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HELICOPTER REMOTE WIND SENSOR SYSTEM DESCRIPTION.(U)
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HELICOPTER REMOTE WIND SENSOR SYSTEM DESCRIPTION

SEPTEMBER 1979

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20. ABSTRACT (Continue on reverse side if necessary and identify by block number)

The helicopter remote wind sensor (HRWS) is an application of a laser Doppler velocimeter. This system description describes the HRWS fabricated by Raytheon Company for the US Army Atmospheric Sciences Laboratory. The HRWS was designed to measure wind fields in an aircraft turbulent environment. The operational emphasis is to augment an attack helicopter's fire control system. This CO_2 laser heterodyne system

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20. ABSTRACT (CONT)

remotely measures three dimensional winds from 1 to 33 meters in front of the sensor. The HRWS was mounted in an external pod compatible with AH-1, AAH and UH-1 wing stores or a fixed wing aircraft equipped with standard wing shackles. The HRWS theory, functions, and subsystem are discussed. Data from flight and ground testing will be presented in a later report.

CONTENTS

FIGURES

TABLES

INTRODUCTION

This report describes the helicopter remote wind sensor built by Raytheon Company for the US Army Atmospheric Sciences Laboratory. The system was designed to fly on an AH-1 helicopter (attack helicopter) to improve the accuracy of free flight armament, e.g., 2.75-in. rockets, by measuring the atmospheric flow field along the rocket path. Measurements were made of the wind field from ranges of 1 to 32 m in front of the sensor and every 2 degrees around a 17-degree half-angle cone. These data allow the calculation of the three dimensional flow field vector as a function of range. The sensor system is located in a pod that can be suspended below the stub wing of the AH-1 helicopter. An artist's concept is shown in figure 1, and the actual system mounted on a UH-1 helicopter (utility helicopter) (used for test purposes) is shown in figure 2.

This report is a system description only and as such does not present analyses, conclusions, or recommendations typically contained in technical test reports.

Doppler Theory

Operation of a laser Doppler velocimeter may be understood with the aid of figure 3. A laser beam is transmitted through an interferometer which passes most of the energy to the target and diverts a small amount to the detector as shown in figure 3A. The transmitted beam is usually passed through a telescope to the target, which in the present case consists of aerosols naturally suspended in the atmosphere. Some of the transmitted energy is backscattered by the target particles and enters the detector by the path shown in figure 3B. This returning energy has been shifted in frequency by an amount proportional to the component of particle velocity parallel to the direction of propagation $V_{||}$, according to the Doppler principle. Figure 4 shows the geometry of the laser system and target velocity vector relationship. Only $V_{||}$ contributes to the Doppler shift.

Two beams fall on the detector at two different frequencies (figure 5): a reference beam at the optical frequency f_o with a power of a few milliwatts and the smaller power signal beam at the new frequency $f_o + f_d$, where f_d is the Doppler shift frequency. When these beams are superimposed, the combination contains energy at the difference of these frequencies.

The difference is the Doppler frequency and is related to the target velocity. A sample spectrum in figure 6 shows the intensity as a function of frequency for a laser velocimeter system signal. There is a peak at 0.7 MHz corresponding to a velocity of about 3.35 m/s for $f_o = \frac{C}{\lambda} = 2.83 \times 10^{13}$ Hz. C = speed of light; λ = wavelength.

CONICAL SCANNING SYSTEM - THEORETICAL BACKGROUND

The use of a coaxial heterodyne laser system to obtain a component of the velocity vector of the atmosphere has been described in the previous paragraphs. A conical scan can also be performed with the beam from the system to measure all three vector components. This technique is described in this portion of the report.

The system geometry is illustrated in figure 7. The cone has an axis $\hat{r}$ which lies in the x-z plane at an angle θ with the z-axis. The half-apex angle of the cone is $\delta/2$. If at time, $t = 0$, the propagation vector $\vec{K}$ is in the x-z plane, then

$$\vec{K} = |K| \left[\cos \frac{\delta}{2} \hat{r} + \sin \frac{\delta}{2} \cos \omega t \, \hat{p}_x + \sin \frac{\delta}{2} \sin \omega t \, \hat{p}_y\right], \qquad (1)$$

where ω is the scanning frequency. P_x and P_y are as defined in figure 7, with P_y normal to P_x. For an arbitrary velocity vector,

$$\vec{V} = V_x \hat{i} + V_y \hat{j} + V_z \hat{k} \qquad (2)$$

The Doppler frequency shift is proportional to the dot product of $\vec{K}$ and $\vec{V}$,

$$f_D = - \frac{\vec{K} \cdot \vec{V}}{\Pi} . \qquad (3)$$

Substituting equations (1) and (2) into equation (3),

$$f_D = - \frac{|K|}{\Pi} \left[\cos \frac{\delta}{2} (V_x \sin \theta + V_z \cos \theta) + \cos \omega t \sin \frac{\delta}{2} (V_x \cos \theta - V_z \sin \theta) + \sin \omega t \sin \frac{\delta}{2} V_y\right]. \qquad (4)$$

The Doppler shift consists of two parts: a DC term f_{DC} equal to $- |K|/\Pi \cos \delta/2 \, (V_x \sin \theta + V_z \cos \theta)$ and a time varying term f_{AC}

equal to $- |K|/\Pi \cos \omega t \sin \delta/2 (V_x \cos \theta - V_z \sin \theta) + \sin \omega t \sin \delta/2 V_y$. Since the angles θ and δ are known, the three velocity components can be determined in the following fashion. At time, $t = \Pi/2\omega$, the magnitude of the time varying term is proportional to the velocity component in the y-direction. Thus

$$V_y = - \frac{\Pi}{|K| \sin \frac{\delta}{2}} f_{AC} \Bigg|_{t = \frac{\Pi}{2\omega}} . \tag{5}$$

The x and z components can be found by the DC term and the AC term at time, t = 0.

$$f_{DC} (t = 0) = - \frac{|K|}{\Pi} \cos \frac{\delta}{2} (V_x \sin \theta + V_z \cos \theta), \tag{6}$$

$$f_{AC} (t = 0) = - \frac{|K|}{\Pi} \sin \frac{\delta}{2} (V_z \cos \theta - V_z \sin \theta). \tag{7}$$

Solving equations (6) and (7) simultaneously,

$$V_x = \frac{\Pi \left[f_{DC} \sin \theta \sin \frac{\delta}{2} - f_{AC}(t = 0) \cos \theta \cos \frac{\delta}{2} \right]}{|K| \cos \frac{\delta}{2} \sin \frac{\delta}{2} (\sin^2 \theta - \cos^2 \theta)}, \tag{8}$$

$$V_z = \frac{\Pi \left[f_{DC} \cos \theta \sin \frac{\delta}{2} - f_{AC}(t = 0) \sin \theta \cos \frac{\delta}{2} \right]}{|K| \cos \frac{\delta}{2} \sin \frac{\delta}{2} (\sin^2 \theta - \cos^2 \theta)}. \tag{9}$$

The principle of least squares is applied by fitting a rectified biased sinusoidal function to the HRWS data. Subsequently, the V_x, V_y, V_z wind components are determined from the coefficients obtained by the fitting process. The equations used are contained in the HRWS flight test report (to be published).

SYSTEM DESCRIPTION

A CO_2 laser heterodyne system has been constructed and mounted in a pod compatible with an AH-1 or a UH-1 helicopter. A block diagram of the system is shown in figure 8. The HRWS can be subdivided into three subsystems: the transmitter, the scanners, and the processor. The transmitter consists of a CO_2 laser and appropriate beam combining optics for heterodyne operation. Two scanners are employed: one scans in range and the other produces a conical scan. The processor includes the detector and its associated biasing network, frequency tracker, and a tape recorder.

The following paragraphs describe the individual components.

Laser

Other studies have determined that a 5-W laser would provide sufficient power to reliably measure helicopter flow fields, a previously developed Raytheon Company 5-W air-cooled laser with low weight and size and an extremely high degree of ruggedness and reliability was selected. This laser was designed as a result of numerous system analyses, which indicated a need for a 5-W lightweight, rugged, air-cooled laser, specifically designed for operation in a field environment. Among the primary design goals for such a device were:

1. All-metal/ceramic laser tube, replacing the fragile glass tube of laboratory lasers

2. Direct air-cooling of the laser tube, eliminating liquid coolants

3. Stable lightweight laser frame.

Raytheon has built and tested a laser meeting these requirements and has successfully flown it as part of a system.[1]

The laser frame consists of a magnesium alloy box within which four invar rods, connecting the two end plates, are contained. An all-ceramic alumina plasma tube is terminated by zinc selenide Brewster angle windows. The mirrors are contained in spherical mounts in the end plates and are prealigned and then permanently fixed in place. The clamp-on fins for cooling are of anodized aluminum. The laser is cooled by room air directed through the laser frame. A photograph of the CO_2 laser is shown in figure 9.

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[1] Contract 33615-73-C-1321, CO_2 Laser Heterodyne Sensor

Tests on this laser conducted during several programs have been impressive, with sustained output power as high as 6 W. The laser has been tested to determine its output power as a function of tube wall temperature, with no ill effects even though no cooling was supplied to the laser for periods in excess of 30 min.

Several design features are incorporated to achieve adequate laser lifetime. A gas reservoir is contained in the central portion surrounding the cathode. The laser uses a split discharge configuration with anodes on both ends and air cooling fins applied to both sections of the tube. A separate, high voltage power supply is used to power each half of the discharge tube. The units chosen for this application are modular, current-regulated supplies and contain an automatic starting circuit to initiate the laser discharge.

Laser lifetime in excess of 200 hr has been obtained with this laser. The lifetime is usually determined by decomposition of the gas; and when the power decreases significantly, the laser may be refilled through a valve on the laser tube incorporated into the design.

Heterodyne Optics

The heterodyne optics consist of a combination of beam splitters and polarizing elements that generate the local oscillator (LO) and the transmitter beams and recombine the collected signal with the LO. Measurements of the detector indicate that LO beams of approximately 1 to 2 mW provide the maximum signal-to-noise ratio (SNR). The reflectivity of the beam splitters has been selected to generate an LO beam of this value. The remainder of the energy is transmitted into a beam expander (described in the next paragraph). After the beam expander collects the scattered energy, the heterodyne optics combine LO and signal and direct this superposition toward the detector (described in paragraph entitled "Receiver"). The polarizing elements of the heterodyne optics assure proper polarization of all beams.

Beam Expander

Table 1 lists the specifications of the beam expander. The optical design resulted in a Galilean telescope with an input negative singlet and an output doublet. Figure 10 shows a back-reflection optical diagram for the surface reflection properties.

Range Scanner

Range scanning is accomplished by translating the input lens of the beam expander described in the previous paragraph. Motion of the input lens translates the location of the focused laser beam as shown

in figure 11. If the input lens were scanned linearly with time, too much time would be spent at closer ranges. Therefore, a nonlinear drive function is electronically generated, which gives the scan as a function of time as shown in figure 12. The results shown in this figure were obtained experimentally from the assembled HRWS.

The range scanner is controlled by a three-position switch. In the off position, the beam becomes collimated (focuses at infinity); in the manual position, a potentiometer permits the operator to focus the beam at any range. The automatic switch position generates a range scan in which the beam varies its focus from 1 to 33 m in approximately 1 s. Under all conditions, a voltage proportional to the range of the focus is available for display and recording.

Conical Scanners

The conical scan is generated by rotating a wedge counterclockwise relative to direction of propagation about the optical axis. The wedge specifications are given in table 2, and the configuration is shown in figure 13. The wedge is driven by a motor whose speed is variable from 0 to 2300 r/min. To insure adequate bearing lifetime, the delivered system was limited to a speed of 1300 r/min. The motor speed is adjusted by a potentiometer and can be displayed on the control panel. As the motor speed decreases below a few hundred revolutions per minute, the laser beam is automatically blocked for safety reasons, preventing propagation into the atmosphere. Although the beam blocking mechanism can be overriden for alignment procedures, a caution light on the control panel warns the operator. In addition to measurements of the motor speed, signals are available every 2.8 degrees of rotation to indicate the angular beam position.

Receiver

The receiver consists of a 10.6μm detector and the associated bias circuit and amplifiers. The requirements of high quantum efficiency and good frequency response dictated the use of a mercury-cadmium-telluride (HgCdTe) detector. The detector is mounted in a dewar which holds liquid nitrogen for a period of at least 6 hr. The electronic circuitry of the receiver was designed to minimize loss of SNR.

The present performance of commercially available metal dewars is very good. With capacities of 0.25 l, the hold time is 6 to 8 hr. Such a dewar fits into a 6-in. cube. Windows are easily dismounted if they need to be replaced. For future operational systems, other cooling techniques are available.

The function of the bias circuit is to provide a voltage supply to the detector and to provide the proper matching between the receiver preamplifier and the detector. A prime requirement in designing this

circuit is to protect the detector from excess voltages or currents that can result from removal of the LO, too high LO power, or loss of coolant in the detector dewar.

A Honeywell HgCdTe detector was purchased, and its parameters were measured. The results of this measurement are given in table 3, and graphs of detector characteristics are shown in figures 14, 15, and 16.

Frequency Tracker

This portion describes the frequency tracker and counter built for the HRWS. The equipment is capable of measuring fixed frequency or sinusoidally frequency modulated inputs. The measured characteristics for the tracker are listed in table 4.

Figure 17 is a block diagram of the frequency tracker. The input band is translated up to a center frequency of f_o by mixing the input with the output of a voltage-controlled oscillator (VCO). The signal is then limited and fed through a frequency discriminator centered near f_o. In tracking operations, the discriminator drives an integrator, the output of which drives the VCO. With the tracking loop closed, the sum of the input and VCO frequencies is constant.

Since the discriminator frequency range is much smaller than the input band, acquisition circuits are included to initially center the translated input in the discriminator band. A ramp generator sweeps the VCO over its range until the acquisition detector detects the presence of a signal in the discriminator band; at this time the loop is closed and the input tracked. The acquisition detector will revert to the search mode when the amplitude of the bandpass filter output is insufficient for tracking.

For automatic calibration purposes the input signal is periodically replaced by the output of a crystal oscillator. The readout circuit makes use of measurements made during this calibration interval to automatically correct for drift in the discriminator and integrator.

The frequency counter (figure 18) counts the positive zero crossings of the VCO during a 160μs period once each angle clock pulse. The sample gate generator provides this accurately timed counter gate along with a pulse immediately after it to clock the output storage register.

The calibration gate generator periodically generates a gate which turns on the 5-MHz calibration oscillator. After waiting for a period of time for the tracker to acquire the 5-MHz calibration signal and for the VCO to settle, a single 160μs sample of the VCO frequency is taken by the calibration counter. The calibration counter output is then subtracted from all frequency readings to automatically correct for discriminator center frequency drift.

A single-bit output is provided which declares data invalid during calibration and under loss of track.

The results of laboratory tests on the frequency tracker for the helicopter wind sensor are summarized. Test results include minimum SNR, frequency accuracy, dynamic range, and acquisition time. Measurements of the SNR (referenced to a 1-MHz noise bandwidth ±by1) required to track are shown in figure 19. The probability of tracking (P_T) is the fraction of the time which the tracker stays in its tracking mode. Due to the extremely sharp nature of the threshold levels at the high end of the band, these measurements were made only to 9 MHz. At 10 MHz, a 17-dB input SNR produced nearly 100 percent tracking.

Measurements of the accuracy of the counter output were made at two signal levels and three frequencies. The results are listed in table 5. Each output measurement shown is the average of five samples of the counter output. The worst error with self-calibration employed is 17.5 kHz.

A 2-MHz input signal was varied in level from -48 dBm to +2 dBm; a 25-kHz shift in measured frequency occurred over this 50 dB of input range. The tracker was then disabled and the harmonic content of the signal was examined with a spectrum analyzer. At the +2 dBm input level, the worst harmonic level generated in the preamp/translator circuitry appeared at a frequency 4 MHz above the translated fundamental. Its level was 24 dB below the translated fundamental.

To obtain the maximum sensitivity possible, the sweep excursion has been increased to 12 MHz; the VCO now sweeps from 36 to 48 MHz. The time required to sweep this range is 240μs, and the time required to dump the integrator is 15μs. The worst case acquisition time is, therefore, 255μs.

Pod

The integration of any system into the compact package required to fit into an AH-1 helicopter is expensive and time-consuming. Additionally, packaging will relate to only a given installation and will require modifications for alternative vehicles. Therefore, it was decided that for this phase of the program the most cost-effective and timely choice would be to build an airworthy pod, similar to the rocket pods currently used on the AH-1, to enclose the system. The mounting brackets from an existing rocket pod were used, and the remainder of the pod was fabricated. Figure 20 shows a picture of the pod before the sensor was installed.

Controls

The sensor in the pod is operated from a control panel (figure 21) in the helicopter. All operations on the system are performed from the control panel except filling the detector with liquid nitrogen, which was done during aircraft preflight. The control panel also has a number of diagnostic capabilities which may be used by maintenance personnel. The controls and their functions are shown in table 6.

The data may be obtained from the data panel as shown in figure 22. These data are suitable for recording on a tape recorder such as the Ampex CP-100 which was used during the flight tests.

Integrated Sensor System

Several photographs of the assembled HRWS system are presented. Figure 23 shows the system with the pod doors open so that details of the components are visible. Figure 24 shows a closed pod. Figure 25 shows the system attached to a UH-1 helicopter during tests. The tape recorder, used for data collection, is clearly visible in figure 26.

CONCLUSIONS

A CO_2 laser Doppler velocimeter has been successfully miniaturized and packaged for airborne operations. The HRWS has successfully functioned as designed in flight tests. The HRWS will measure wind speed and direction and potentially can augment a fire control system for attack helicopters. The HRWS can readily be used as a diagnostic tool for aircraft wind field investigation and as a ground-based wind sensor to measure atmosphere winds.

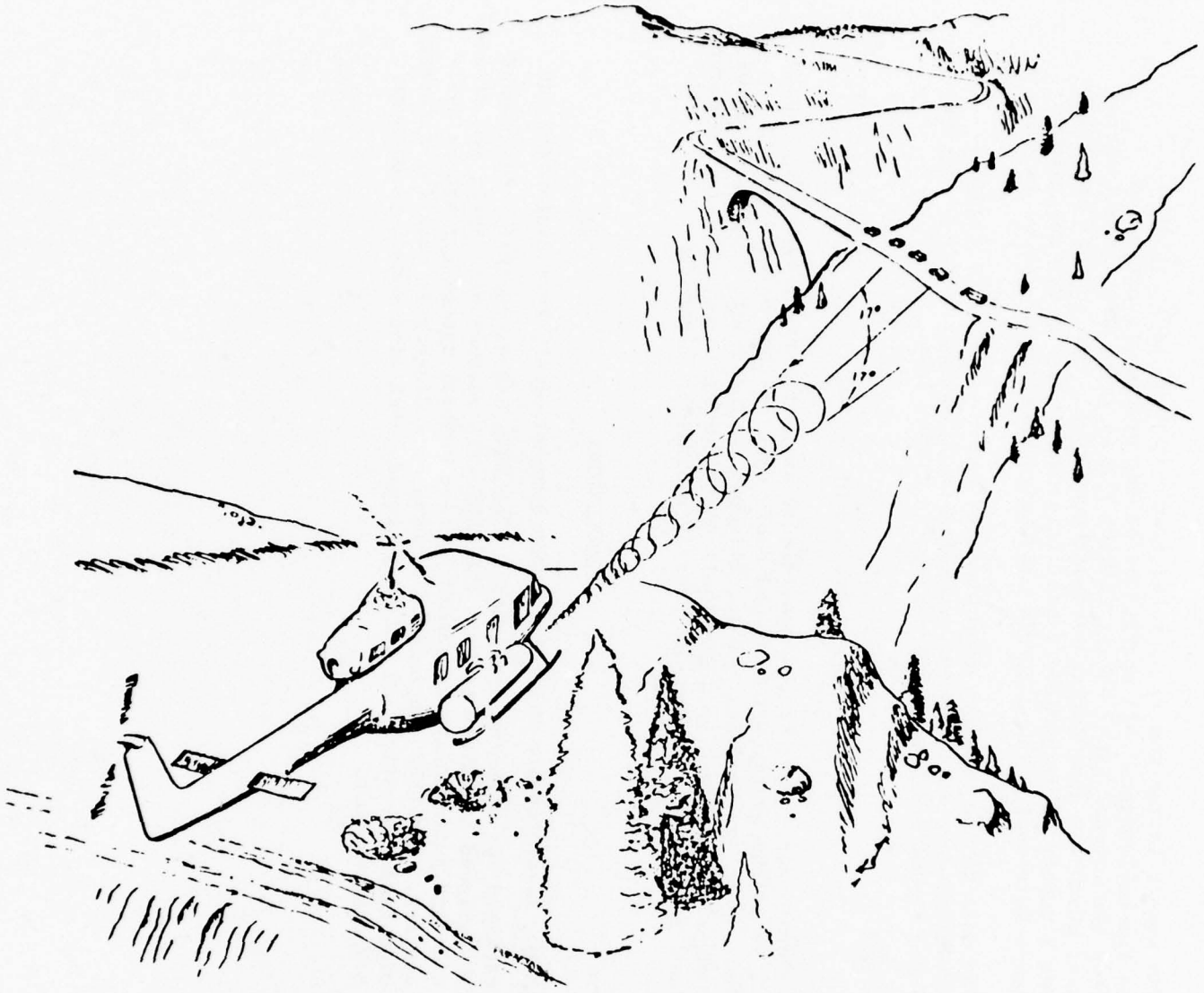
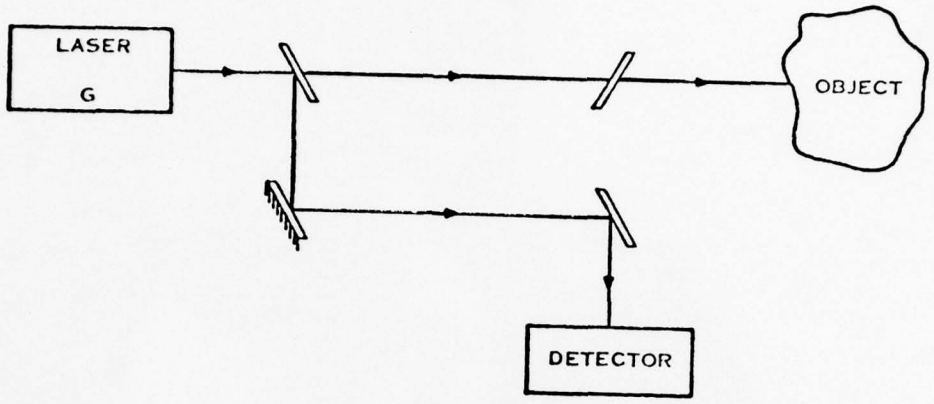
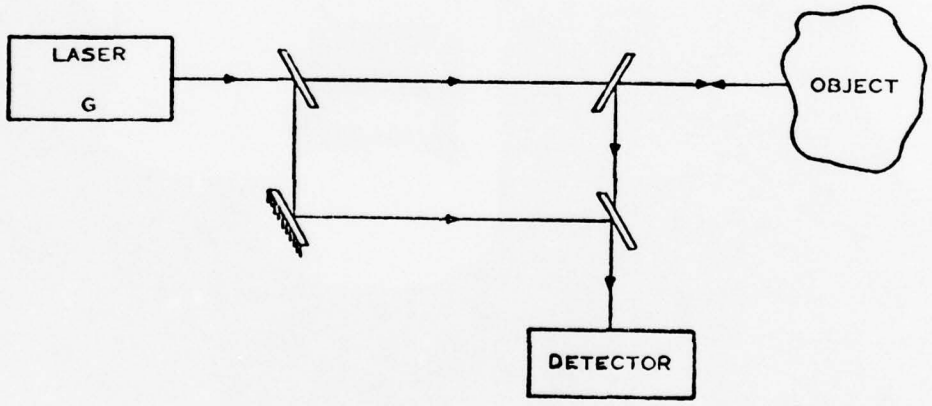


Figure 1. Artist's concept of helicopter remote wind sensor (HRWS).

Figure 2. Helicopter remote wind sensor (pod open).



A. Paths of transmitted beam.



B. Paths of transmitted and received beams.

Figure 3. Laser Doppler velocimeter transmission and reception.

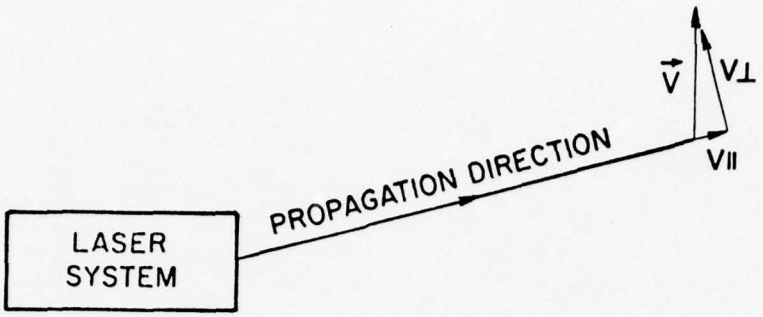


Figure 4. Laser Doppler velocity measurement.

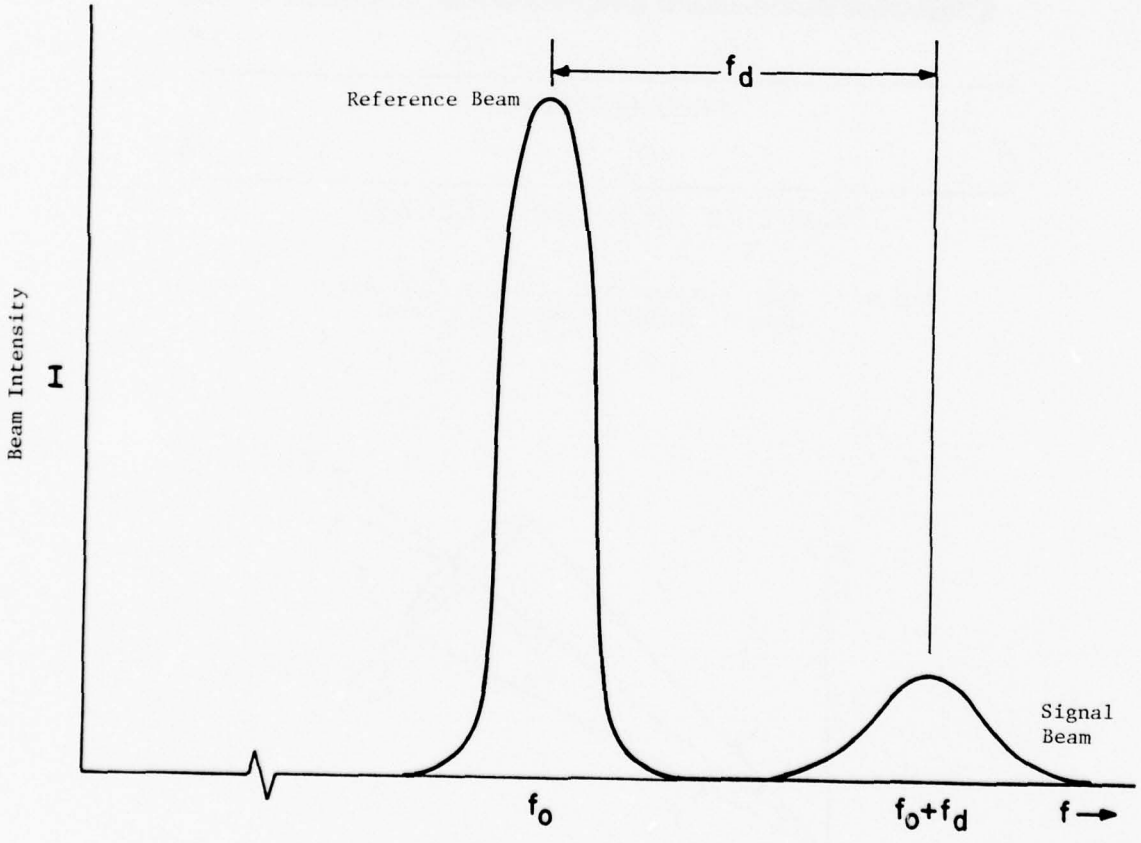


Figure 5. Illustration of optical heterodyning.

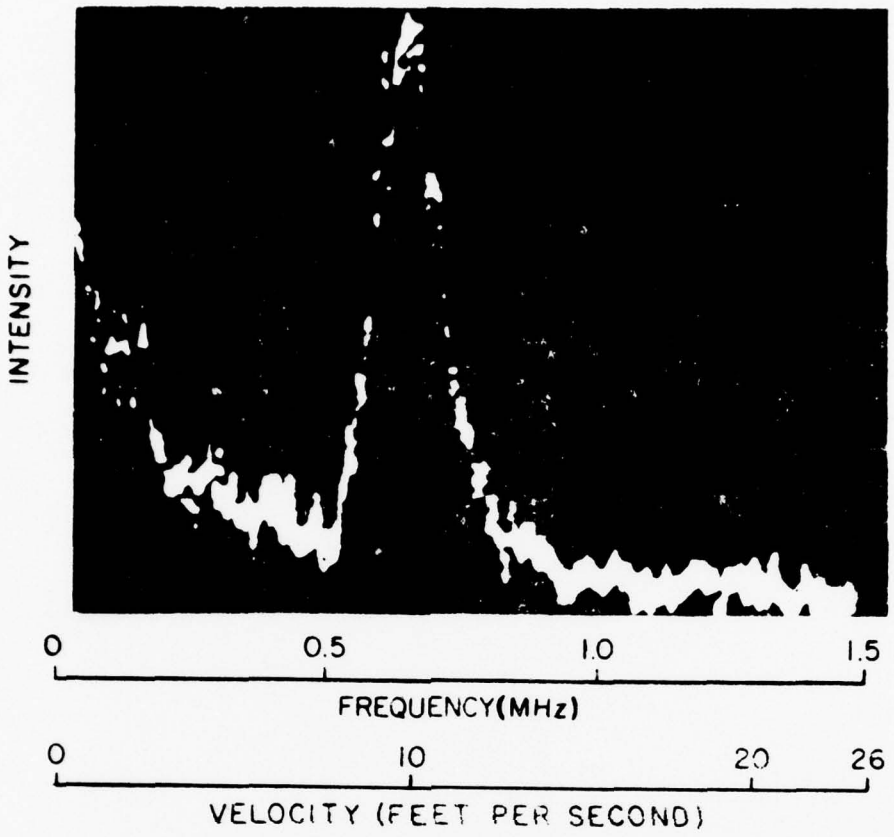


Figure 6. Sample ground wind velocity distribution from a coaxial nonscanning system.

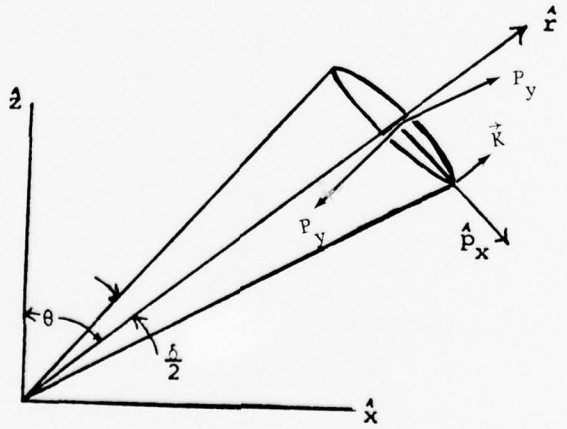


Figure 7. System geometry.

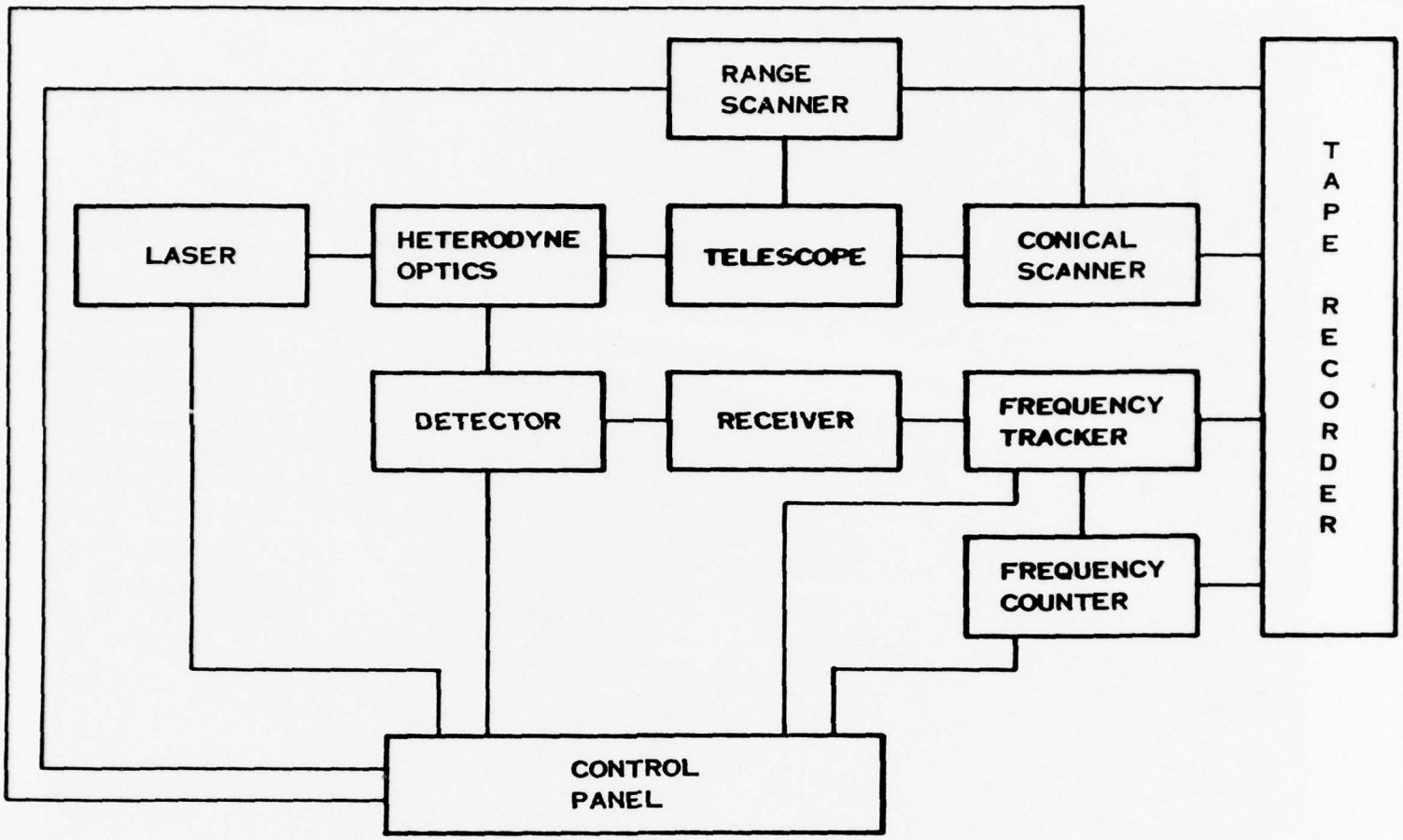


Figure 8. Block design of helicopter remote wind sensor.

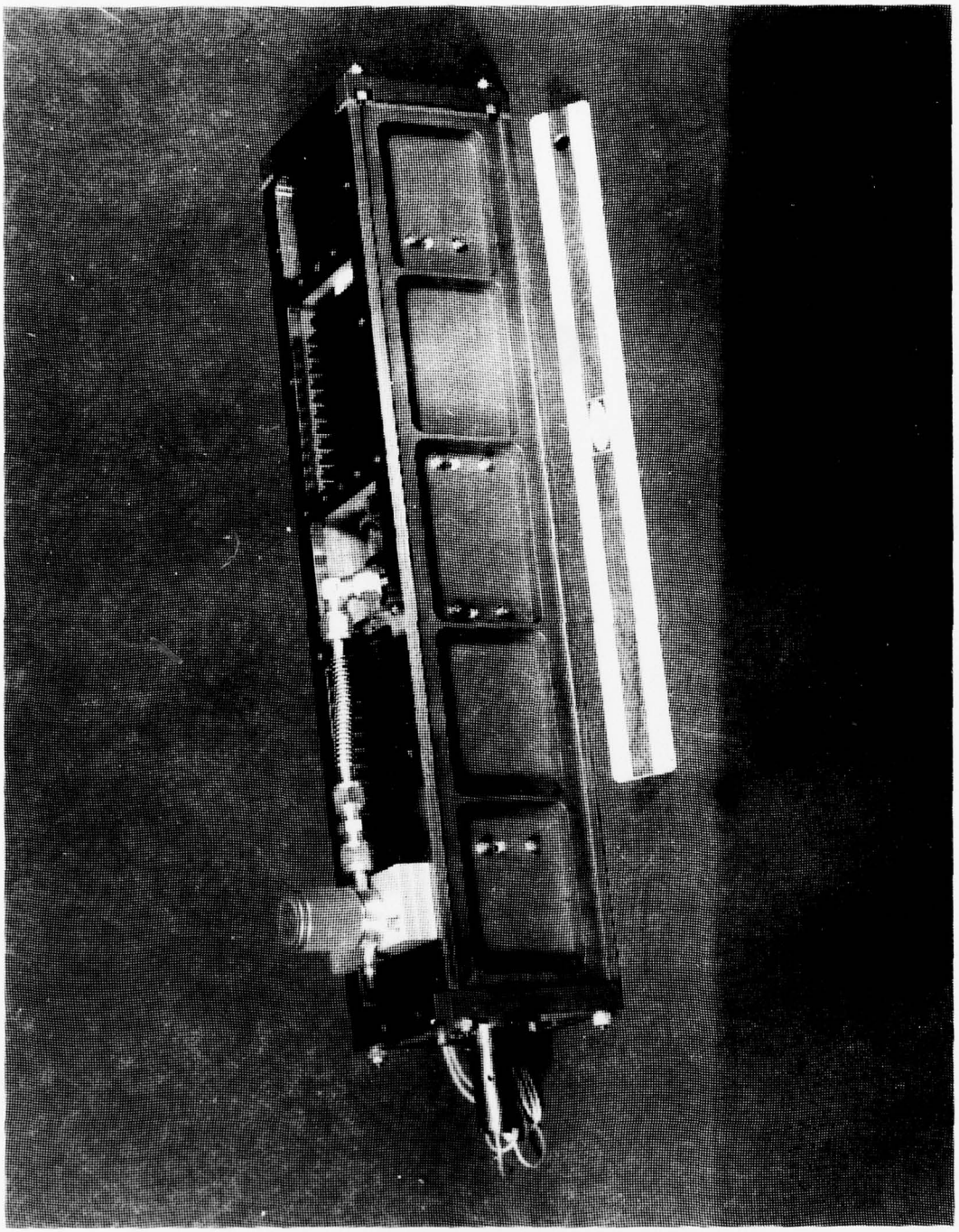


Figure 9. CO_2 laser (scale [in photo] is 12 inches).

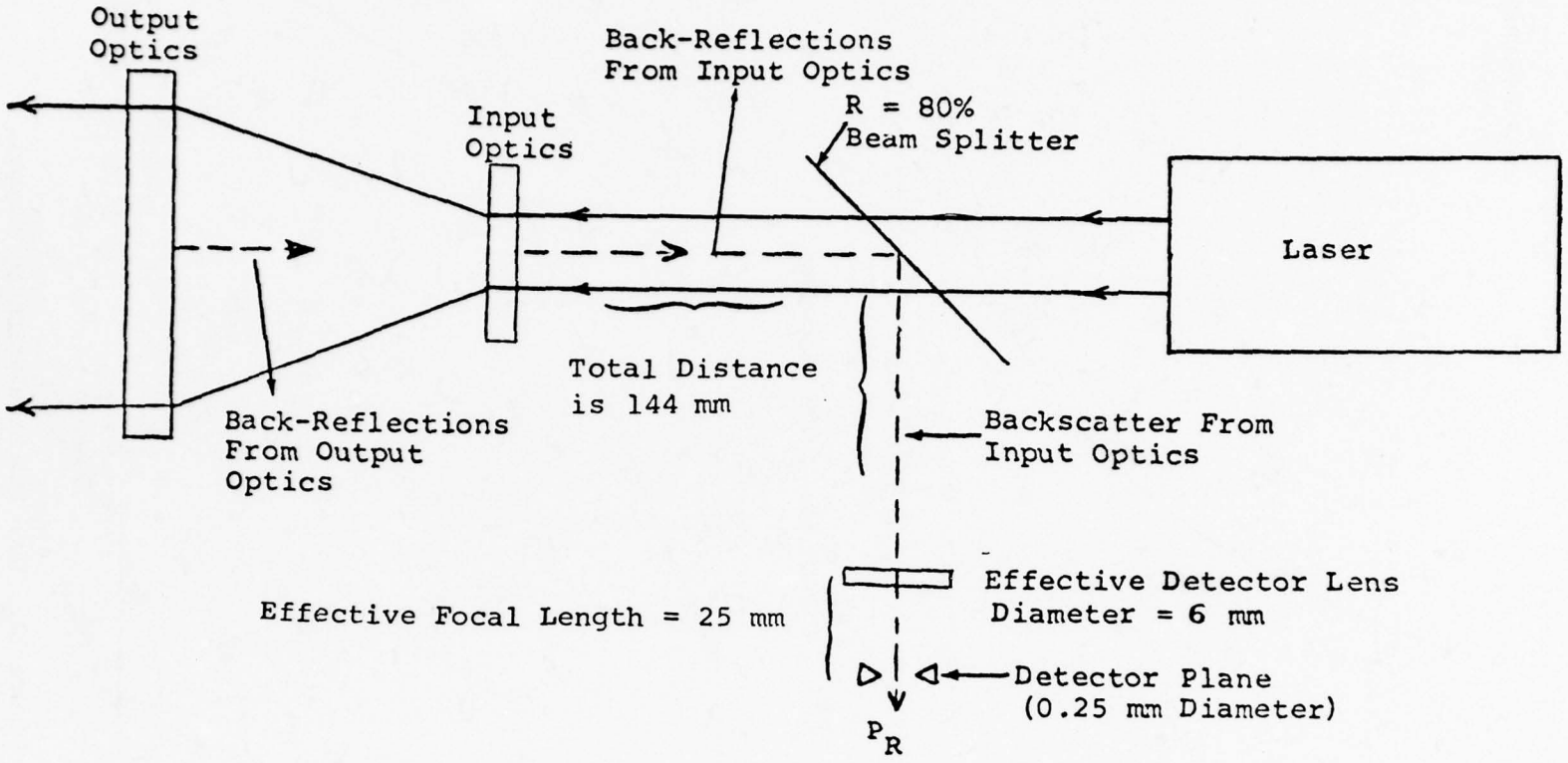


Figure 10. Back-reflection optical diagram.

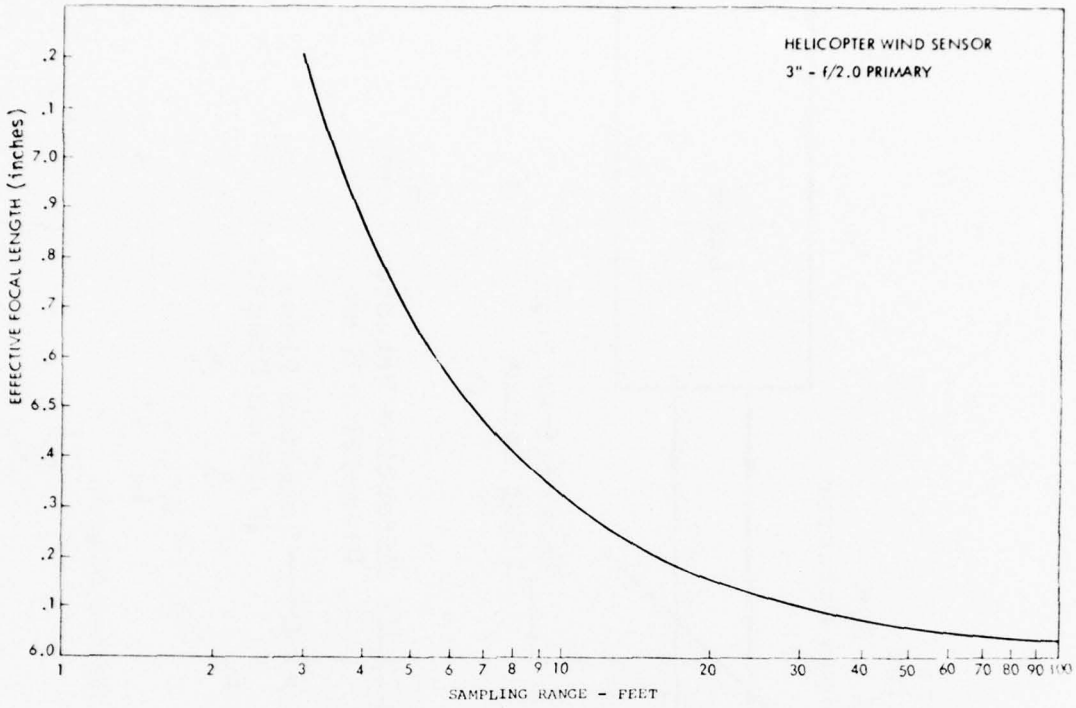


Figure 11. Range scanning.

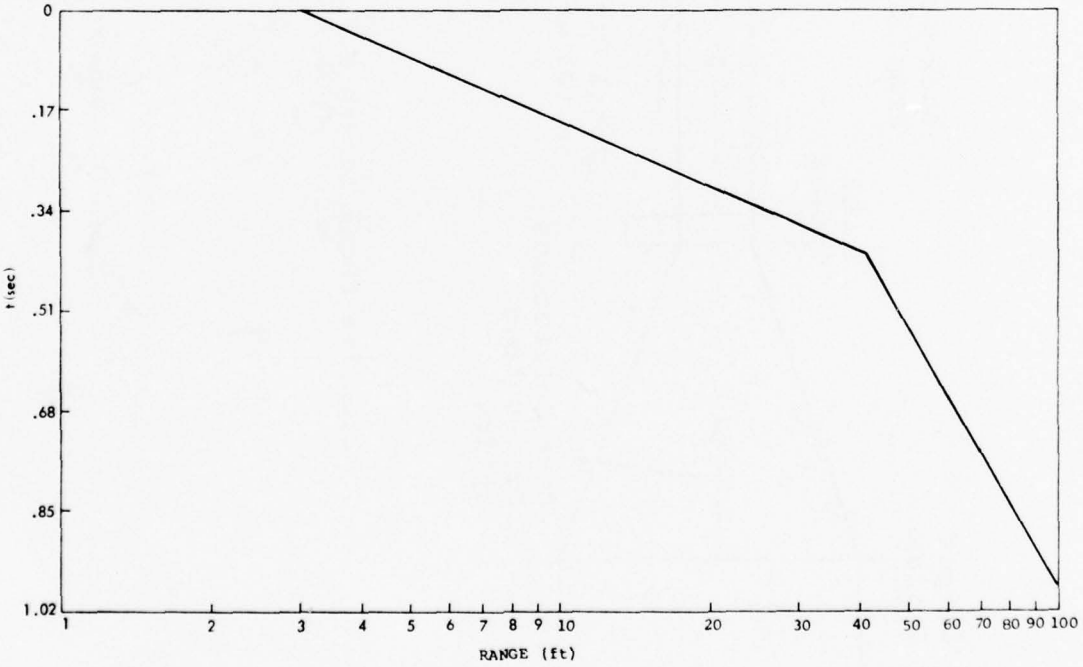


Figure 12. Time function of range scanner.

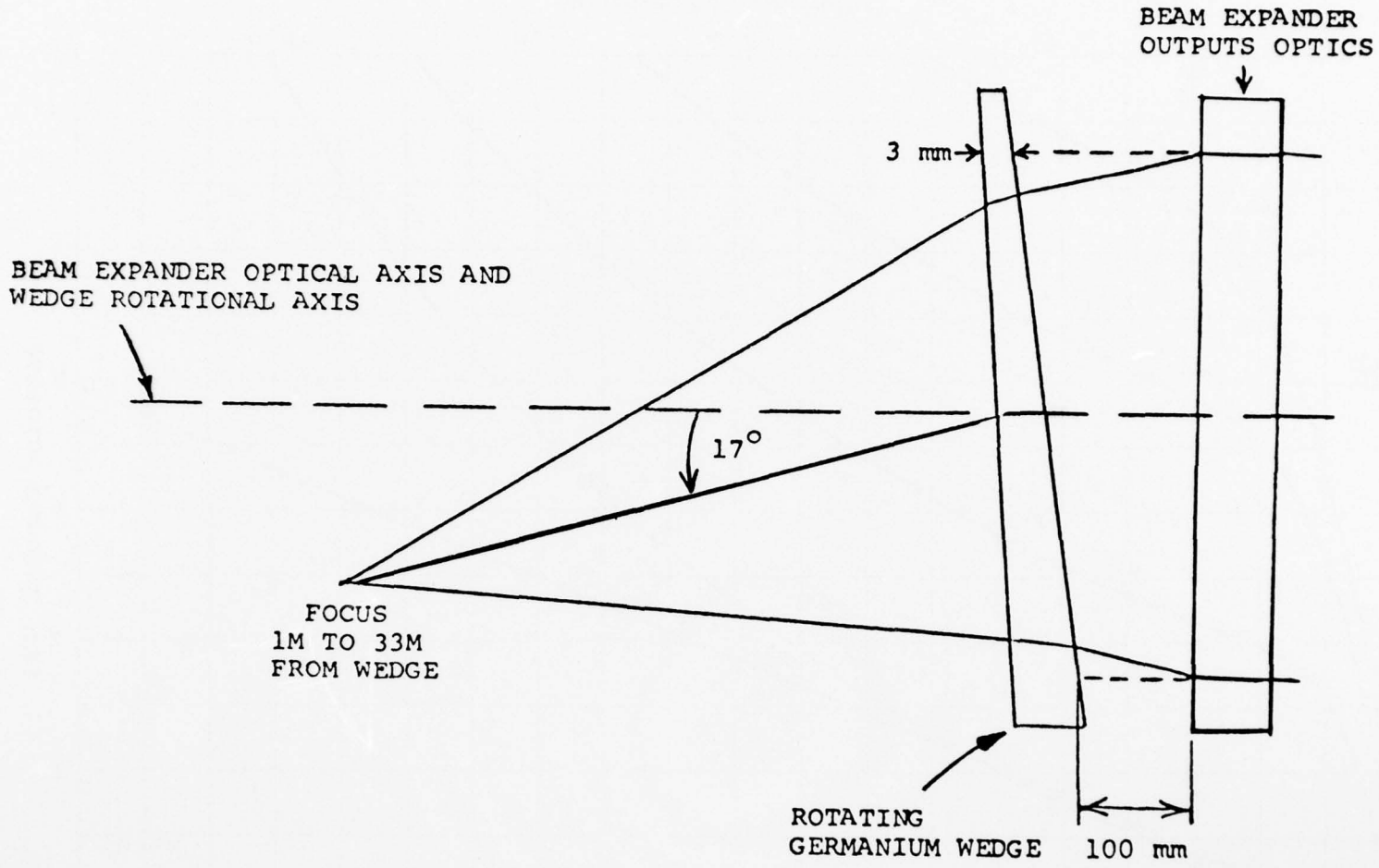


Figure 13. Angle scanner optical layout.

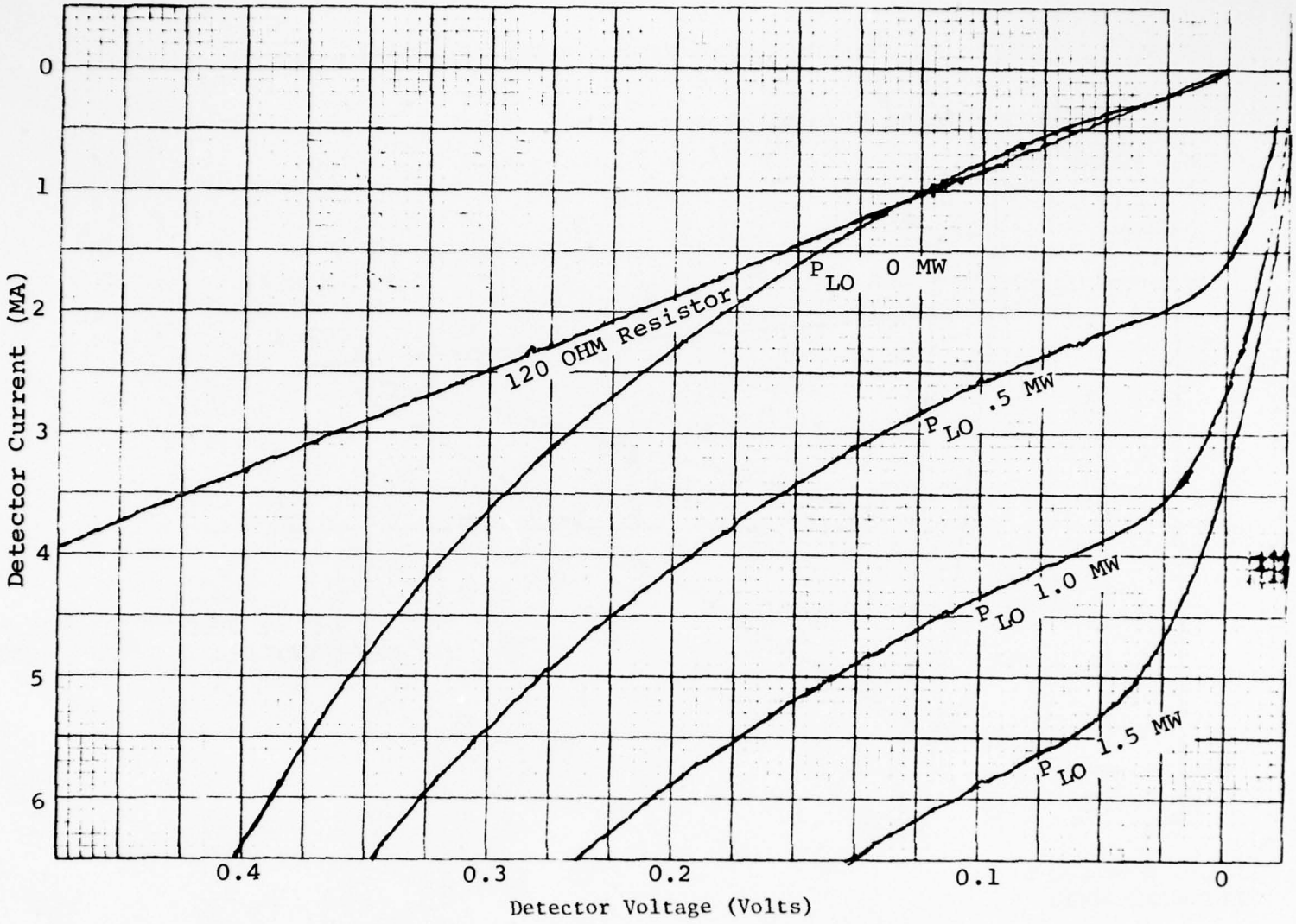


Figure 14. Detector characteristics.

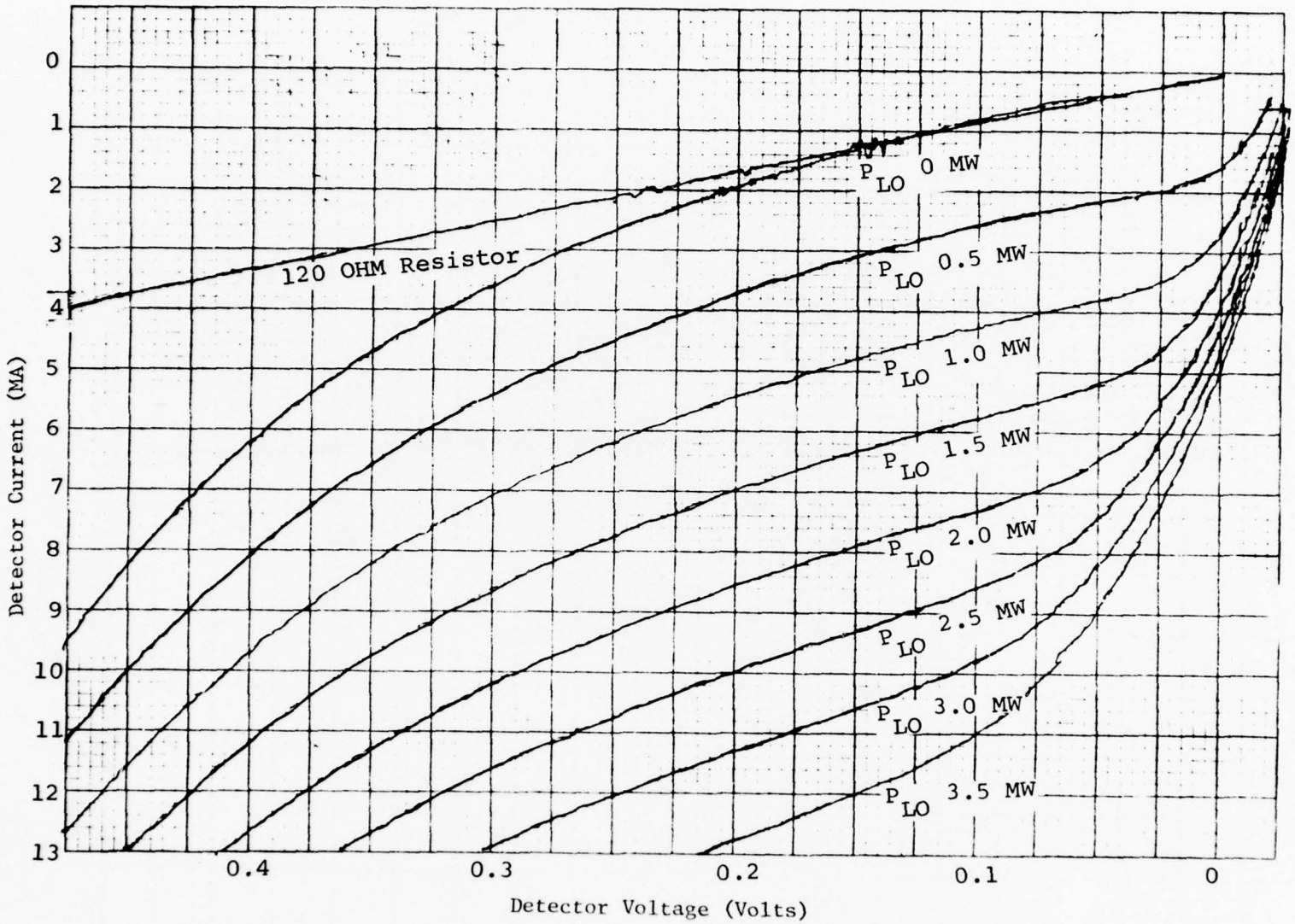


Figure 15. Detector characteristics.

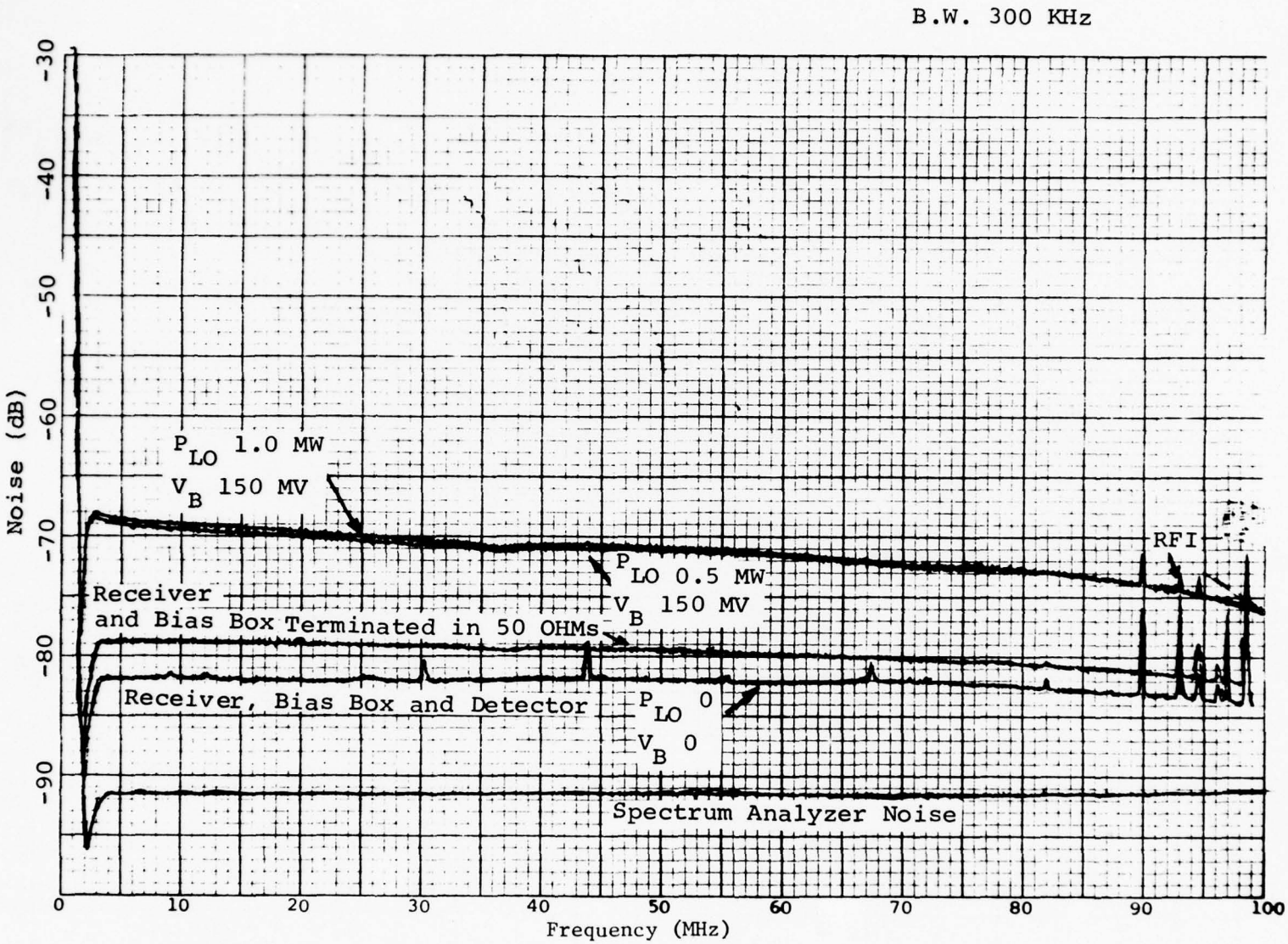


Figure 16. Detector characteristics.

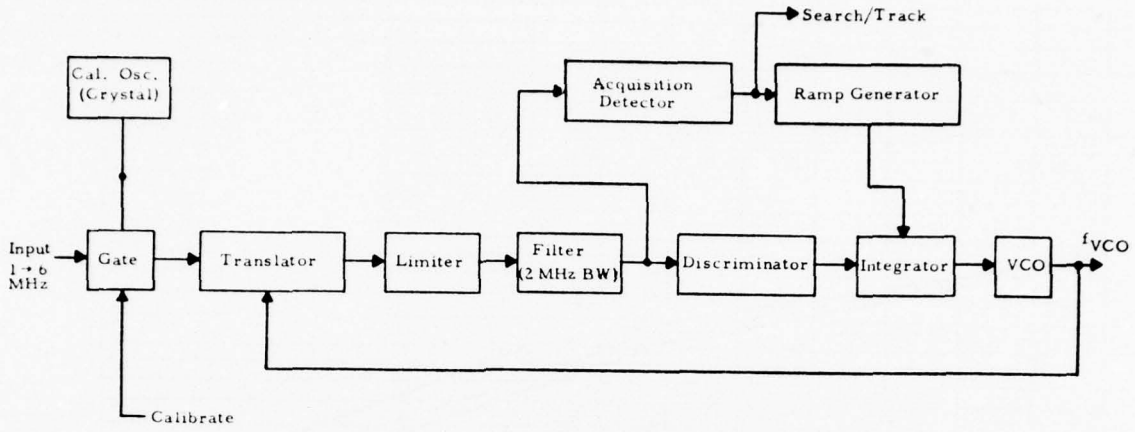


Figure 17. Remote wind sensor frequency tracker.

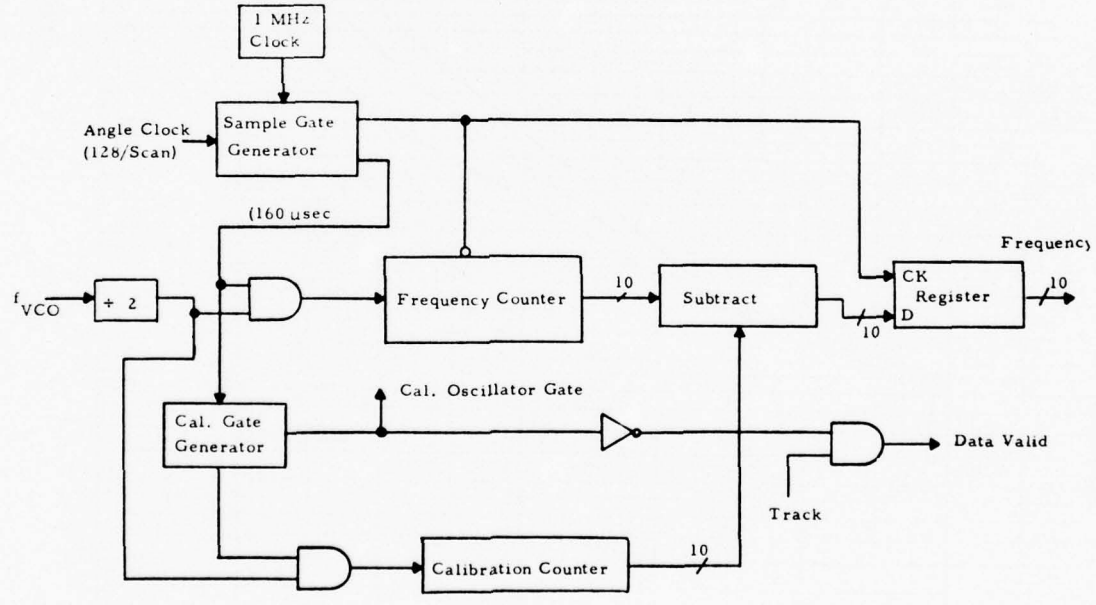


Figure 18. Frequency counter.

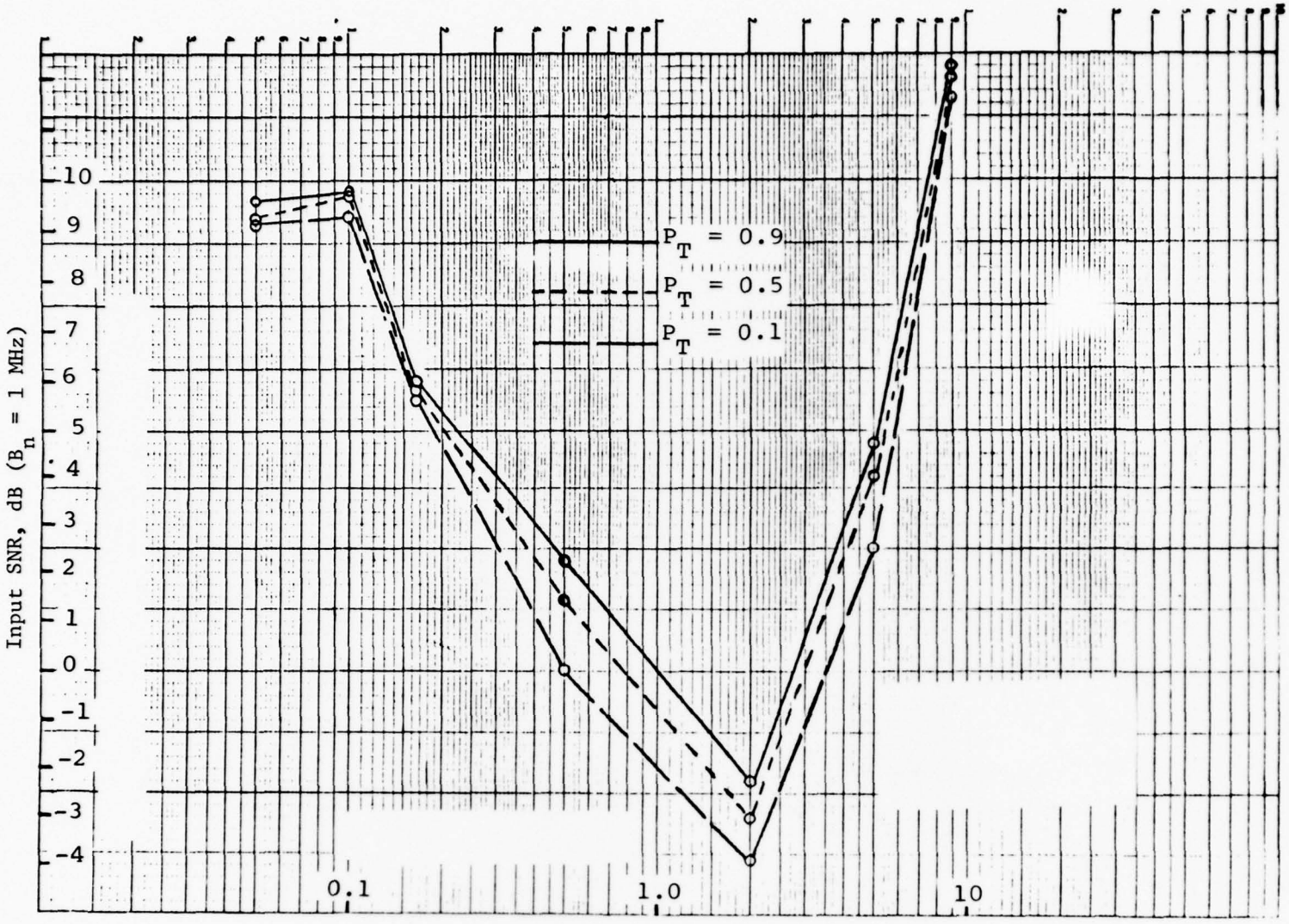


Figure 19. Minimum SNR vs frequency.

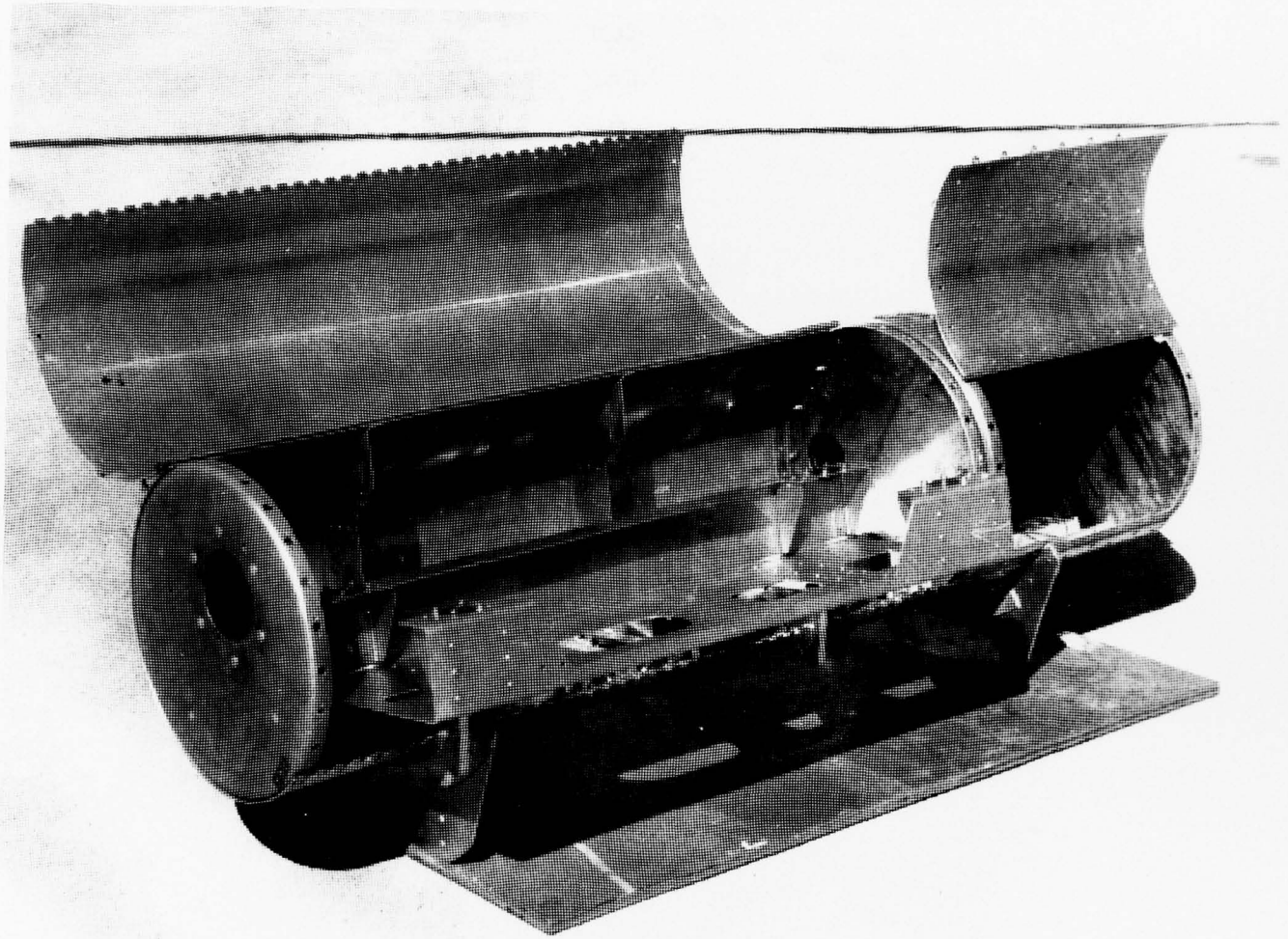


Figure 20. Helicopter pod.

Figure 21. HRWS control panel.

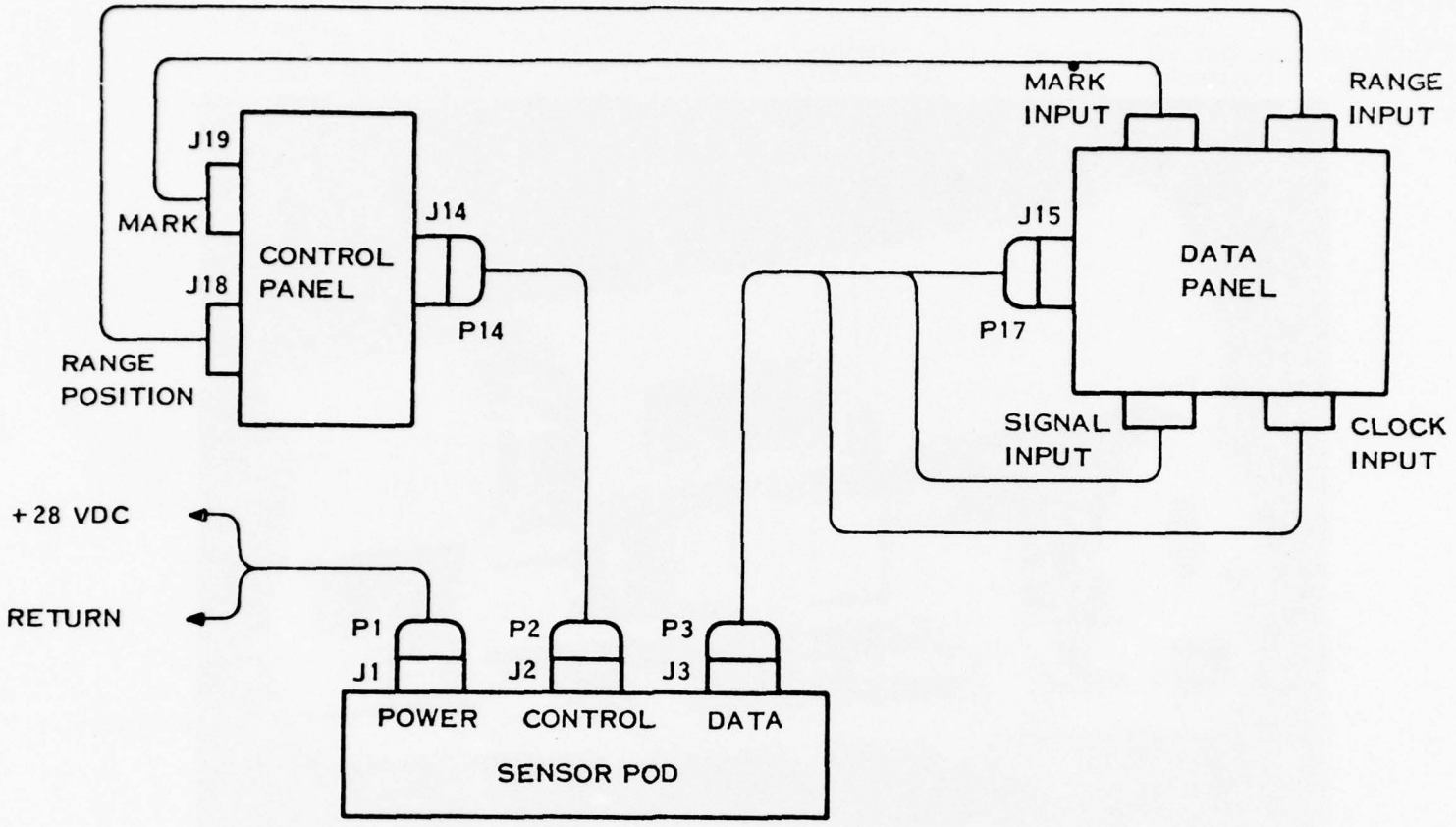


Figure 22. System interconnection.

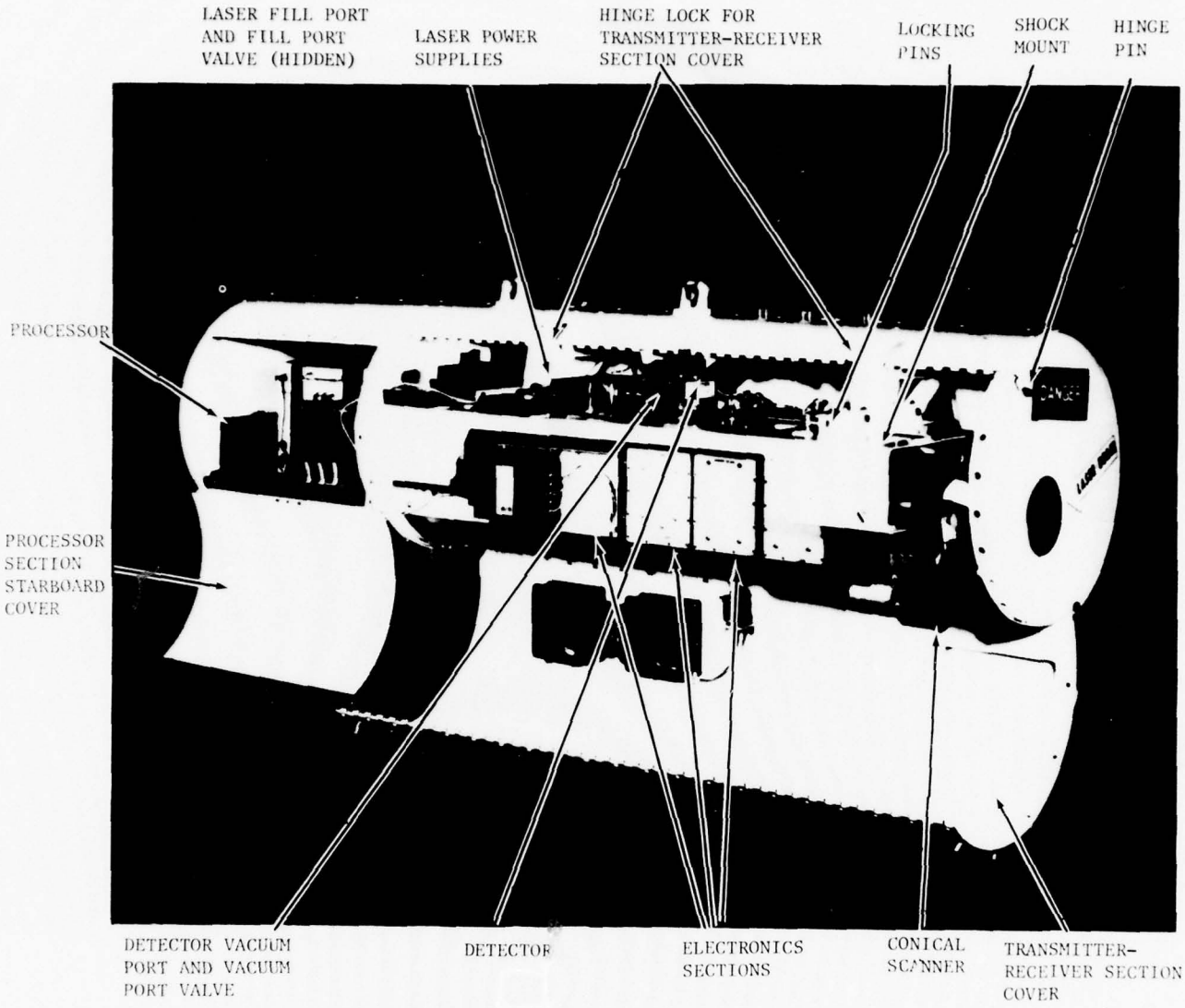


Figure 23. Helicopter carried remote wind sensor with corners open.

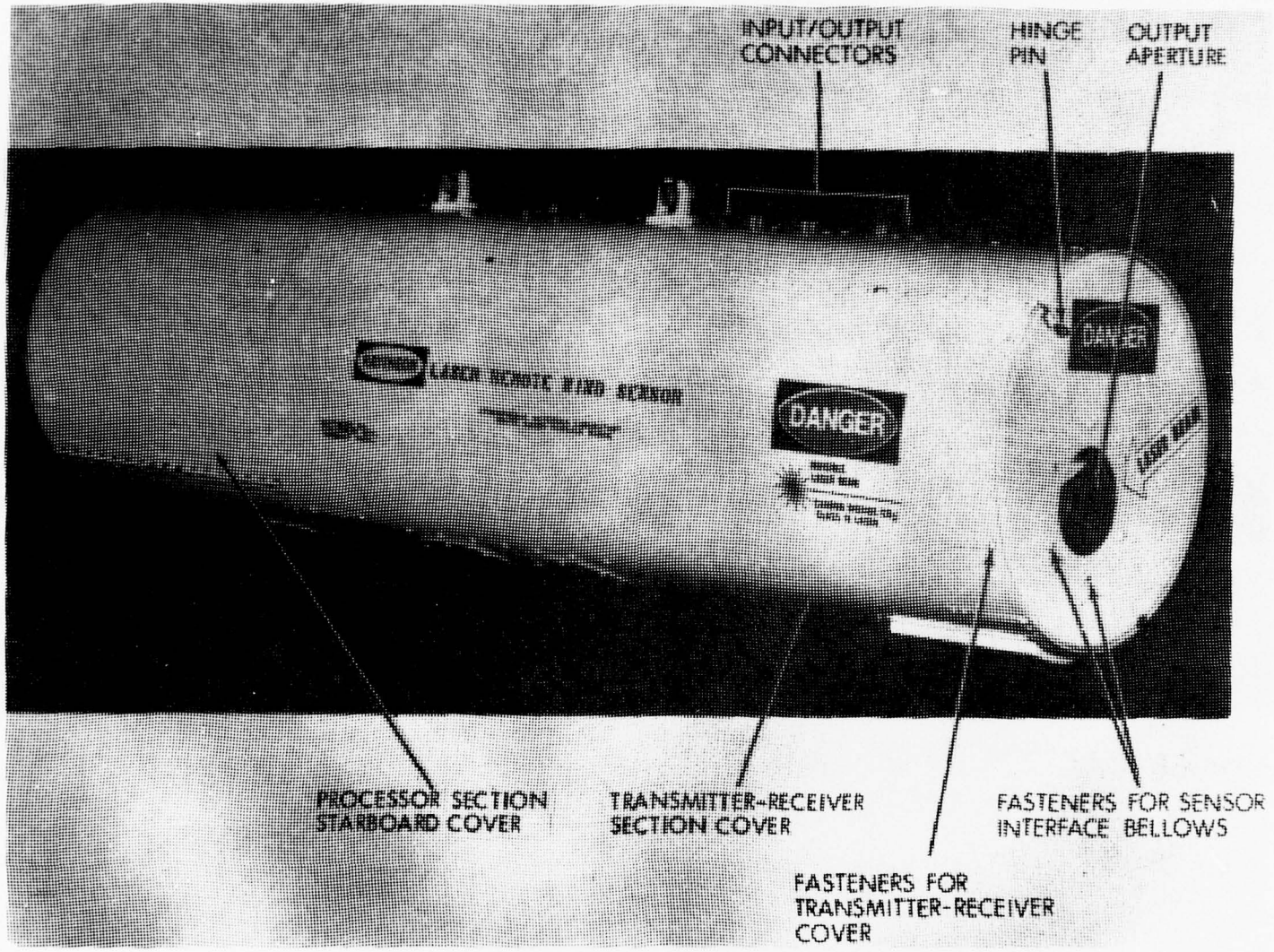


Figure 24. Helicopter carried remote wind sensor.

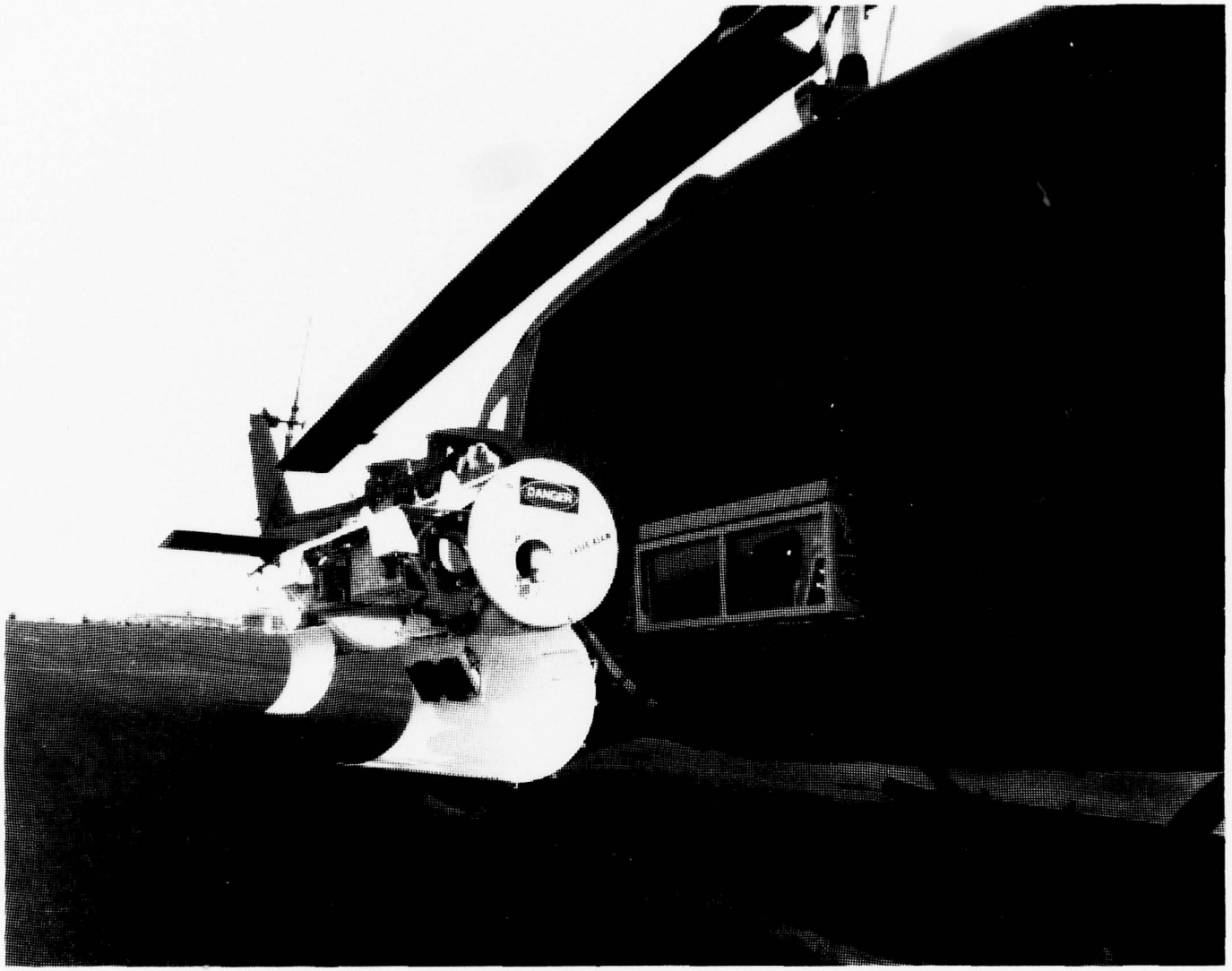


Figure 25. Mounted helicopter remote wind sensor.

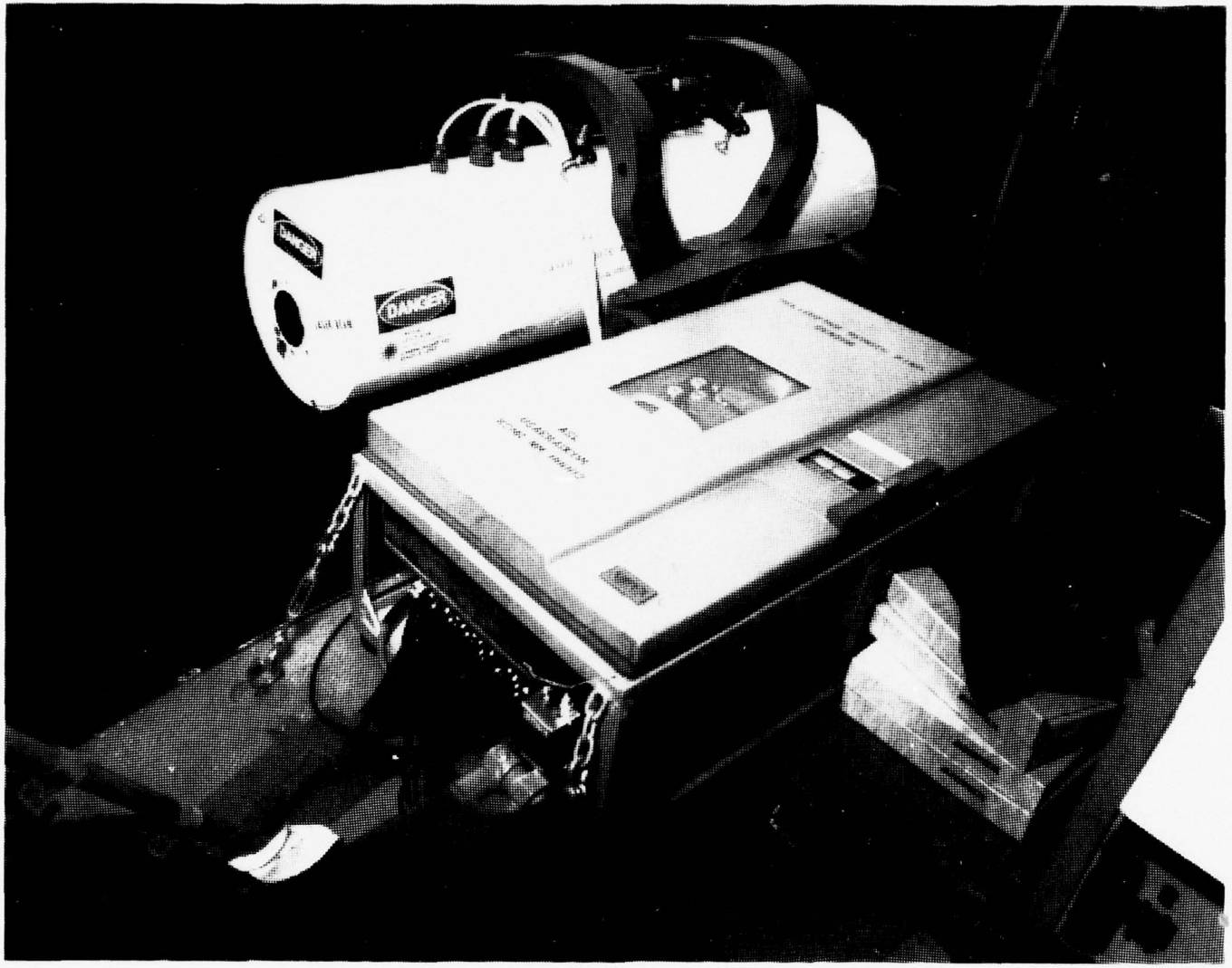


Figure 26. Sensor and tape recorder.

TABLE 1. BEAM EXPANDER SPECIFICATIONS

Input Beam Properties	
Diameter	2.6 mm @ $1/e^2$ points
Shape	Gaussian; TEM_{00} mode
Wavelength	10.6μm
Power	6 W, CW
Wavefront	Nearly plane wave
Output Beam Properties	
Diameter	66 mm @ $1/e^2$ points at ∞ focus (with no diffraction effects)
Focus	1 to 32 m goal, 2 to 32 m firm
Aberrations	The diameter of the blur circle in which 87% of the total power is contained shall be less than:
	$\frac{2.6\ \lambda R}{D}$ 2 times diffraction limit for Gaussian beam truncated at $1/e^2$ points
where λ = 10.6μm D = 83 mm R = Range to focus	
Input Optics	
Clear aperture	5.2 mm
Material	Germanium
Coatings	AR $\leq$ 1/2% per surface
Field angle	On axis
Focal length	Negative

Table 1 (cont)

Output Optics

Clear aperture	83 mm
Material	Germanium
Coatings	AR $\leq$ 1/2% per surface
Field angle	On axis
Focal length	Positive: effective focal length = 153 mm

Paraxial Magnification

25.4 @ ∞ focus range

Surface Reflection Properties

1. The input optics should minimize the back-reflected power onto the detector to less than 50μW (see figure 10).

2. The output optics back-reflected power onto the detector should be less than 100μW.

TABLE 2. WEDGE SPECIFICATIONS

Total beam deviation	17 deg
Material	Germanium
Clear aperture	83 mm
Closest distance to beam expander output optics	100 mm
Minimum wedge thickness	3 mm
Focus range	1 to 32 m

TABLE 3. MEASURED DETECTOR PARAMETERS

Detector mfg	Honeywell Inc., Radiation Center
Detector type	LK146E9, Mercury Cadmium Telluride (HgCdTe)
Detector serial No.	T-3

Detector Parameter	
Quantum efficiency*	42%
Frequency response (-3 dB)*	70 MHz
Maximum LO power	$\geq$ 5 mW
Operating wavelength	P-20 (10.6μm)
Dark current**	1.3 mA
Dynamic resistance*	83 Ω
Noise***	-9.8 dB
Series resistance	10.4 Ω
Static resistance*	30.6 Ω

*At operating point of 1 mW LO power with 0.15-V reverse bias and a detector current of 4.9 mA.

**At 0.15-V reverse bias.

***Departure from LO shot noise limited operation. Biased as in * above and at a frequency of 10 MHz.

TABLE 4. TRACKER PARAMETERS

Input	
Frequency coverage (-3 dB)	0.1 to 8.5 MHz
Spectral width	≤ 2 MHz
SNR (1 MHz noise bandwidth)	8 dB
Rate of frequency change	500 MHz/s
Scan rate	20 to 40 Hz
Frequency deviation (P-P)	4 MHz
Discriminator	
Center frequency	60 MHz
Bandwidth	2 MHz
Acquisition	
Acquisition time (max)	255μs
VCO sweep rate	50 kHz/μs
Samples integrated	80
Tracker	
Time constant	100μs
Angle lag (max)	1.5 deg
Frequency lag (@ max df/dt)	50 kHz
Output	
Samples/scan	128
Maximum rate	5120 Hz
LSB	12.5 kHz
Bits	10

TABLE 5. FREQUENCY MEASUREMENTS

Input SNR* (db)	Actual Frequency (MHz)	Measured Frequency	
		Calibrated (MHz)	Uncalibrated (MHz)
15	8.000	8.005	8.048
35	8.000	8.000	8.040
15	1.000	1.0075	1.050
35	1.000	1.0075	1.050
15	0.100	0.1175	0.160
35	0.100	0.1125	0.1475

*Referenced to 1 MHz noise bandwidth

TABLE 6. CONTROLS FOR HRWS

Control	Positions	Function
Laser	On-off	Turn on HV to laser
LVPS	On-off	Turn on low voltage power supplies for control circuits
Sensor	Run	For normal operation
	Pwr	Inserts beam diverter to measure power of transmitter
	No scan	Overrides beam blocking system to permit nonscanning measurements
		Conical scan switch must be on; RPM adjust at zero
Range	Auto	Performs approximately 1 Hz scan
	Manual	Moves scanner to manually selected range
	Off	Defocuses beam for safety
RPM adjust		Continuously variable RPM from 0 to 1200 for conical scanner
PZT adjust		Adjusts piezoelectric transducer voltage to optimize laser operation
Range adjust		Adjusts position of range scanner; only operational in manual RANGE mode
Conical scan	On-off	Turns on scan motor; CAUTION preset RPM adjust to zero
Proc	On-off	Turns on processor
	Cal	Turns on 5 MHz oscillator and turns off input signal
Cal test	Test	Turns on input signal

Table 6 (Cont)

Control	Positions	Function
Cal disable		Disable operation of the self-calibration circuit.
Clock disable	Down to disable	Disconnects 2-deg clock pulse and replaces it with internal clock
Sweep disable		Disables VCO sweep and sets VCO to 48 MHz
Track disable		Disables acquisition logic so VCO sweeps continuously; pin J3 of the translator can be displayed on an oscilloscope to show the entire spectrum with 2 MHz resolution.
Low air light		Indicates insufficient air flow for laser cooling; the sensor is operable with this light illuminated.
No scan light		Indicates that the no-scan mode is activated and a hazard condition can exist if the transmitted beam is not adequately blocked

Control	Position	Measured Quantity	Units
Monitor selector	I_A, I_B	Laser current	Milliamperes
	Pwr	Laser power	Watts
	V_D	Detector voltage	Volts
	I_D	Detector Current	Milliamperes
	RPM	Scan speed	Revolutions per minute
	Range	Range	See calibration
	-15 V, +15 V, +5 V	Low voltage power supplies	Volts

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