM1
U2=NUM2
U3=NUM3
IY2=U2/PI2DLT*SQRT(50.0*Q22)+.5
NTERM=IY2
NTERM1=NTERM+1
NTERM16=NTERM*16
NC=NTERM16+1
NT=2*NTERM16
NTA1536=NT+1536
R55=0.5/R11
CL=-0.5/A22
CM=-0.5/A33
SI=-0.5/A11
C *****GRID Y1,Y2 AND YA *****
EDG1=PI/U1
EDG2=PIDLT/U2
DO 40 I=1,NUM1
  X=I-1
  X=X/U1
40  Y1(I)=-PI+X*PI2+EDG1
DO 50 I=1,NUM2
  X=I-1
  X=X/U2
50  Y2(I)=-PIDLT+X*PI2DLT+EDG2
DO 51 I=1,IY2
  X=I
  X=X/U2
51  Y3(I)=X
DO 55 I=1,NUM1
  COSY(I)=COS(Y1(I))
55  SINY(I)=SIN(Y1(I))
S1(1;16)=COSY(1;16)/R11
S2(1;16)=SINY(1;16)/R11
CALL VPROP(SINY,1)
CALL VPROP(COSY,1)
XP=0.5*SQRT(A33)
YY3=(NUM3-1.0)/2.0+1.0
IY3=YY3
DO 60 I=1,NUM3
  XX=I-YY3
60  YA(I)=Y3EST+XP*XX
C ***** DYNAMIC EXPONENTIALS *****
IF(IY2.EQ.0)GO TO 153
DO 150 I=1,IY2
150  EXPON(I)=EXP(-0.5/Q22*(Y3(I)**2))
DO 152 I=1,IY2
  EXPON(I)=EXPON(IY2+1-I)
152  EXPON(IY2+1+I)=EXPON(I)

```

```

153   EXPON(IY2+1)=0.5
      IYY=2*IY2+1
      LTERM2=0
      DO 720 I=1,16
      DO 720 K=1,16
      XNUM=(K-I)*XP+ALFD*(YA(I)-1.)
      XNUM=-0.5/Q33*XNUM**2
      EXP33(I,K)=0.0
      IF(XNUM.LT.-27.)GO TO 720
      EXP33(I,K)=EXP(XNUM)
720  CONTINUE
      LTERM=0
      LTERM16=LTERM*16
      LTERM1=LTERM+1
      LC=LTERM16+1
      NS=NTA1536*LTERM+NC
      I=0
      DO 339 K=1,16
      DO 339 N=1,16
      DO 339 J=NTERM1,IYY
      I=I+1
339  D(I)=EXPON(J)*EXP33(N,K)
C     ***** INITIAL DENSITY *****
      IJK=0
      DO 160 K=1,16
      ZZZ=CM*(YA(K)-Y3EST)**2
      DO 160 J=1,96
      YYY=ZZZ+CL*(Y2(J)-Y2EST)**2
      DO 160 I=1,16
      IJK=IJK+1
      XXX=YYY+SI*(Y1(I)-Y1EST)**2
      IF(XXX.LT.-27.)GO TO 159
      JN(IJK)=EXP(XXX)
      GO TO 160
159  JN(IJK)=0.0
160  CONTINUE
C     ***** INTEGER ARRANGEMENT *****
      DO 225 I=1,26112
225  B3(I)=B'1'
      I=0
      DO 300 K=1,16
      DO 300 J=1,96
      I=I+17
300  B3(I)=B'0'
      DO 320 K=1,3071,2
      JNF(K)=(K-1)*8
320  JNF(K+1)=JNF(K)
      I=0
      J1=0
      DO 332 K=1,16
      DO 332 J=1,96
      I=I+1
      I1=J1+MOD(23-(J-1)/6,16)
      JNS(I)=I1
332  J1=J1+32
      DELJ(1;17)=11./12.
      DELJ(18;17)=0.75

```

```

DELJ(35;17)=7./12.
DELJ(52;17)=5./12.
DELJ(69;17)=0.25
DELJ(86;17)=1./12.
CALL VPROP1(DELJ)
11 CONTINUE
KOUNT=1
CALL GAUSS(JSEED,DEV1,Y1EST,X1)
XDAT(KOUNT,1)=X1
CALL GAUSS(JSEED,DEV2,Y2EST,X2)
CALL GAUSS(JSEED,DEV4,Y3EST,X3)
ACOS=X3*COS(X1)
ASIN=X3*SIN(X1)
CALL GAUSS(JSEED,DEV3,ACOS,Z1)
CALL GAUSS(JSEED,DEV3,ASIN,Z2)
GO TO 470
450 CONTINUE
X1=X1+X2*DELT
XDAT(KOUNT,1)=X1
CALL GAUSS(JSEED,DEVQ2,X2,X2)
X3=X3*BET+ALFD
CALL GAUSS(JSEED,DEVQ3,X3,X3)
ACOS=X3*COS(X1)
ASIN=X3*SIN(X1)
CALL GAUSS(JSEED,DEV3,ACOS,Z1)
CALL GAUSS(JSEED,DEV3,ASIN,Z2)
XP=0.5*SQRT(ALOSS)
DO 600 I=1,NUM3
  XX=I-YY3
  YA(I)=X3EST+XX*XP
600 CONTINUE
DO 730 I=1,16
DO 730 K=1,16
XNUM=(K-I)*XP+ALFD*(YA(I)-1.)
XNUM=-0.5/Q33*XNUM**2
EXP33(I,K)=0.0
IF(XNUM.LT.-27.)GO TO 730
EXP33(I,K)=EXP(XNUM)
730 CONTINUE
I=0
DO 340 K=1,16
DO 340 N=1,16
DO 340 J=NTERM1,IYY
I=I+1
340 D(I)=EXPON(J)*EXP33(N,K)
470 CONTINUE
C ***** SENSOR FUNCTION *****
CALL Q3CLOCKS(T,TT)
S1(1;16)=Z1*S1(1;16)
S2(1;16)=Z2*S2(1;16)
S1(1;16)=S1(1;16)+S2(1;16)
J=1
DO 500 K=1,16
S2(1;16)=S1(1;16)*YA(K)
SN2(1;16)=VEXP(S2(1;16);SN2(1;16))

```

```

CALL VPROP(SN2,0)
SN1(J;1536)=SN2(1;1536)
S2(1;16)=-R55*YA(K)*YA(K)
SN2(1;16)=VEXP(S2(1;16);SN2(1;16))
CALL VPROP(SN2,0)
SN1(J;1536)=SN1(J;1536)*SN2(1;1536)
500 J=J+1536
C ***** MAIN LOOP STARTS *****
CALL Q8VXTDV(X'02',0,KJNF,0,DB,0,DJNA)
CALL Q8VXTDV(X'02',0,KJNS,0,DC,0,DJN1)
JNA(1;26112)=JN1(2;26112)-JN1(1;26112)
JNA(1;26112)=DELJ(1;26112)*JNA(1;26112)
JN1(1;26112)=JN1(1;26112)+JNA(1;26112)
JNA(1;24576)=Q8VCMPRS(JN1(1;26112),B3(1;26112);JNA(1;24576))
JN1(1;24576)=JNA(1;24576)
J=1
I=1
DO 510 K=1,16
N=I+NTERM16
JNA(N;1536)=JN1(J;1536)
JNA(I;NTERM16)=JNA(I+1536;NTERM16)
JNA(N+1536;NTERM16)=JN1(J;NTERM16)
J=J+1536
510 I=I+NTA1536
N=1
I1=0
JK=1
DO 700 I=1,16
JN1(N;1536)=0.0
DO 690 K=1,16
J1=NS+I1
J2=NS+I1
DO 680 J=1,NTERM1
JN(1;1536)=JNA(J1;1536)+JNA(J2;1536)
JN(1;1536)=JN(1;1536)*D(JK)
JN1(N;1536)=JN1(N;1536)+JN(1;1536)
JK=JK+1
J1=J1+16
680 J2=J2-16
690 CONTINUE
N=N+1536
700 I1=I1+NTA1536
JN1(1;24576)=JN1(1;24576)*SN1(1;24576)
CNORM=SUMLOG(JN1)
IF(CNORM.LT.1.0E-20)CNORM=1.0
CNORM=1./CNORM
JN(1;24576)=CNORM*JN1(1;24576)
JNA(1;24576)=COSY(1;2457)*JN(1;24576)
CHAT=SUMLOG(JNA)
JNA(1;24576)=SINY(1;24576)*JN(1;24576)
SHAT=SUMLOG(JNA)
CXHAT=ATAN2(SHAT,CHAT)
J=1
DO 343 K=1,16
YB(J;1536)=YA(K)

```

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343 J=J+1536
      JNA(1;24576)=YB(1;24576)*JN(1;24576)
      X3EST=SUMLOG(JNA)
      JNA(1;24576)=JNA(1;24576)*YB(1;24576)
      ALOSS=SUMLOG(JNA)
      ALOSS=ALOSS-X3EST*X3EST
C ***** MAIN LOOP ENDS *****
      CALL Q3CLOCKS(TNLF,TT)
      WRITE(6,201)KOUNT,X1,X2,X3,Z1,Z2,CXHAT,X3EST,ALOSS
201  FORMAT( 15,1X,1P3E14.6,4X,1P2E14.6,4X,1P3E14.6 )
      WRITE(6,8880)TNLF
8880  FORMAT(1PE12.6)
      IF(KOUNT.EQ.NO2)GO TO 505
      KOUNT=KOUNT+1
      GO TO 450
505  CONTINUE
      SUMP=0.0
      SUMC=0.0
      XDAT(KOUNT,2)=CXHAT
      XDAT(KOUNT,3)=ALOSS
      XDAT(KOUNT,4)=X3EST
      DO 1501 I=31,NO2
          XD=ABS(XDAT(I,1)-XDAT(I,2))
1498  CONTINUE
          IF(XD.GT.PI)GO TO 1499
          GO TO 1500
1499  XD=XD-PI2
          GO TO 1498
1500  SUMP=SUMP+XD**2
1501  CONTINUE
      H=NO2-30
      SUMP=SUMP/H
      XNSAMP=NSAMP
      XAA=XNSAMP+1.0
      SUMP1=(SUMP+XNSAMP*SUMP1)/XAA
      DSUMP1=ALOG10(SUMP1)*10.0
      DO 1601 I=31,NO2
          XD=ABS(XDAT(I,1)-XDAT(I,4))
1698  CONTINUE
          IF(XD.GT.PI)GO TO 1699
          GO TO 1700
1699  XD=XD-PI2
          GO TO 1698
1700  SUMC=SUMC+XD**2
1601  CONTINUE
      SUMC=SUMC/H
      SUMP2=(SUMC+XNSAMP*SUMP2)/XAA
      DSUMP2=ALOG10(SUMP2)*10.0
      WRITE(6,1511)NO3,SUMP1,DSUMP1,SUMP2,DSUMP2
1511  FORMAT(110,1P4E14.6)
      NSAMP=NSAMP+1
      IF(ISAMP.EQ.NO3)GO TO 2200
      ISAMP=ISAMP+1
      GO TO 11

```

2200 CONTINUE

STOP

END

FUNCTION SUMLOG(A)

REAL A(26112),C(12288)

C(1;12288)=A(1;12288)+A(12289;12288)

C(1;6144)=C(1;6144)+C(6145;6144)

C(1;3072)=C(1;3072)+C(3073;3072)

C(1;1536)=C(1;1536)+C(1537;1536)

C(1;768)=C(1;768)+C(769;768)

C(1;384)=C(1;384)+C(385;384)

C(1;192)=C(1;192)+C(193;192)

C(1;96)=C(1;96)+C(97;96)

C(1;48)=C(1;48)+C(49;48)

C(1;24)=C(1;24)+C(25;24)

C(1;12)=C(1;12)+C(13;12)

C(1;6)=C(1;6)+C(7;6)

C(1;3)=C(1;3)+C(4;3)

SUMLOG=C(1)+C(2)+C(3)

RETURN

END

SUBROUTINE VPROP(A,I)

REAL A(24576)

IF(I.EQ.2)GO TO 10

A(17;16)=A(1;16)

A(33;32)=A(1;32)

A(65;32)=A(1;32)

10 A(97;96)=A(1;96)

A(193;192)=A(1;192)

A(385;384)=A(1;384)

A(769;768)=A(1;768)

IF(I.EQ.0)RETURN

A(1537;1536)=A(1;1536)

A(3073;3072)=A(1;3072)

A(6145;6144)=A(1;6144)

A(12289;12288)=A(1;12288)

RETURN

END

SUBROUTINE VPROP1(A)

REAL A(26112)

A(103;102)=A(1;102)

A(205;204)=A(1;204)

A(409;408)=A(1;408)

A(817;816)=A(1;816)

A(1633;1632)=A(1;1632)

A(2265;2264)=A(1;2264)

A(3265;3264)=A(1;3264)

A(6529;6528)=A(1;6528)

A(13057;13056)=A(1;13056)

RETURN

END

FUNCTION RNNF(NS,MODE)

```

DIMENSION NS(2), NC(2)
COMMON /RN/ N1, N2, MP, T1, T2
DATA M1, M2/244734, 158551/
C      MODE=0 TO CONTINUE, OTHERWISE RESTART WITH
C      INTEGER NUMBER NS(1)*2**18+NS(2)
      IF (MODE) 10, 100, 10
10  N1=NS(1)
    N2=NS(2)
    T1=2.**(-18)
    T2=2.**(-36)
    MP=2**18
100 DO 200 I=1,2
    GO TO (110,120),I
110 K=M2*N2
    GO TO 190
120 K=M1*N2+M2*N1+KD
190 KD=K/MP
200 NC(I)=K-KD*MP
    N1=NC(2)
    N2=NC(1)
    XN1=N1
    XN2=N2
    RNNF=XN1*T1+XN2*T2
    RETURN
    END
    SUBROUTINE GAUSS(JS,SD,XM,X)
    DIMENSION NST(2)
    COMMON /RN/ N1, N2, MC, T1, T2
    COMMON /GN/ J,XR(2)
    IF (J) 10, 10, 20
10  J=2
    TWOPI=8.*ATAN(1.)
    NST(1)=244734
    NST(2)=158551
    NST(1)=102943
    NST(2)=185617
    XR(1)=RNNF(NST,1)
    GO TO 35
20  GO TO (30,40), J
30  J=2
    XR(1)=RNNF(NST,0)
35  XR(2)=RNNF(NST,0)
    X1=SQRT(ABS(-2.*ALOG(XR(1))))
    XR(2)=TWOPI*XR(2)
    XR(1)=X1*SIN(XR(2))
    XR(2)=X1*COS(XR(2))
    X=XR(1)*SD+XM
    RETURN
40  J=1
    X=XR(2)*SD+XM
    RETURN
    END

```

```

1.0      0.01      1.0      1.4      16
0.0      0.0      -3.0     0.1     0.01     16     96     130     1
END OF FILE

```