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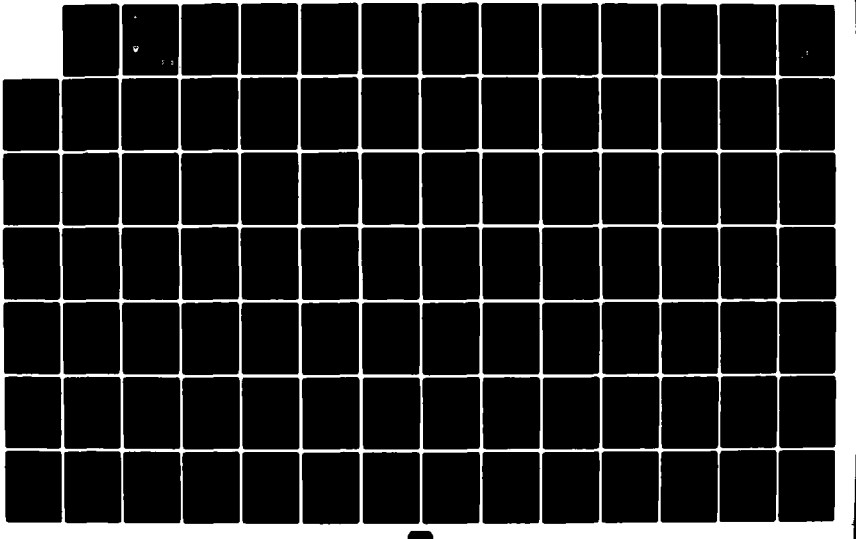

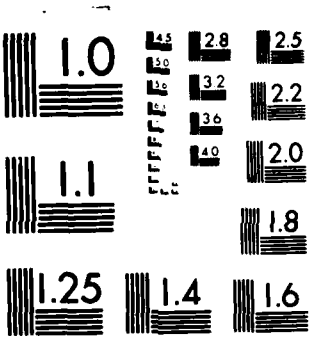

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TECHNICAL REPORT RE-83-7

ANALYTICAL RESEARCH BY COMPUTER SIMULATION OF DEVELOPMENTAL POLARIMETRIC/FREQUENCY AGILE PULSED RADARS

R. F. Russell and F. W. Sedenquist
Advanced Sensors Directorate
US Army Missile Laboratory

December 1982

U.S. ARMY MISSILE COMMAND
Redstone Arsenal, Alabama 35809

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A new generation of radar systems that exploit the polarization characteristics of various targets and clutter are under development. This report examines the methods of simulating these new techniques and presents typical results.

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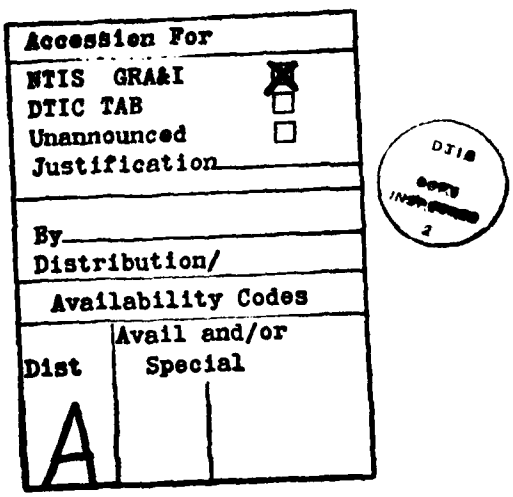


CONTENTS

Page No.

I. INTRODUCTION 1

II. MATH MODEL DEVELOPMENT 1

A. Polarization Definition 1
B. Plane Waves 1
C. Special Cases of Polarization 3
D. Polarization Notation 6
E. Scattering Matrix. 7
F. Scattering Matrix for Simple Objects 8
G. Radar Range Scaling. 11
H. Noise Generation 12
I. Antenna Isolation 13
J. Frequency Agility and Intra-Range Resolution 14

III. RADAR SIMULATION 15

A. Signal Processing 15

IV. SIMULATION UTILIZATION 18

A. General Simulation Outputs 18
B. Antenna Isolation. 19
C. Signal to Noise Ratio. 19
D. Signal to Clutter. 19

V. CONCLUSIONS AND RECOMMENDATIONS 19

APPENDIX - Simulation Flow Charts and Program Listing 59

I. INTRODUCTION

Radio frequency (rf) systems that utilize the polarization characteristics of the target and environment to detect, track, identify, etc., are referred to as rf polarimetric systems. These systems usually combine the polarization characteristics with frequency agility for increased range resolution. Examples of such radar systems are the Multi-environment Active Radio Frequency Seeker (MARFS), the Advance Indirect Fire System (AIFS), the Helicopter All Weather Fire Control and Acquisition Radar (HAWFCAR), and the Polarimetric Technology Seeker (PTS) as well as various other R&D radars under development by numerous independent contractors in private industry.

The demand for a more complete understanding of the techniques and processess employed in various programs has precipitated the development of the polarimetric radar simulation. This document covers the mathematical analysis required as background, the computer simulation model, and typical results. Recommendations for future expansions of this model are also addressed.

II. MATH MODEL DEVELOPMENT

A. Polarization Definition

The concept of polarization and the associated conventions are vital to the understanding of the use of the polarization scattering matrix. The definitions of polarization have been traditionally either the physics or the engineering convention. Either convention will provide the same general answer but with different notation. Therefore, the convention to be used throughout this analysis is as stated in the IEEE STD 211-1977 "IEEE Standard Definitions of Terms for Radio Wave Propagation".

Linearly Polarized Wave - An electromagnetic wave whose electric and magnetic field vectors always lie along fixed lines at a given point. (Page 9.)

Left-handed (counterclockwise) polarized wave - An elliptically polarized electromagnetic wave in which the rotation of the electric field vector with time is counterclockwise for a stationary observer looking in the direction of the wave normal.

NOTE: For an observer looking from a receiver toward the apparent source of the wave, the direction of rotation is reversed. (Page 9.)

The definition of right-handed is found on page 12 and is the same as above with the word clockwise used instead of counterclockwise.

B. Plane Waves

For a plane time harmonic electromagnetic wave traveling in free space the electric field intensity vector $\bar{E}(t)$, and the magnetic field intensity vector $\bar{H}(t)$ are always orthogonal to one another and have directions specified by the right hand rule as defined in the complex Poynting vector ($\bar{S}$).

$$\bar{S} = \bar{E} \times \bar{H}$$

Since $\overline{E}$ and $\overline{H}$ are always coupled together, it is customary to specify the $\overline{E}(t)$ vector only in describing the plane wave. The plane wave can be specified by its amplitude, frequency, direction of propagation, and polarization.

The vector wave equations for waves in free space* can be written as

$$\nabla^2\overline{E} + k^2\overline{E} = 0$$

$$\nabla^2\overline{H} + k^2\overline{H} = 0$$

where k is the complex wave number. The rectangular components of $\overline{E}$ and $\overline{H}$ satisfy the complex scalor wave equation (commonly called the Helmholtz equation):

$$\nabla^2\Psi + k^2\Psi = 0$$

The solution to the Helmholtz equation for one component, say x, thus reduces to

$$\frac{d^2E_x}{d_z{}^2} + k^2E_x = 0$$

which is the one dimensional Helmoholtz equation. The equation has solutions that are linear combinations of e^{jkz} and e^{-jkz}. We can choose to work with either of these solutions, though, in engineering we generally use the form e^{-jkz}: in particular, consider the solution

$$E_x = E_o e^{-jkz}$$

This satisfies the $\nabla\cdot\overline{E}=0$, and is therefore a possible electromagnetic field.

To interpret this solution, let E_o be the rms value; then the instantaneous field is found to be

$$\xi x = \sqrt{2}\, E_o \cos(wt-kz)$$
$$\xi y = \sqrt{2}\, E_o \cos(wt-kz)$$

For conventions' sake, the x direction will be the horizontal polarization and the y direction will be the vertical polarization. In general the two waves need not have the same phase. Again, for convention, it will be assumed that all phase shifts between the two waves are referenced to the horizontal wave. Therefore, the final form of the wave equation can be written as

$$\overline{E}_x(r,t) = E_H e^{j(wt-kz)}\overline{a}_x$$

$$\overline{E}_y(r,t) = E_v e^{j(wt-kz\pm\beta_o)}\overline{a}_y$$

*"Time-Harmonic Electromagnetic Fields" Roger F. Harrington, McGraw-Hill Book Co., 1961.

where

E_H is electric field strength polarized in the horizontal direction.
E_V is electric field strength polarized in the vertical direction.
ω is the radian frequency of the transmitted wave.
k is the complex wave number.*
t is time
z is distance (when z = range to target z = R)
β_o is the phase difference between horizontal and vertical electric field waves at the transmitting antenna. ($-\pi \leq \beta_o \leq \pi$)
$\underline{a}_y$ is a unit vector in the Y direction (vertical)
$\underline{a}_x$ is a unit vector in the X direction (horizontal)
a_z is a unit vector in the Z direction (range)

C. Special Cases of Polarization

1. Linear (see Figure 1)

$\beta_o = 0$

$\overline{V}_T = E_v e^{j(wt-kz)}\underline{a}_y$ or $E_v \cos(wt-kz)\underline{a}_y$

$\overline{H}_T = E_H e^{j(wt-kz)}\underline{a}_x$ or $E_H \cos(wt-k_z)\underline{a}_x$

$\rho = \arctan V_T/H_T = \arctan E_v/E_H$

*$(k=k'-jk")$ where k' is the intrinsic phase constant and k" is the intrinsic attenuation constant. When no attenuation is assumed $k = k' = 2\pi/\lambda$.

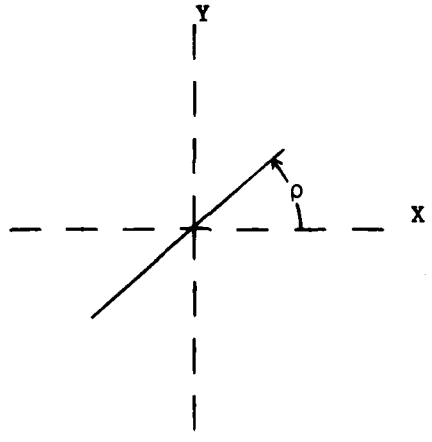


Figure 1 Linear polarization

ρ equal zero is referred to as horizontal polarization.
ρ equal ninety degrees is referred to as vertical polarization.
ρ equal forty five degrees is 45 degree linear polarization.

2. Circular (see Figures 2 and 3)

$\beta_o = \pm 90°$

$E_v = E_H = E$

Left hand circular $\beta_o = 90°$ or $\pi/2$ radians.

The loci is a circle of radius E. The electric field vector is constant in magnitude. When looking in the direction of travel the electric field vector rotates counterclockwise: when looking against the direction of travel the vector rotates clockwise.

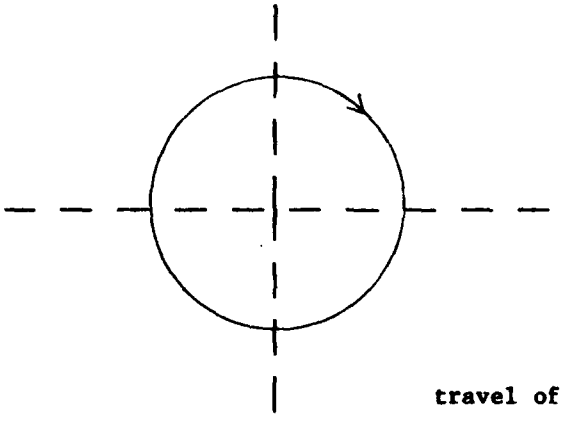


Figure 2. Left hand circular polarization.

Right hand circular

$\beta_o = -90°$ or $-\pi/2$ radians

$E_v = E_H = E$

The Loci is a circle of radius E. However, the electric field is rotating clockwise with time when viewed in the direction of travel, and counterclockwise when the observation is made looking against the direction of travel.

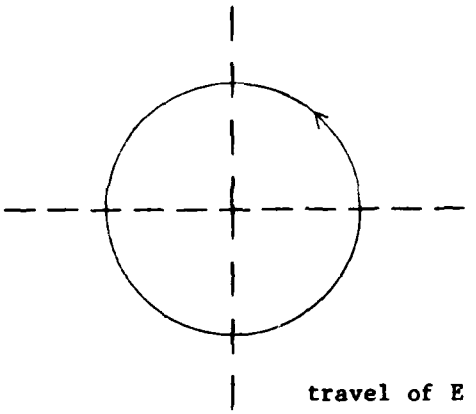


Figure 3. Right hand circular polarization

3. Elliptical (see Figure 4)

   $\sin \beta_o > 0$ left hand elliptical
   $\sin \beta_o < 0$ right hand elliptical

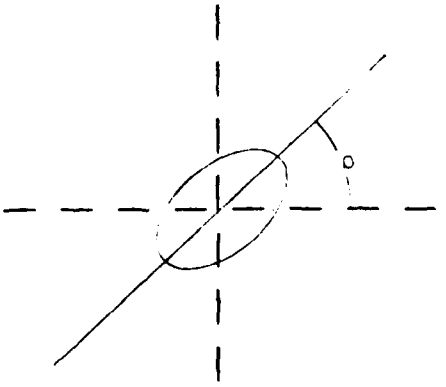


Figure 4. Elliptical polarization

The angle ρ is the angle to the major axis e and is dependent upon the ratio of E_V, E_H, and β_o.

D. Polarization Notation

As previously presented, a pure left hand circular polarized wave electric field may be shown as

$$\bar{E}_{T_L} = E[\cos(wt-kz)\bar{a}_x - \sin(wt-kz)\bar{a}_y]$$

and a pure right hand circular polarization as

$$\bar{E}_{T_R} = E[\cos(wt-kz)\bar{a}_x + \sin(wt-kz)\bar{a}_y]$$

For simplicity the time dependency given originally as e^{jwt} may be suppressed or removed and a circular wave can be represented as

$$\bar{E}_{T_L} = E[\cos(-kz)\bar{a}_x - \sin(-kz)\bar{a}_y]$$

$$\bar{E}_{T_R} = E[\cos(-kz)\bar{a}_x + \sin(-kz)\bar{a}_y]$$

Assuming the electric field at the transmitter (z=0) is 90° (plus or minus) out of phase in the H and V direction, or

$$\bar{E}_{T_L} = E\bar{a}_x + Ee^{j\pi/2}\bar{a}_y$$

$$E_{T_L} = E + jE \quad \text{for left hand circular (see Figure 5)}$$

Therefore

$$E_{T_R} = E - jE \quad \text{for right hand circular}$$

$H_T = E \cos(wt-kz)\, a_x$
$V_T = E \cos(wt-kz+90°)ay$

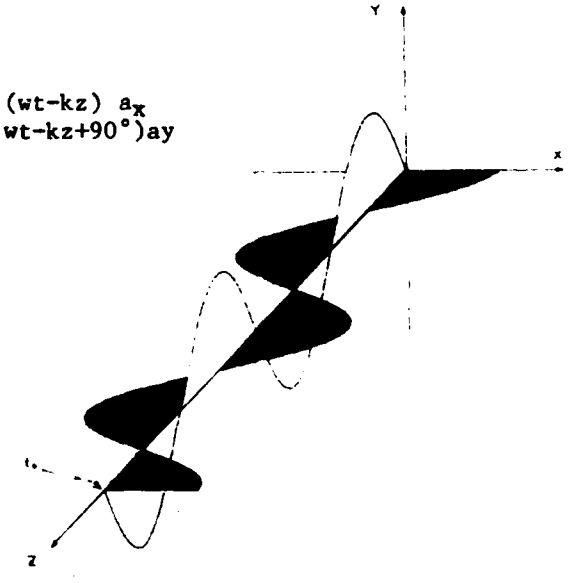


Figure 5. Left hand circular wave traveling in Z direction.

E. Scattering Matrix

Scattering of a wave by objects in the field of view is modeled by the polarization scattering matrix as

$$[\overline{E^R}] = [S][\overline{E^T}] \cdot \frac{1}{\sqrt{4\pi R^2}}$$

where $\overline{E}^R$ is received electric field vector

$\overline{E}^T$ is transmitted electric field vector

$$[S] = \begin{bmatrix} S_{HH} & S_{HV} \\ S_{VH} & S_{VV} \end{bmatrix}$$

It can be shown that the scattering matrix is related to radar cross section in the following manner:

$$[S] = \begin{bmatrix} \sqrt{\sigma_{HH}}e^{j\phi_{HH}} & \sqrt{\sigma_{HV}}e^{j\phi_{HV}} \\ \sqrt{\sigma_{VH}}e^{j\phi_{VH}} & \sqrt{\sigma_{VV}}e^{j\phi_{VV}} \end{bmatrix}$$

For convenience, the $\frac{1}{\sqrt{4\pi R^2}}$ term is usually dropped and the received electric field components are shown to be related by

$$[\overline{E^R}] = [S]\ [\overline{E^T}]$$

Therefore, for a monostatic radar the voltage at the antenna terminals is related as

$$[\overline{E^R}] = \begin{bmatrix} E_H \\ E_V \end{bmatrix} = [S] \begin{bmatrix} E_H^T \\ E_V^T \end{bmatrix} \cdot \frac{1}{K}$$

where K is some factor that represents the appropriate radar range scaling, which for most calculations is not considered unless the absolute received voltage is required.

For a left hand circular transmitted wave this becomes

$$[E^R] = \begin{bmatrix} E_H \\ E_V \end{bmatrix} = [S] \begin{bmatrix} E_e j(wt-2kR) \\ E_e j(wt-2kR+\pi/2) \end{bmatrix}$$

where R is now the one way range to the target from the radar.

In short hand notation this can be written as

$$[E^R] = [S] \begin{bmatrix} E^T \\ jE^T \end{bmatrix}$$

F. Scattering Matrix for Simple Objects

The polarization scattering matrix in its most generic form is written as

$$[S] = \begin{bmatrix} S_{11}e^{j\phi_{11}} & S_{21}e^{j\phi_{21}} \\ S_{12}e^{j\phi_{12}} & S_{22}e^{j\phi_{22}} \end{bmatrix}$$

where the subscripts refer to orthogonal components, the first subscript being receive, and the second transmit.

In the linearly polarized form this becomes

$$[S] = \begin{bmatrix} S_{HH}e^{j\phi_{HH}} & S_{HV}e^{j\phi_{HV}} \\ S_{VH}e^{j\phi_{VH}} & S_{VV}e^{j\phi_{VV}} \end{bmatrix}$$

In the circular polarized form this becomes

$$[S] = \begin{bmatrix} S_{RR}e^{j\phi_{RR}} & S_{RL}e^{j\phi_{RL}} \\ S_{LR}e^{j\phi_{LR}} & S_{LL}e^{j\phi_{LL}} \end{bmatrix}$$

where R refers to right hand circular, and L to left hand circular.

In this analysis where a circular wave (right or left) is broken into its horizontal and vertical components the linearly polarized scattering matrix must be used. However, the same results could be obtained by using the circular scattering matrix and not breaking down the electric field into orthogonal components of E_H and E_V.

Consider an odd bounce reflector (a flat plate) that totally reflects the transmitted wave. The linear scattering matrix elements can be written as

$$S_{11} = S_{HH},\ S_{12} = S_{HV},\ S_{21} = S_{VH},\ S_{22} = S_{VV}$$

The return from an impinging horizontal electrical field will have the same magnitude but the phase will shift 180°. The same is true for an impinging vertical electric field. A horizonal electic field striking a flat plate and being received in the vertical direction is zero. The same is true for transmit vertical/receive horizonal.

The scattering matrix for a flat plate is therefore

$$[S]_{FP} = \begin{bmatrix} -1 & 0 \\ 0 & -1 \end{bmatrix}$$

It should be noted that the 180° phase shift is due to the electromagnetic boundary condition of zero tangential field at the surface of a perfect conductor. This matrix could have been written as

$$[S]_{FP} = \begin{bmatrix} e^{-j\pi} & 0 \\ 0 & e^{-j\pi} \end{bmatrix}$$

noting that $e = -1$.

The same odd bounce reflector in a circlar scattering matrix would have the following elements

$$S_{11} = S_{RR};\quad S_{12} = S_{RL};\quad S_{21} = S_{LR};\quad S_{22} = S_{LL}$$

The rotation of the return from a circularly polarized wave will be the reverse of the rotation of the transmitted wave; that is, right hand transmitted becomes left hand received. Because there is only a cross polarization component, the circular scattering matrix for a flat plate (odd) is

$$[S]_{FP} = \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix}$$

The equivalence of linear polarization and circular polarization can best be seen from examples of both worked in parallel. Assume a left hand circular transmit into a flat plate.

The circular (left hand) transmit is written in a linear system as

$$E_L^T = E_{\bar{a}_x} + jE_{\bar{a}_y}$$

A circular transmit system is written in circular notation as

$$E^T_{\bar{a}r} = E^T_{\bar{R}} + E^T_{\bar{L}}$$

where $\bar{R}$ is a unit vector rotating in the right hand direction

$\bar{L}$ is a unit vector rotating in the left hand direction

LINEAR (FLAT PLATE)	CIRCULAR (FLAT PLATE)
$\begin{bmatrix} E_H^R \\ E_V^R \end{bmatrix} = \begin{bmatrix} -1 & 0 \\ 0 & -1 \end{bmatrix} \begin{bmatrix} E_H^T \\ jE_V^T \end{bmatrix}$	$\begin{bmatrix} E_R^R \\ E_L^R \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix} \begin{bmatrix} 0 \\ E_L^T \end{bmatrix}$
$E_H^R = -E_H^T$	$E_R^R = E_L^T$
$E_V^R = -jE_V^T$	$E_L^R = 0$
Noting that the direction of travel has reversed the received wave is of the form	In the circular form the received wave is of the form
$E^R = -E_H^T - jE_V^T$	$E = E_L^T\overline{R} + 0$
which is a right hand circular wave traveling in the -z direction.	which is a right hand circular wave.

Figure 6. Scattering characteristics.

The analytical relationships developed for the simulation are based upon the linear scattering characteristics of a few simple shapes, classified to some degree by the number of reflecting surfaces encountered.

1. Odd bounce scattering matrix (flat plate, trihedral corner reflector) for linear polarization (see Figure 6)

$$\begin{bmatrix} e^{-j\pi} & 0 \\ 0 & e^{-j\pi} \end{bmatrix} \quad \text{or} \quad \begin{bmatrix} -1 & 0 \\ 0 & -1 \end{bmatrix}$$

2. Even bounce scattering matrix (diplane) for linear polarization

$$\begin{bmatrix} \cos 2\theta & \sin 2\theta \\ \sin 2\theta & -\cos 2\theta \end{bmatrix}$$

where θ is the angle of rotation of the diplane relative to the horizontal. (See Figure 7).

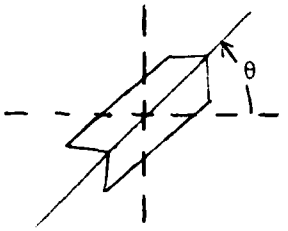


Figure 7. Diplane rotation angle.

3. Dipole matrix for linear polarization (Figure 8).

$$\begin{bmatrix} -\cos 2\theta & -\cos\theta\sin\theta \\ -\cos\theta\sin\theta & -\sin 2\theta \end{bmatrix}$$

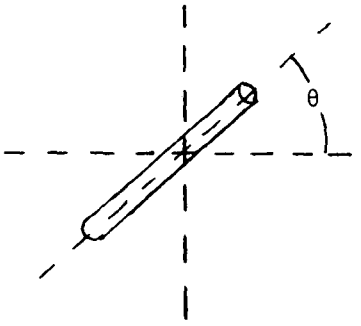


Figure 8. Dipole rotation angle.

It is assumed that superposition holds such that a complex target may be modeled by an ensemble of these even and odd bounce targets or scatterers with the inclusion of their respective ranges.

G. Radar Range Scaling

The basic radar equation to determine the power received at the radar is

$$P_r = \frac{P_t G^2 \lambda^2 \sigma}{(4\pi)^3 R^4 L_s}$$

where P_r is power received in watts
P_t is peak power transmitted in watts
G is antenna gain (unitless)
λ is wavelength in meters
σ is radar cross-section in meters squared
R is range to target in meters
L_s is system loss (unitless)

Because this analysis is performed in the voltage domain the standard radar equation must be modified to be expressed in terms of voltage.

$$P_r = (V_{peak}/\sqrt{2})^2/Z$$

where Vpeak is peak voltage received
Z is impedance (assumed 50 ohms)

Therefore, the peak voltage (Vpeak) is

$$V_{peak} = \sqrt{2*Z*P_r}$$

By removing the radar cross-section from P_r, P_r becomes a constant radar scalar which when used with the field voltage obtained from the scattering matrix defines the peak received voltage.

$$V_{peak} = \sqrt{\frac{2ZP_tG^2\lambda^2}{(4\pi)^3R^4L_s}} \cdot \sqrt{\sigma}$$

H. Noise Generation

If the radar were operated in a perfectly noise free environment so that no external noise sources accompanied the desired signal, there would still exist an unavoidable component of noise generated by the thermal motion of the conduction electrons in the receiver input stages. This is called thermal noise and is directly proportional to the temperature of the ohmic portions of the circuit, and the receiver bandwidth. The available thermal-noise power generated by a receiver of bandwidth B_n (in Hz) at temperature T (degrees Kelvin) is equal to:

$$\text{average available power} = KTB_n$$

where K is Boltzmann's constant (1.38×10^{-23} joule/deg)

No matter whether the noise is generated by a thermal mechanism or by some other mechanism, the total noise at the output of the receiver may be considered to be equal to the thermal-noise power obtained from an ideal receiver multiplied by a factor called noise figure (NF). The noise figure (NF) of a receiver is defined by the equation:

$$NF = \frac{\text{noise out of practical receiver}}{\text{noise out of ideal receiver at Std Temp } (T_o)}$$

The standard temperature is taken to be 290° K.

Therefore,

$$\text{average available power} = KT_oBNF$$

Assuming this to be the available average power at the input stages of a radar, the ohmic load is assumed to be matched as in the simple circuit diagram in Figure 9.

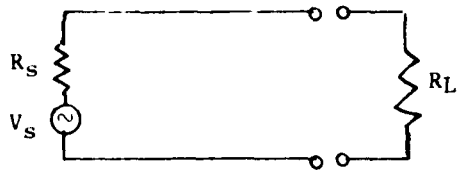


Figure 9. Equivalent noise circuit.

NOTE: Load resistance (R_L) is matched to source resistance (Rs).

Therefore, the RMS voltage available at the source can be calculated as

$$V_s = 2V_L$$

$$V_L = \sqrt{KT_oBNFR_L} \text{ and } \quad V_s = 2\sqrt{KT_oBNFR_L}$$

The noise entering the IF amplifier is assumed to be Gaussian, with a probability-density function given by

$$p(v)dv = \frac{1}{\sqrt{2\pi\Psi}} \exp\left(\frac{-v^2}{2\Psi_o}\right) dv$$

where p(v) dv is the probability of finding the noise voltage between the value of v and v + dv, Ψ_o is the variance, or mean-square value of the noise voltage. The mean value of v is taken to be zero.

Therefore, the mean-square value is taken to be V_L^2 or KT_oBNFR_L and the standard deviation by definition is

$$SD = \sqrt{KT_oBN_FR_L}$$

I. Antenna Isolaton

When two antennas (or elements) are widely separated the energy coupled between them is small, and the influence of the receiving antenna on the current excitation and pattern of the transmitting antenna is negligible. As the antennas (or elements) are brought closer together the coupling between them increases.

Isolation of a polarimetric antenna is represented as two antennas that cross couple energy during transmit and receive. Considering only the transmit cycle the coupling can be represented as in Figure 10.

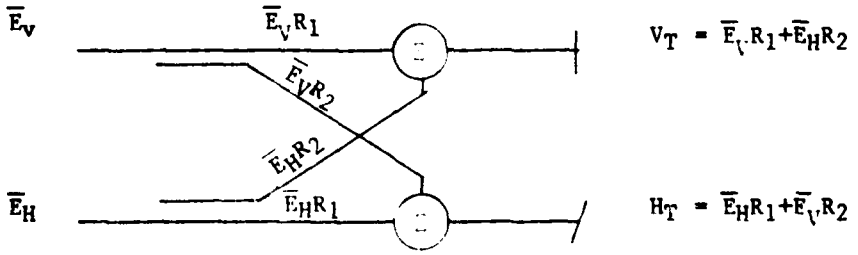


Figure 10. Transmit isolation.

In Figure 10 $\overline{E}_V$ and $\overline{E}_H$ are input to the antenna, and $\overline{V}_T$ and $\overline{H}_T$ are antenna outputs.

$$R_2 = 10^{-ISOL/20} \qquad R_1\sqrt{1-10^{ISOL/10}}$$

ISOL is the antenna isoluation in dB one way, always positive.

Assuming reciprocity, the isolation upon receive is

$$\overline{E}_{VR} = \overline{V}_R R_1 + \overline{H}_R R_2$$

$$\overline{E}_{HR} = \overline{H}_R R_1 + \overline{V}_R R_2$$

where $\overline{E}_{VR}$ and $\overline{E}_{HR}$ are the input signals to the receiver and $\overline{H}_R$ and $\overline{V}_R$ are the inputs at the antenna plane.

Phase stability is assumed across the antenna plane.

J. Frequency Agility and Intra-Range Resolution

Range resolution is usually defined as the distance at which two targets can be resolved in range. In the conventional radar this is defined by the pulse width of the transmitted wave as $\Delta R = (C\tau/2)$

where ΔR is range resolution (m)
C is velocity of light (m/sec)
τ is radar pulsewidth (sec)

Considering the radar to have a match receiver τ, equal to one over receiver bandwidth, ΔR becomes

$$\Delta R = \frac{C}{2B}$$

where B is bandwidth in Hertz.

Either of the two equations can be used to calculate the range resolution of a radar. However, the latter equation is the more general form and can be ulitized in calculating range resolution in conventional radar, pulse compression radar, and frequency agile radar, as well as in hybrids of these such as the pulse compression frequency agile radar.

Ruttenberg showed in 1967 a method that increased range resolution with a non-coherent source. This involved a frequency agile scheme that summed the pulses after they were received (coherent on receive) and delayed by 1/PRF. Since then the use of a fully coherent radar utilizing frequency agility, pulse to pulse, and the Digital Fourier Transform, has demonstrated a range resolution technique that does not require delay lines as did Ruttenberg's technique. The coherent pulses are fed to a DFT (usually the same size as the number of frequency shifts) and frequency is transformed into time (via the DFT) with intra-range resolution of the system following the same range resolution equation.

$$\Delta R = \frac{C}{2B}$$

where B is now the frequency agile bandwidth.

Gjessing, in his book "Adaptive Radar in Remote Sensing" shows that the amplitude spectrum of the scattered field is the Fourier transform of the delay function f(t). Thus, if the target is at some distance d, the delay function will oscillate with a period c/2d. Therefore, by the use of a multifrequency radar system, the resolution of the radar can be increased as the bandwidth of the agile radar increases. The complex Fourier transform will provide the true reference, while the amplitude only Fourier transform will provide the relative distances between the resolvable elements.

III. RADAR SIMULATION

The functional diagram of a polarimetric radar is shown in Figure 11. Functionally the model is a frequency agile coherent radar model. If a non-coherent radar model is desired the signal processor section can be modified. The frequency agile waveform selects the transmitter and coherent local oscillator frequency. The transmitter energy is split (coupled) to the dual polarized antenna with a $\pm 90^{o}$ phase shifter in the vertical channel, resulting in either right hand or left hand circular polarization. Zeroing one channel or the other results in horizontal or vertical polarization. Removing the phase shifter and adjusting the splitter results in slant polarization of any desired angle. If the cross coupling in the antenna section is large enough the result is an elliptical wave.

Energy reflected from the target area is received in both the horizontal and vertical antennas with cross coupling, and passed to the coherent Inphase (I) and Quadrature (Q) detectors, resulting in IF detected signals of horizontal I and Q, and vertical I and Q.

This radar simulation configuration allows maximum versatility by providing for circular, elliptical, and linear polarization transmission. Receiving horizontal and vertical with antenna cross coupling allows the signals of circular and elliptical, and horizonal and vertical, either coherent or non-coherent. The configuration shown in Figure 11 is not intended to imply prefered hardware configuration, but rather to depict a radar simulator which can be used to simulate a large number of pulsed polarimetric radars in order to evaluate proposed radars and signal processors.

A. Signal Processing

1. General

Outputs from the radar simulation (HI, HQ and VI, VQ) are input to the signal processing software where they are combined to form various types of received signals and the respective inverse Fourier transforms. The radar signals available from the the signal processor as plots are:

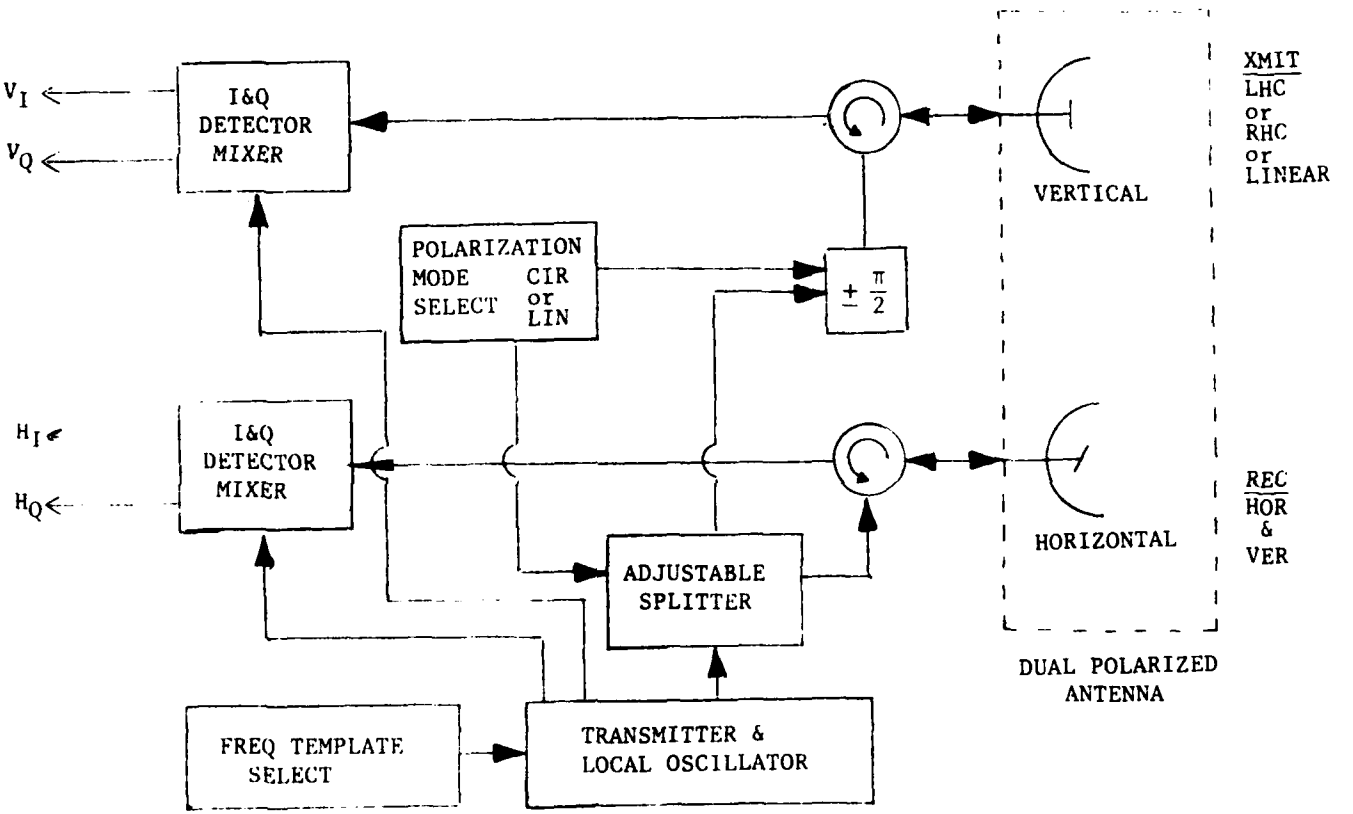
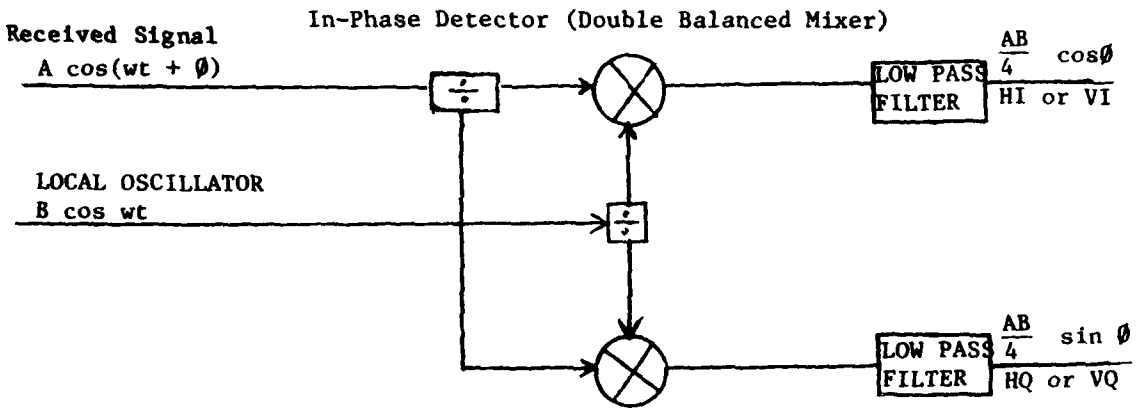


Figure 11. Radar configuration.

a. Peak Horizontal voltage
b. Peak Vertical voltage
c. Phase between Horizontal and Vertical
d. Peak Left Hand Circular (LHC) voltage
e. Peak Right Hand Circular (RHC) voltage
f. Phase between LHC and RHC
g. Scatterer Locations (inverse FFT's of coherent signals)
h. Scatterer Seperation (inverse FFT's of non coherent signals)
i. Phase plots of FFT's

2. Linear Polarizaton

Functionally the coherent horizontal and vertical signals are processed as shown in Figure 12. The resulting inphase and quadrature signals are then loaded into a complex array, and an inverse FFT is performed. The resulting lines represent the location of the scatterers relative to the leading edge of the radar range gate.



Horizontal Received Signal = HI+jHQ
Vertical Received Signal = VI+jVQ

Figure 12. Linear coherent detection block diagram.

Loading the real portion of the complex array with the amplitude magnitudes only, and performing an inverse FFT, results in lines that represent the separation of scatterers relative to each other. There is one line (neglecting sidelobes) for each combination of pairs of scatterers.

$$\text{No. lines} = \sum_{i=1}^{N} (N-i)$$

where N is the number of reflectors.

3. Circular Polarization

Figure 13 is a functional block diagram of a linear to circular tranformation. Inputting horizontal and vertical inphase and quadrature results in left and right hand circular polarized signals. This can be represented in matrix notation as:

$$\begin{bmatrix} RHc \\ LHc \end{bmatrix} = \frac{1}{\sqrt{2}} \begin{bmatrix} 1-j \\ 1\ \ j \end{bmatrix} \begin{bmatrix} H \\ V \end{bmatrix}$$

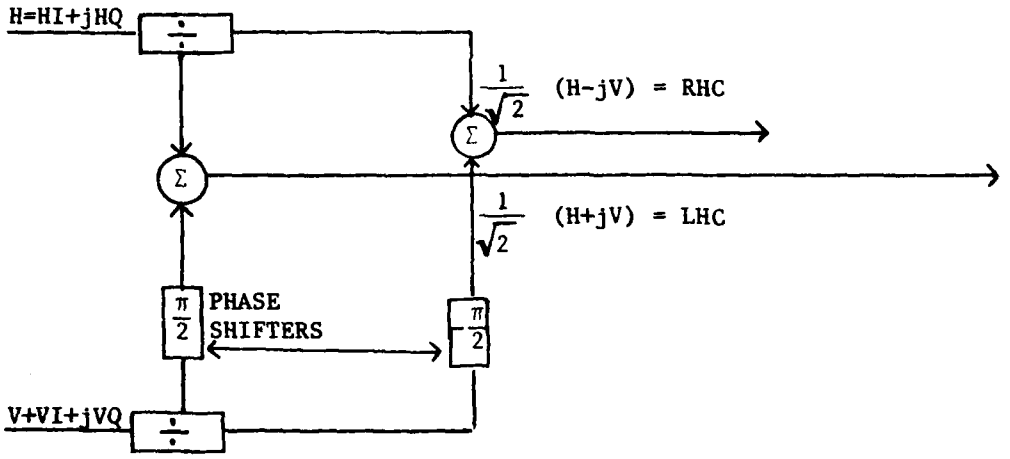


Figure 13. Block diagram of linear to circular polarization converter.

Loading the resulting RHC or LHC into a complex array, and performing an inverse FFT, results in lines that represent the location of scatterers relative to the leading edge of the radar range gate.

Loading the real portion of the complex array with the amplitude magnitudes only, and performing an inverse FFT, results in lines that represent the seperation of scatterers relative to each other.

IV. SIMULATION ULITIZATION

A. General Simulation Outputs

Figures 14 through 35 are the 22 output plots of the simulation. Four received amplitudes (Horizontal, Vertical, LHC and RHC) are plotted as a function of the transmit frequency. The header data presented at the top of each plot define the radar operating parameters used in the calculations. All plot headers for each input parameter set are identical. The phase angle between received Horizontal and Vertical or RHC and LHC is plotted as a function of transmit frequency and is the angle Beta discussed in paragraph 2.2. The remaining 16 plots are inverse FFT's, amplitude and phase, plotted as a function of intra-range gate resolution. The FFT amplitude plots labeled I&Q provide scatterer location from the leading edge of the range cell while the plots labeled peak amplitude provide scatterer separation depending on whether the amplitude data were loaded into the FFT as complex I&Q or as real amplitude only. FFT's were loaded in ascending order with the received signal from the lowest transmit frequency in location one.

B. Antenna Isolation

Utilizing the simulator program, and varying the amount of one way antenna polarization isolation, can reveal the effects of isolation on the polarimetric outputs of a system. This is exampled by Figures 28 through 32 which show the LHC and RHC scatterer locations for a four target array with 30 dB one way polarization isolation. Figures 36 and 37 present the same conditions with 10 dB isolation for comparison. Comparing these plots one can observe the cross coupling from one channel to the other.

While this example is presented for LHC and RHC output it is obvious that the other outputs of the simulation may also be examined. Antenna isolation effects on horizontal or vertical outputs, in either the coherent or non-coherent mode, as well as other combinations of transmitter and receiver polarization configurations, can be examined.

C. Signal to Noise Ratio

System noise effects on polarimetric outputs can be examined in two ways: first, by increasing target range (reducing signal strength); second, by increasing the receiver noise figure (increase system noise). Examples of these are given in Figures 17 and 21 (horizontal and vertical scatterer locations for a greater than 30 dB signal to noise ratio), and Figures 38 and 39 (for a signal to noise ratio of 8 dB).

Inputting a clutter model and varying the system noise will allow examination of the effects of the clutter to noise problem on signal processing. System noise can be increased by elevating the receiver noise figure.

D. Signal to Clutter

The utilization of the simulator program to explore the effects of clutter on signal detection will be highly dependent upon the target and clutter models used. A model for clutter in polarimetric form that has been truly verified has yet to be developed. Therefore, in order to demonstrate the use of the program the following example will be used: a contrived target model of one and one half meters radar length, made up of five reflectors randomly spaced, and having a radar cross section of five square meters each (Figures 40 through 43); clutter made up of fifty randomly selected location, orientation, and type spaced reflectors of 0.1 square meters each. This example has a total signal to clutter ratio of 25/5, or 7 dB, and is shown in the horizontal and vertical location plots in Figures 44 and 47. Figures 48 through 51 show the same configuration with the clutter cross section increased to one square meter per reflector. This represents a signal to clutter ratio of -3.0 dB (25/50).

V. CONCLUSIONS AND RECOMMENDATIONS

A digital simulation has been developed to investigate various aspects of a frequency agile, polarimetric pulsed radar system. While the simulation is not all inclusive and will undoubtedly be refined and updated for years to come, it is a useful tool for evaluating both hardware and software effects on

the next generation of radars. The simulation validation was performed by comparison with an operational radar. The validation has been excluded as the data were acquired from a contractor's IR&D radar. Any government agency desiring more information relative to the validation should contact the authors at AV 746-4061.

Major limiting factors to simulation results are the target and clutter scattering models which remain basically undefined at this time. It is recommended that all models and data in the future be done in scattering matrix format so that the entire radar signature will be available for future radar hardware and simulator designers. Without such data and validated models the radar system analyst and designer will continue to suffer from the so called "Sedenquist Effect"; that is, put two radar engineers in a room and say the word "clutter"; return years later and they will still be trying to define and agree as to what clutter is.

Future plans for the polarimetric radar simulator include the addition of doppler, tracking errors, jamming, attenuation back scatter due to weather, and realistic model development.

This simulation does not include cross range positional effects of scatterers. All cross range scatterer positions within the antenna beamwidth are collapsed to a single radial range bin. The inclusion of doppler will provide the second "cross-range" dimension for two dimensional analysis.

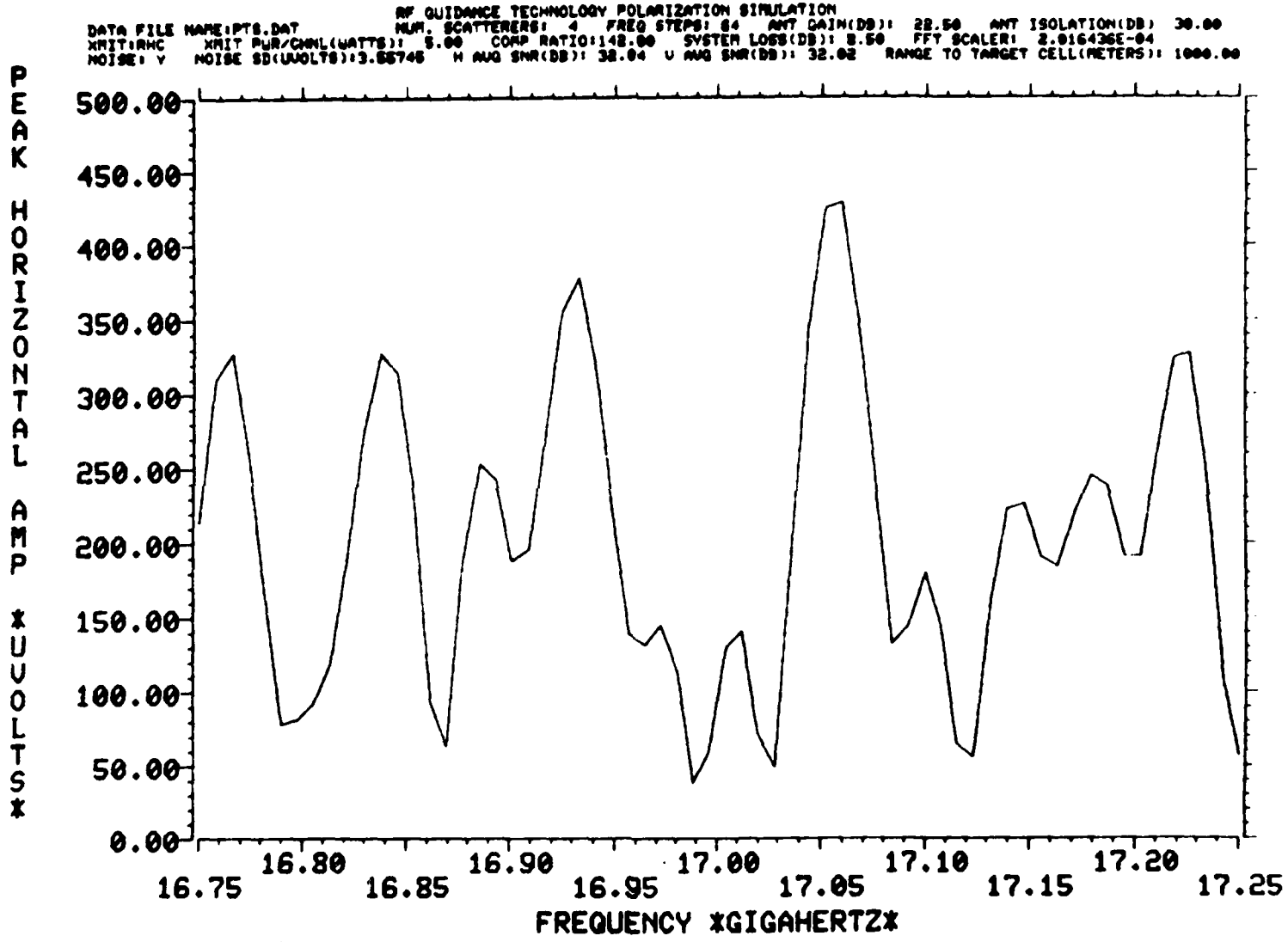


Figure 14. Peak horizontal amplitude vs. frequency at 30 dB antenna isolation.

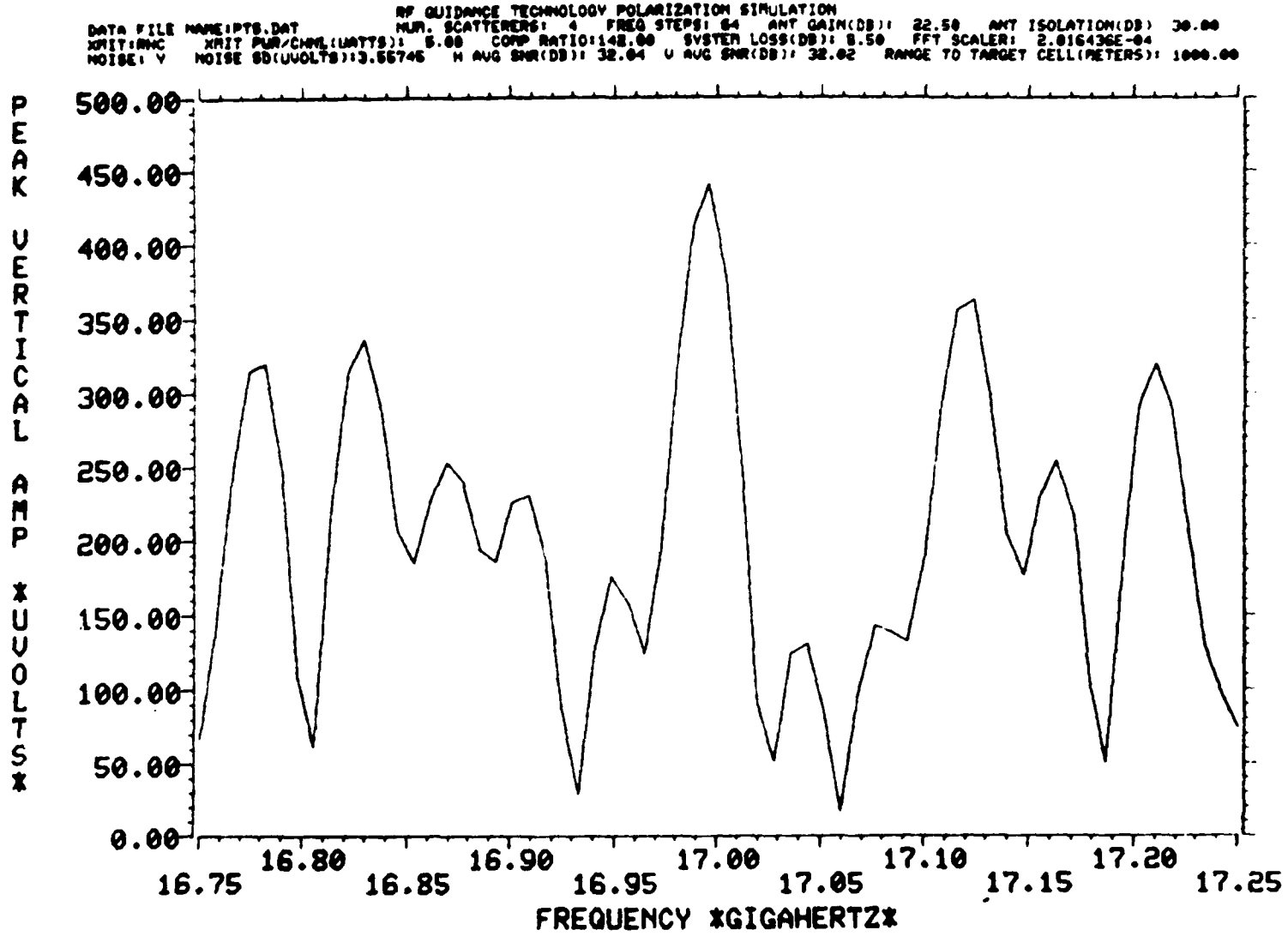


Figure 15. Peak vertical amplitude vs. frequency at 30 dB antenna isolation.

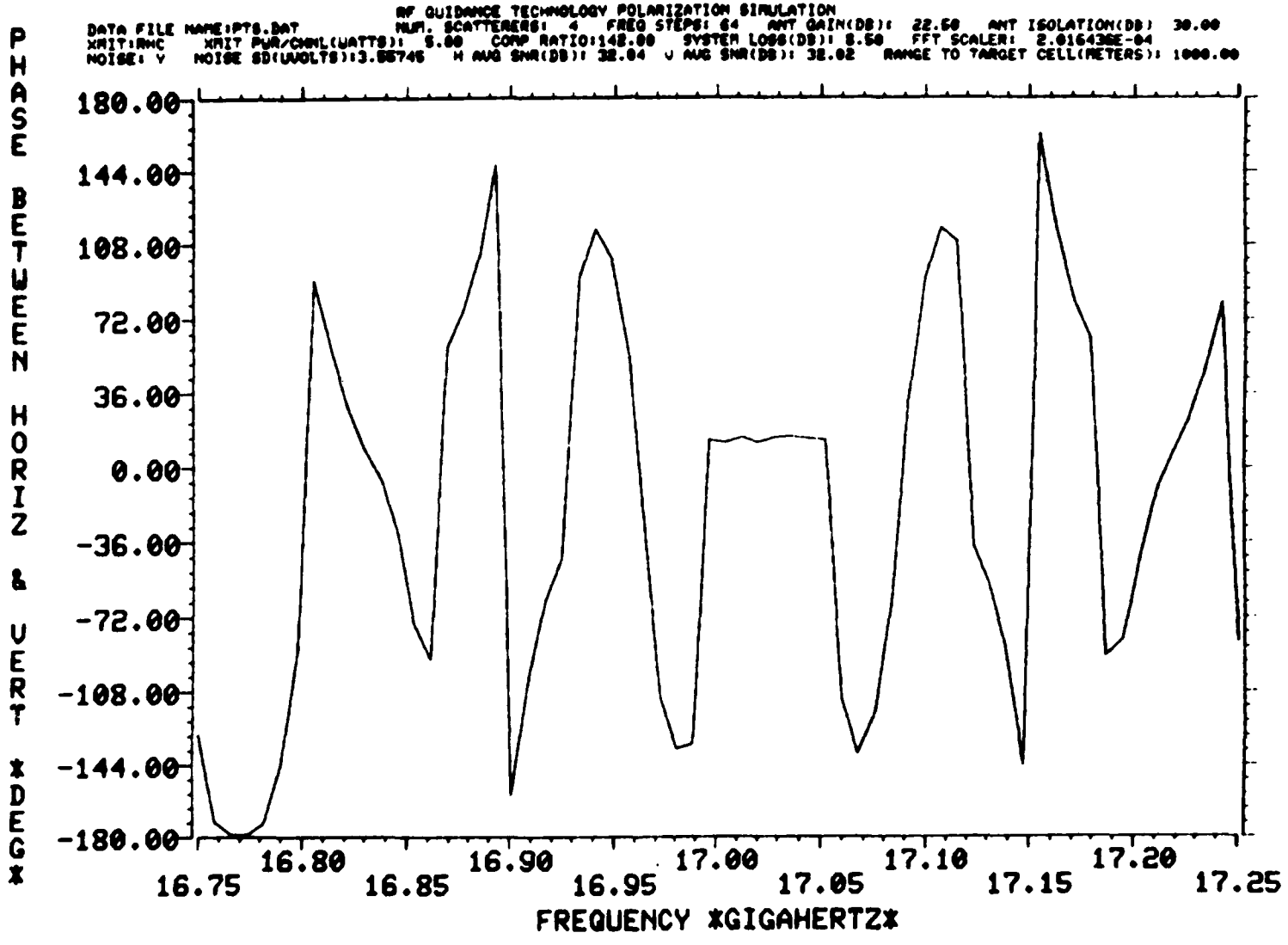


Figure 16. Phase angle between horizontal and vertical at 30 dB antenna isolation.

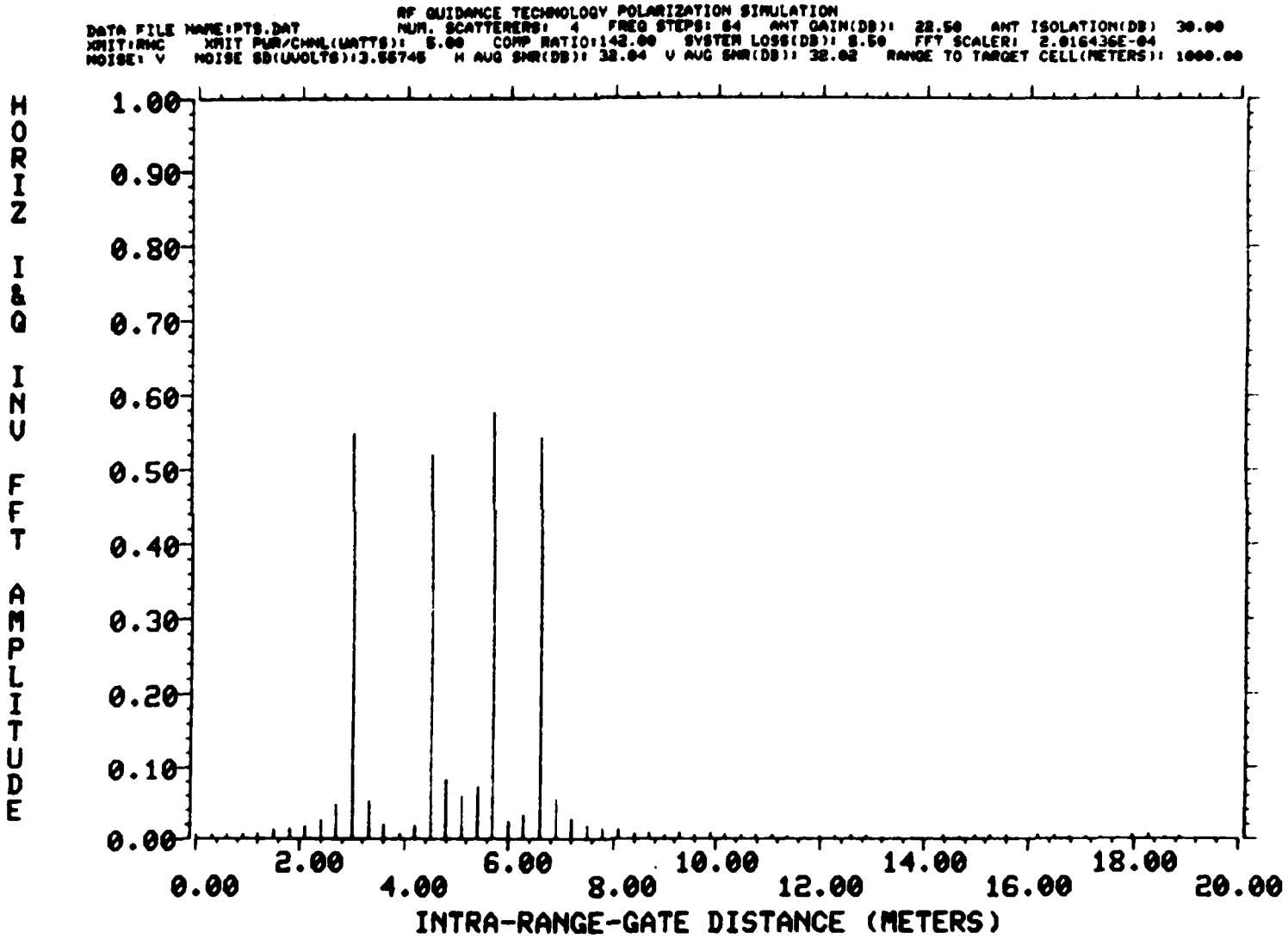


Figure 17. Inverse FFT of horizontal I&Q at 30 dB antenna isolation.

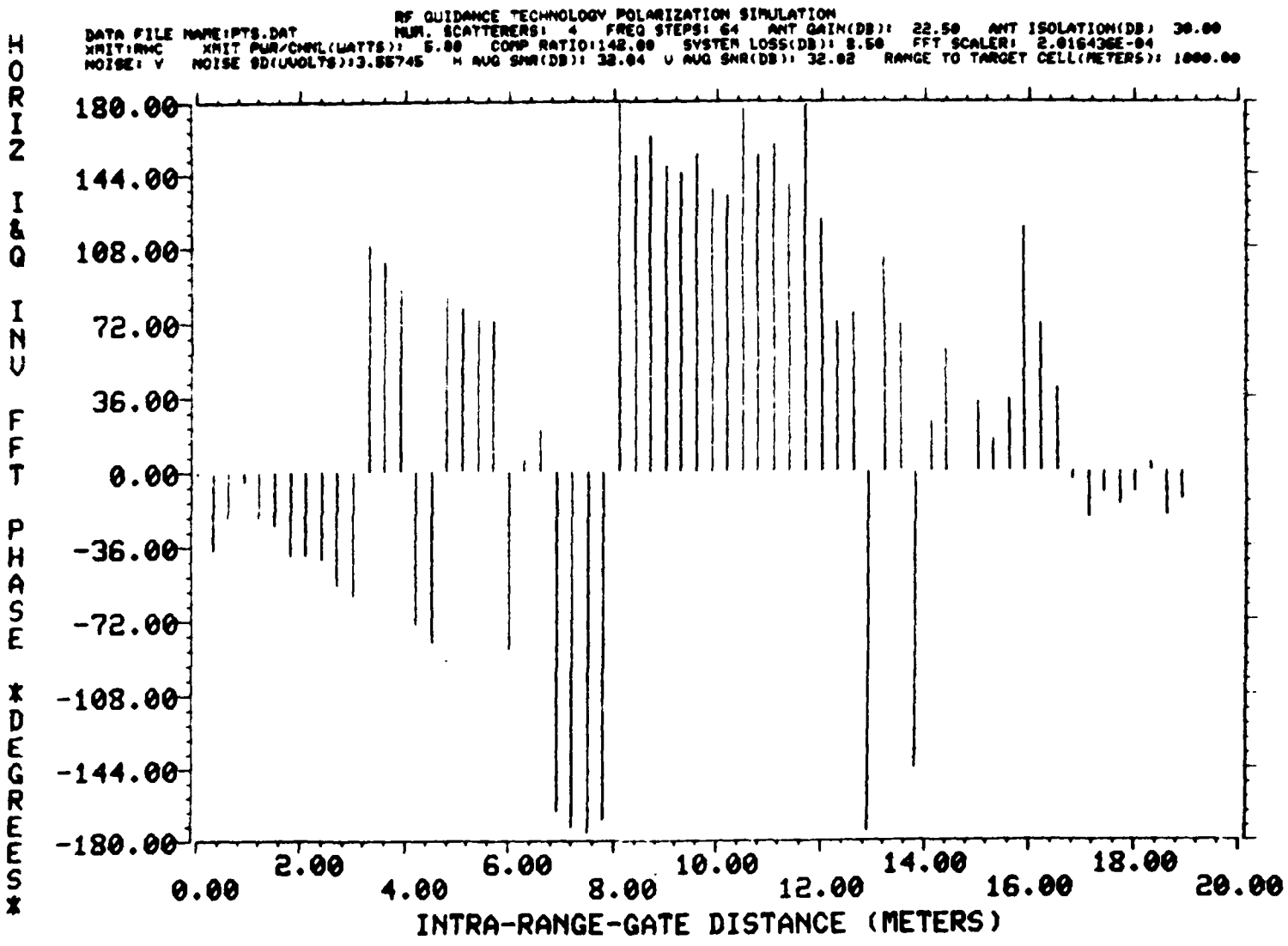


Figure 18. Inverse FFT phase angle for horizontal I&Q at 30 dB antenna isolation.

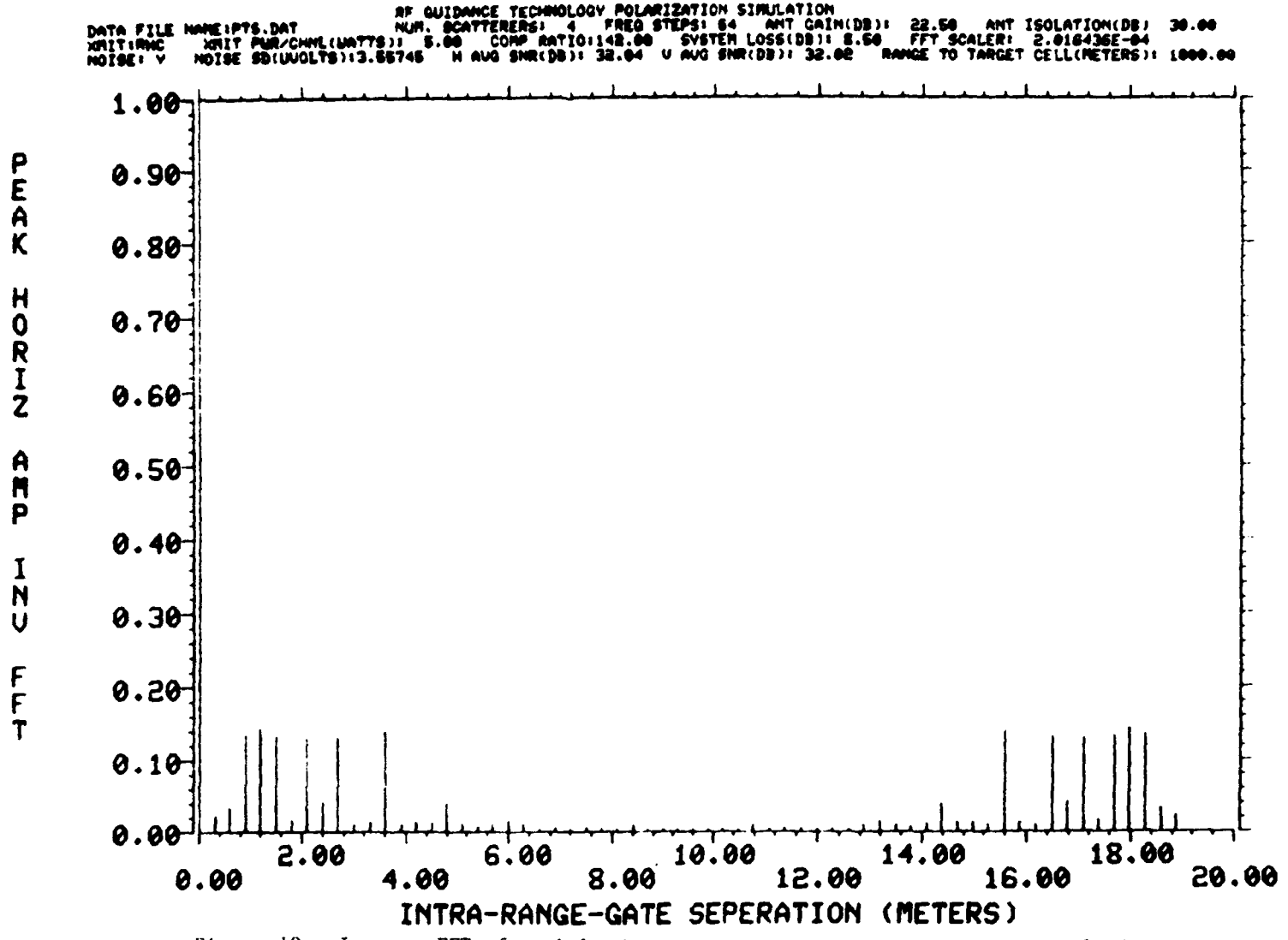


Figure 19. Inverse FFT of peak horizontal amplitude at 30 dB antenna isolation.

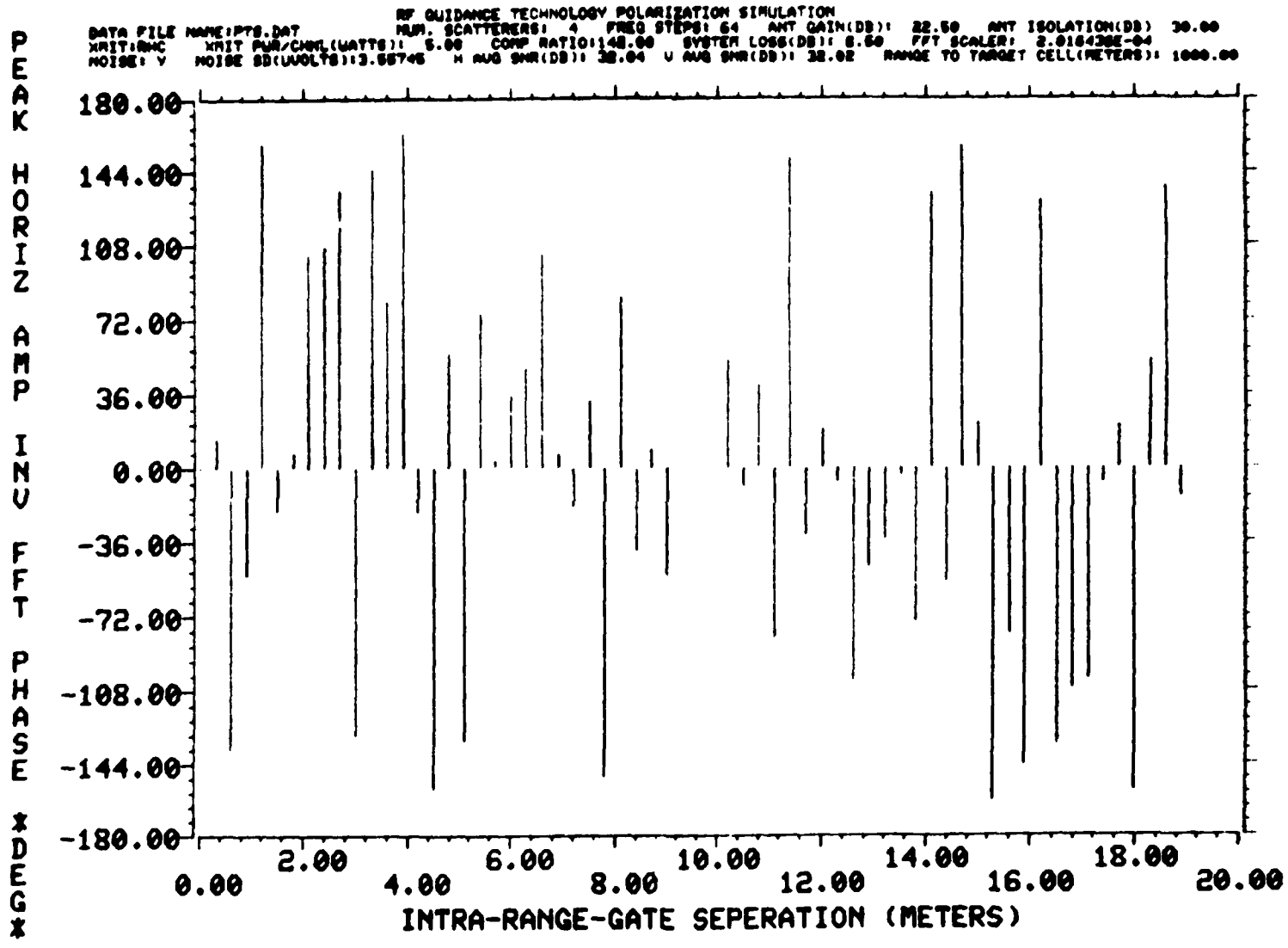


Figure 20. Inverse FFT phase angle of peak horizontal amplitude at 30 dB antenna isolation.

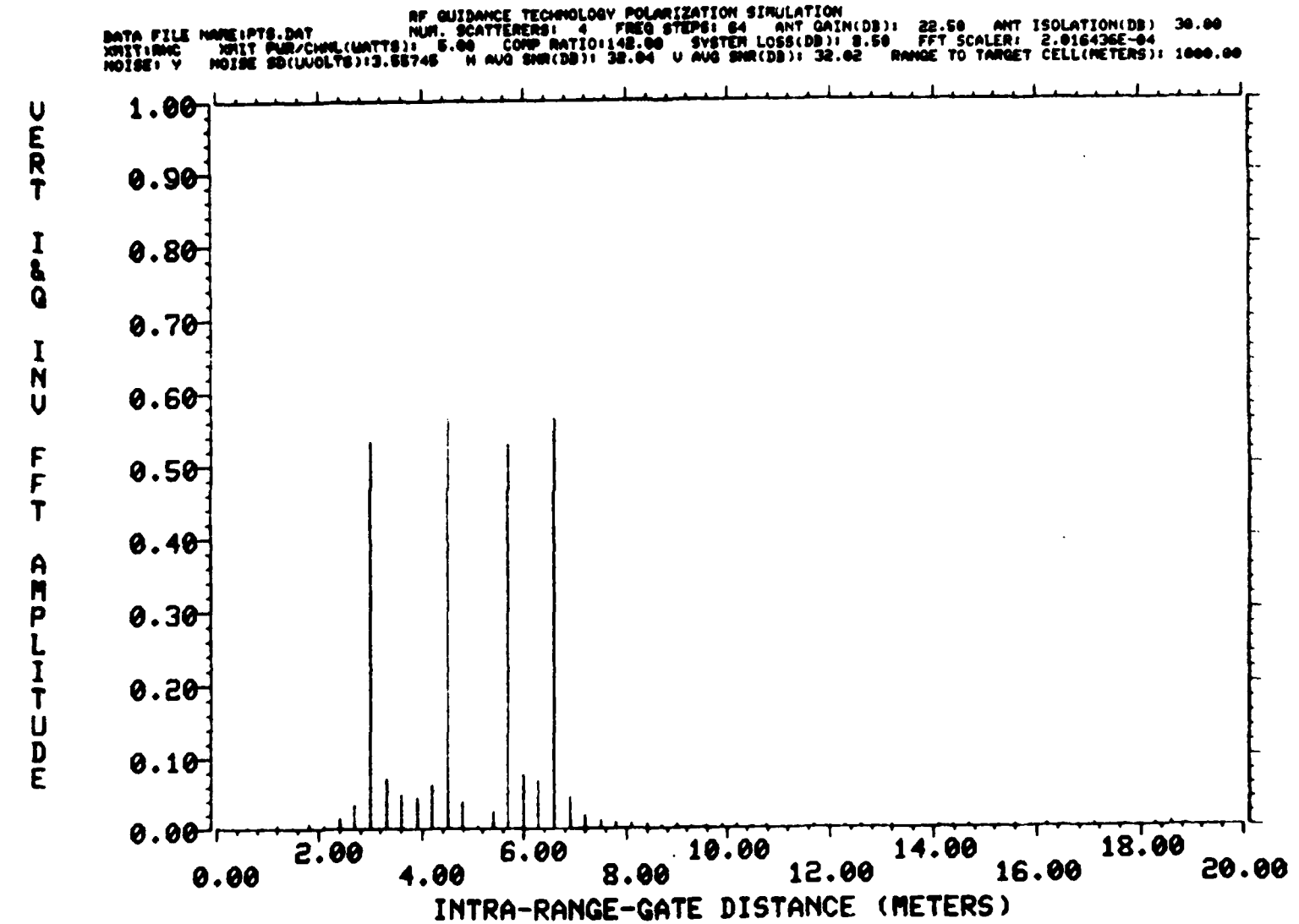


Figure 21. Inverse FFT of vertical I&Q at 30 dB antenna isolation.

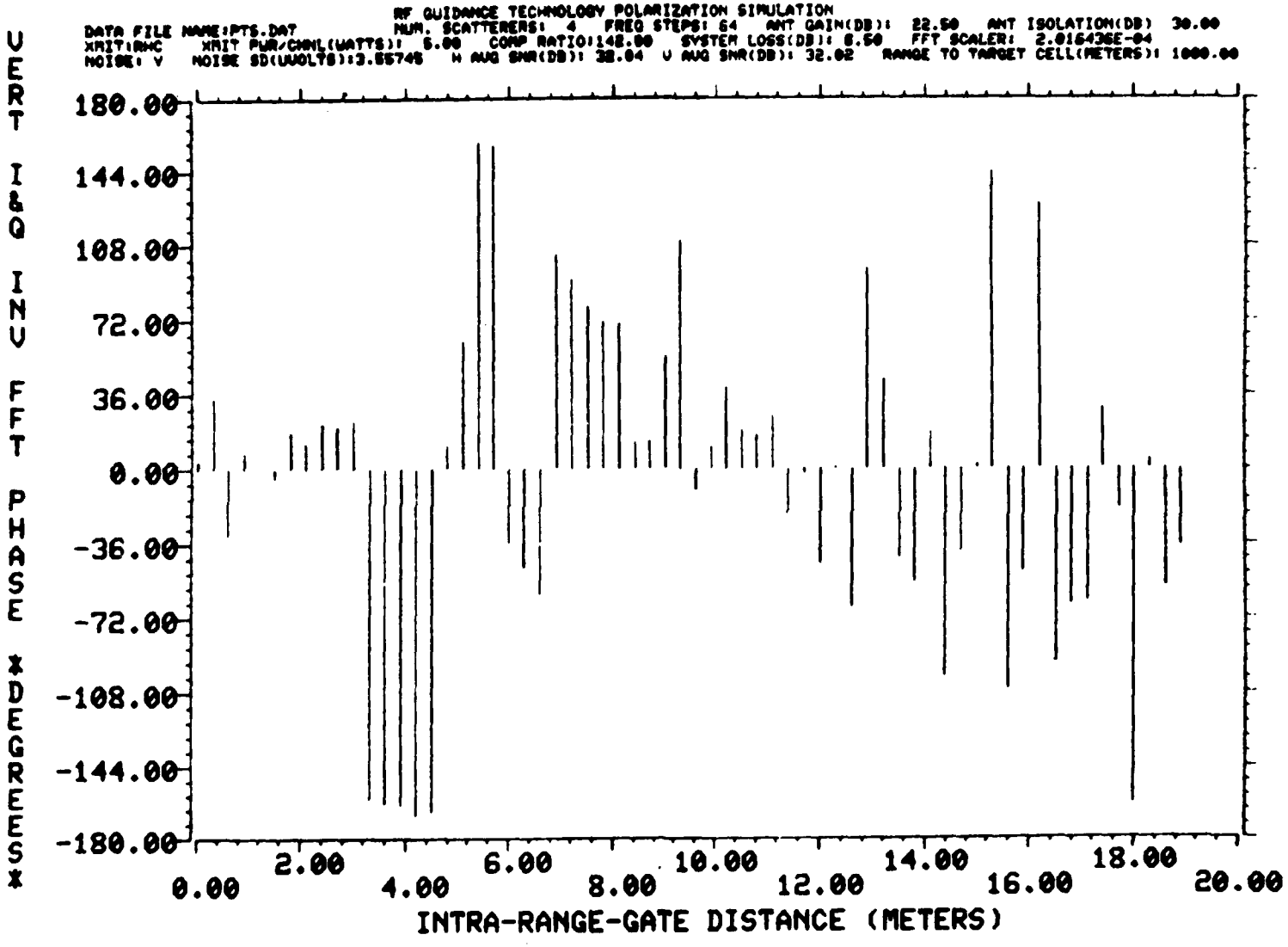


Figure 22. Inverse FFT phase angle for vertical I&Q at 30 dB antenna isolation.

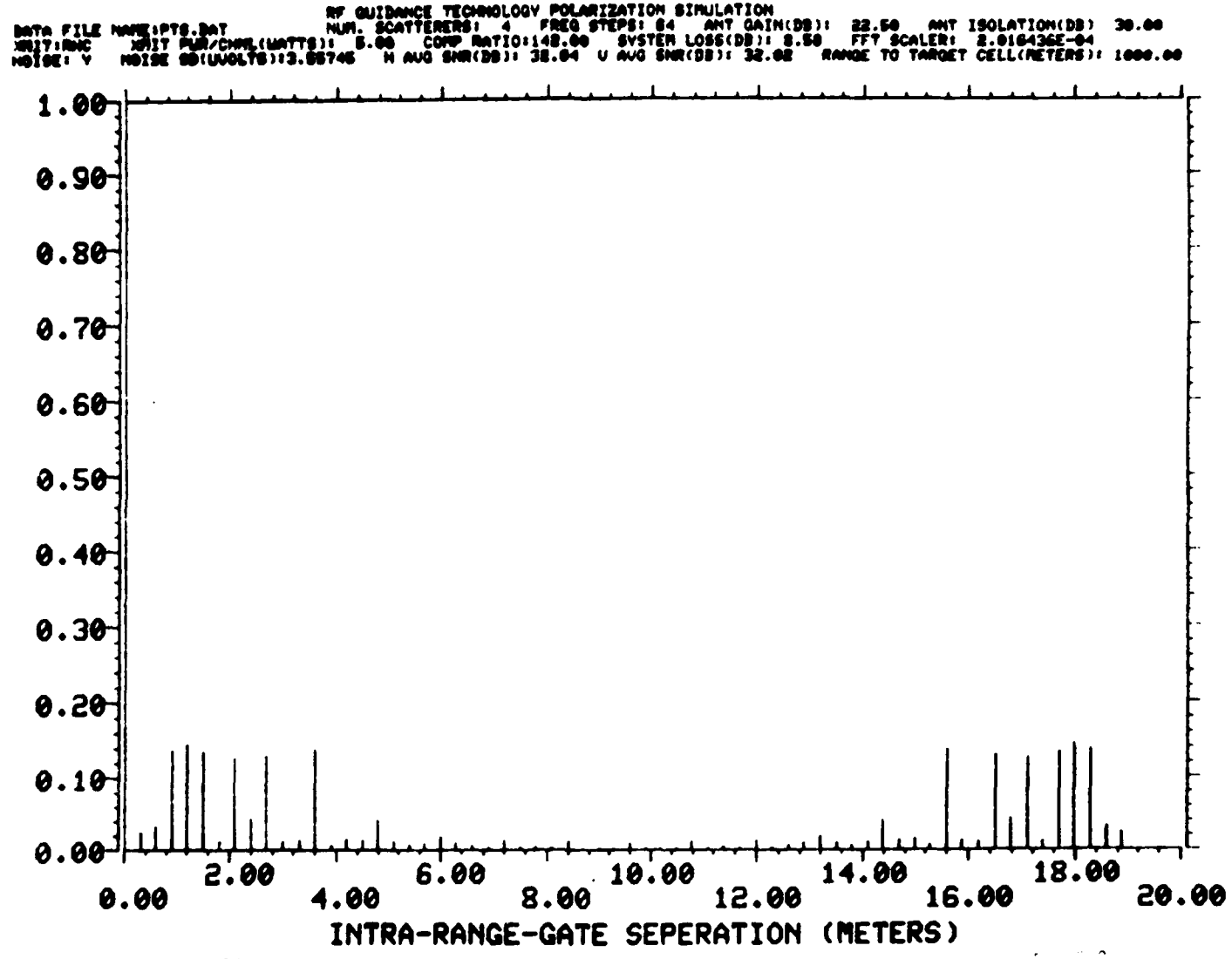


Figure 23. Inverse FFT of peak vertical amplitude at 30 dB antenna isolation.

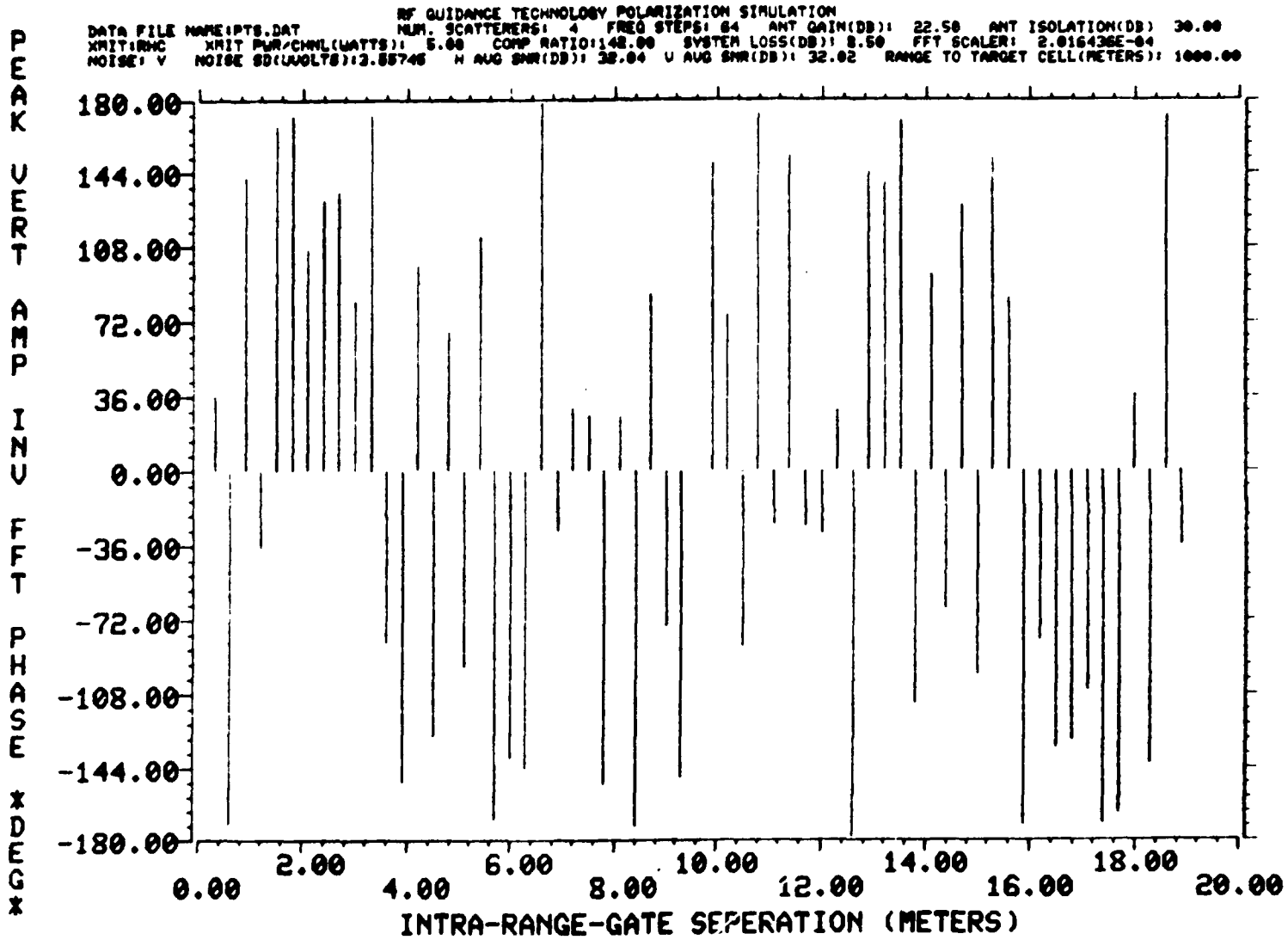


Figure 24. Inverse FFT phase angle of peak vertical amplitude at 30 dB antenna isolation.

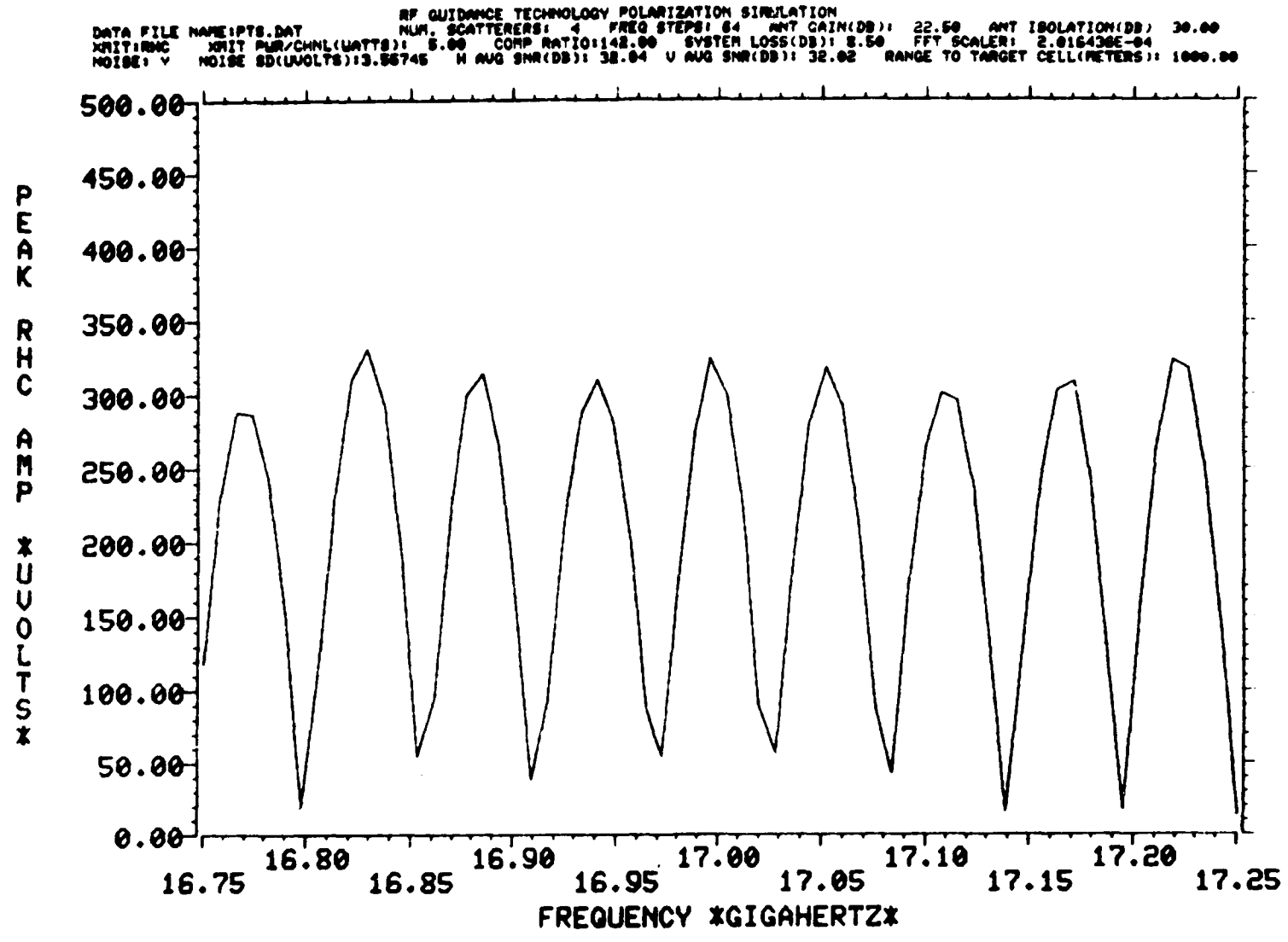


Figure 25. Peak RHC amplitude vs. frequency at 30 dB antenna isolation.

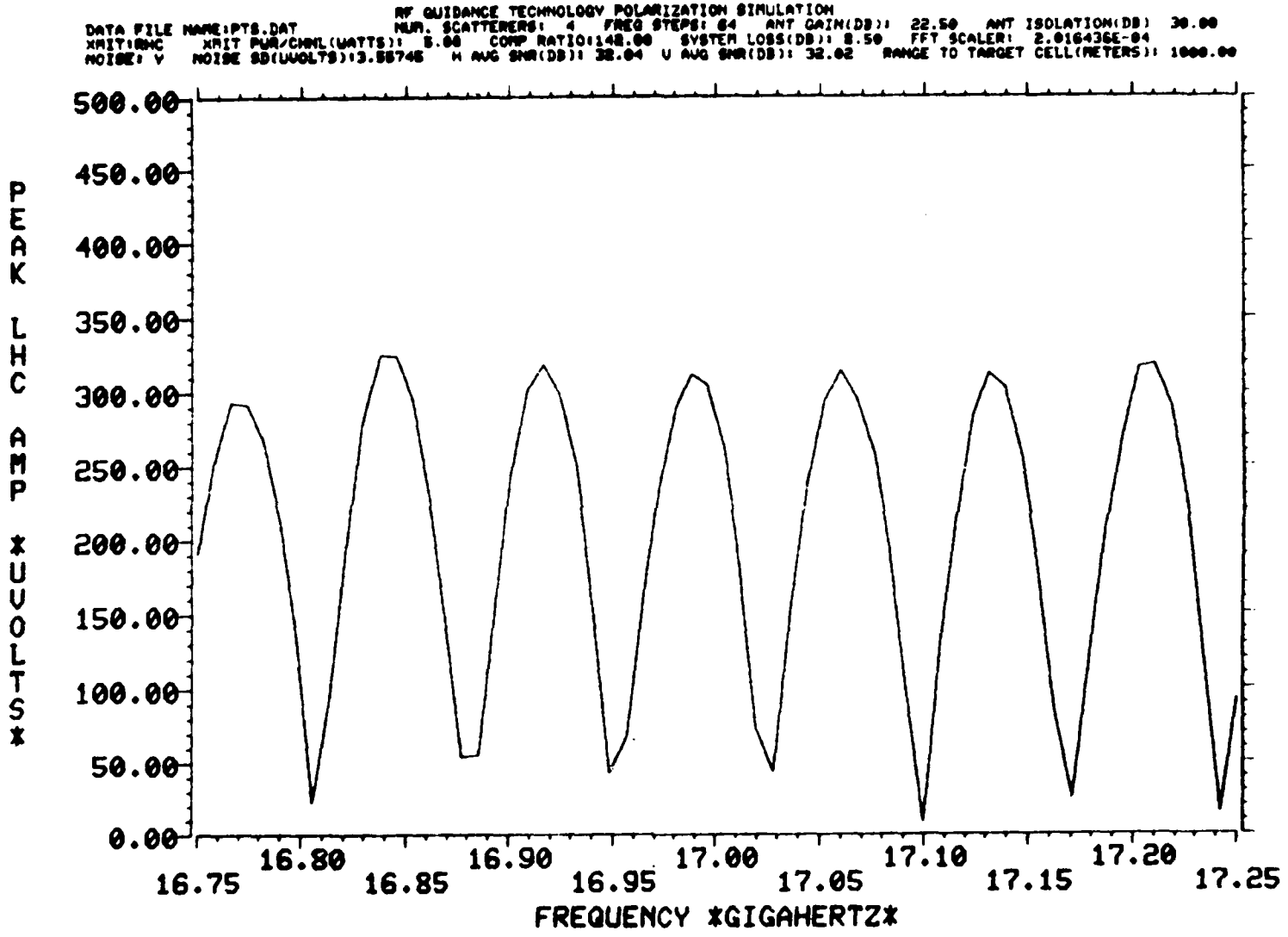


Figure 26. Peak LHC amplitude vs. frequency at 30 dB antenna isolation.

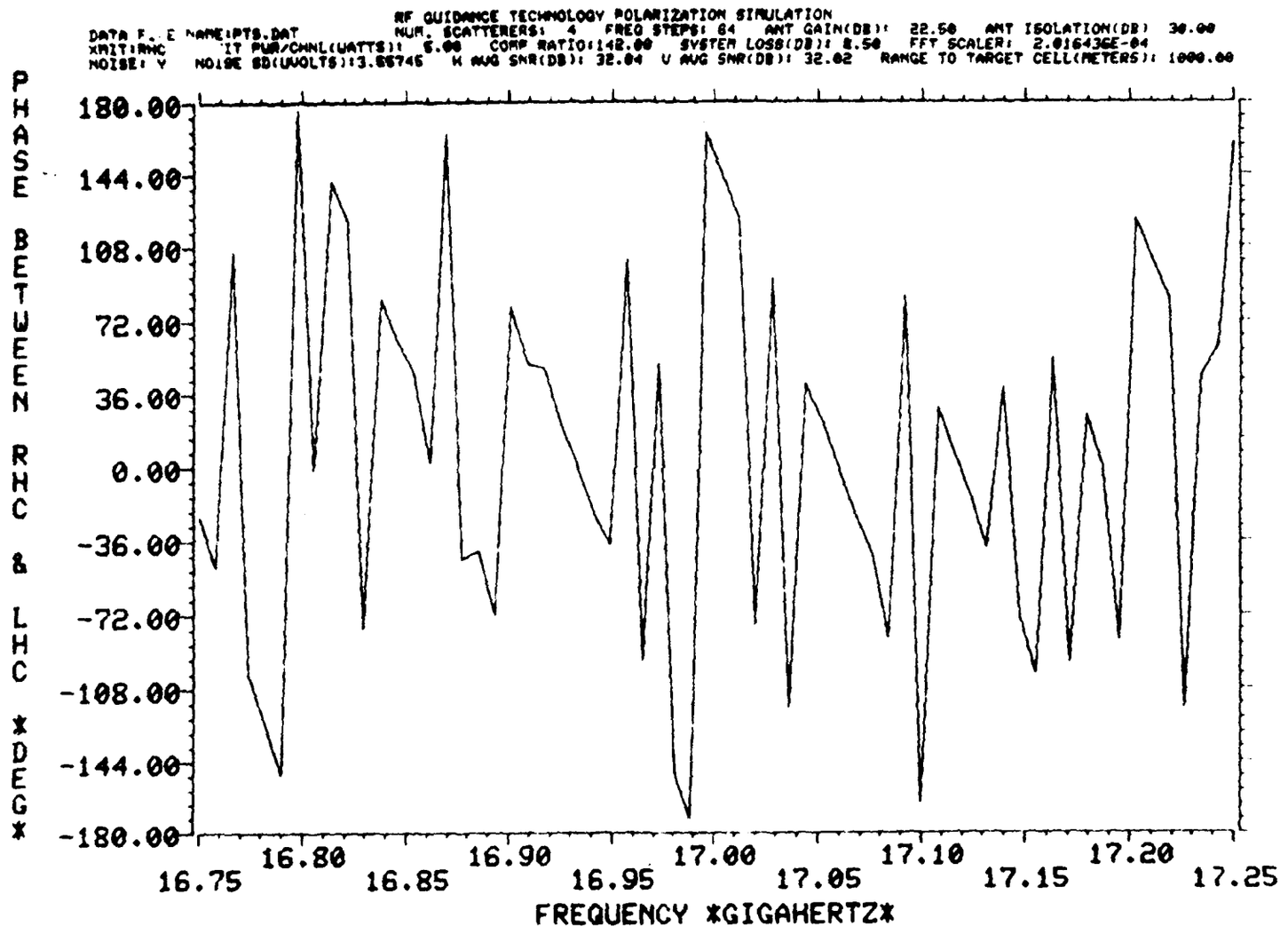


Figure 27. Phase angle between RHC and LHC at 30 dB antenna isolation.

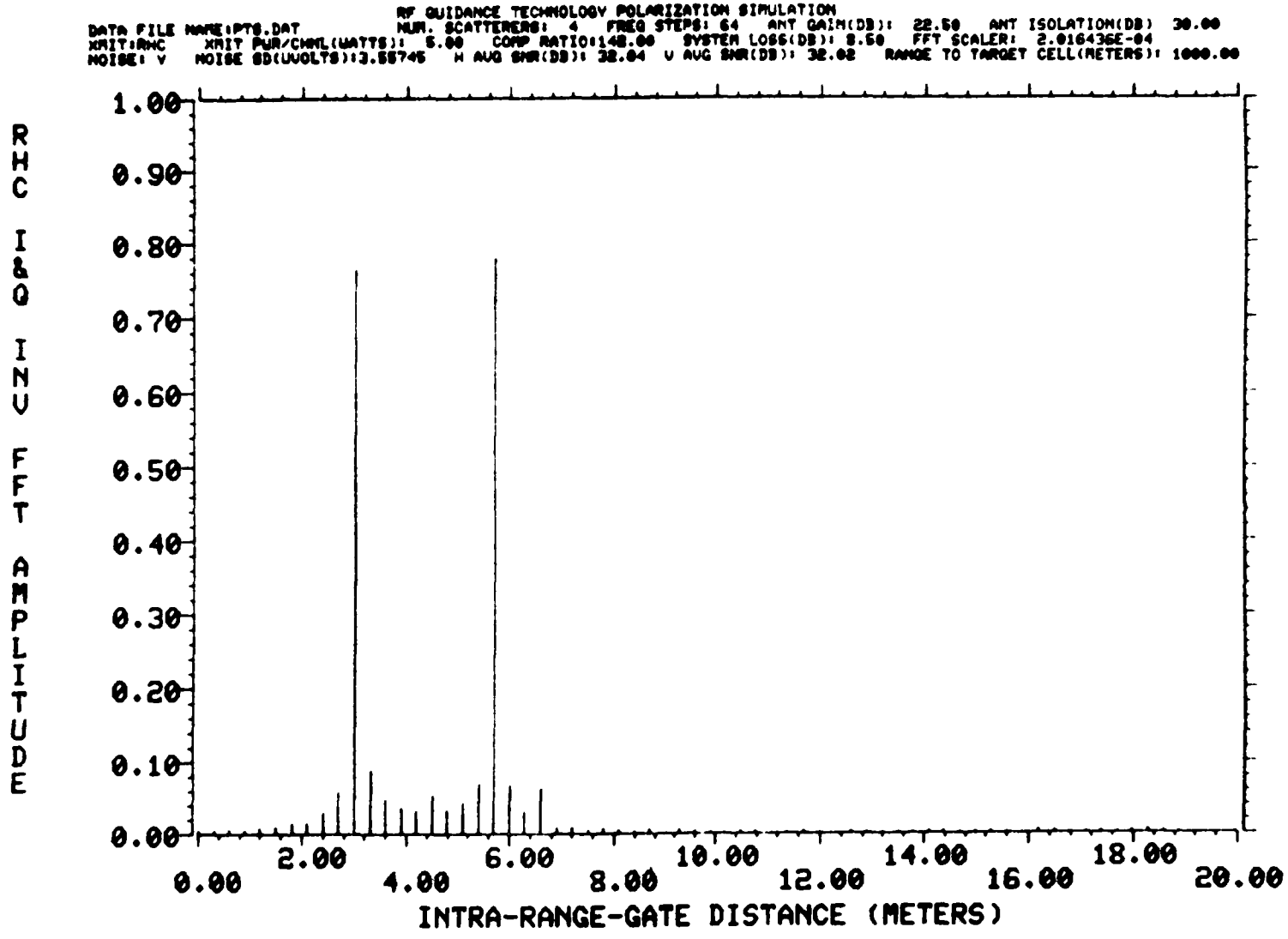


Figure 28. Inverse FFT of RHC I&Q at 30 dB antenna isolation.

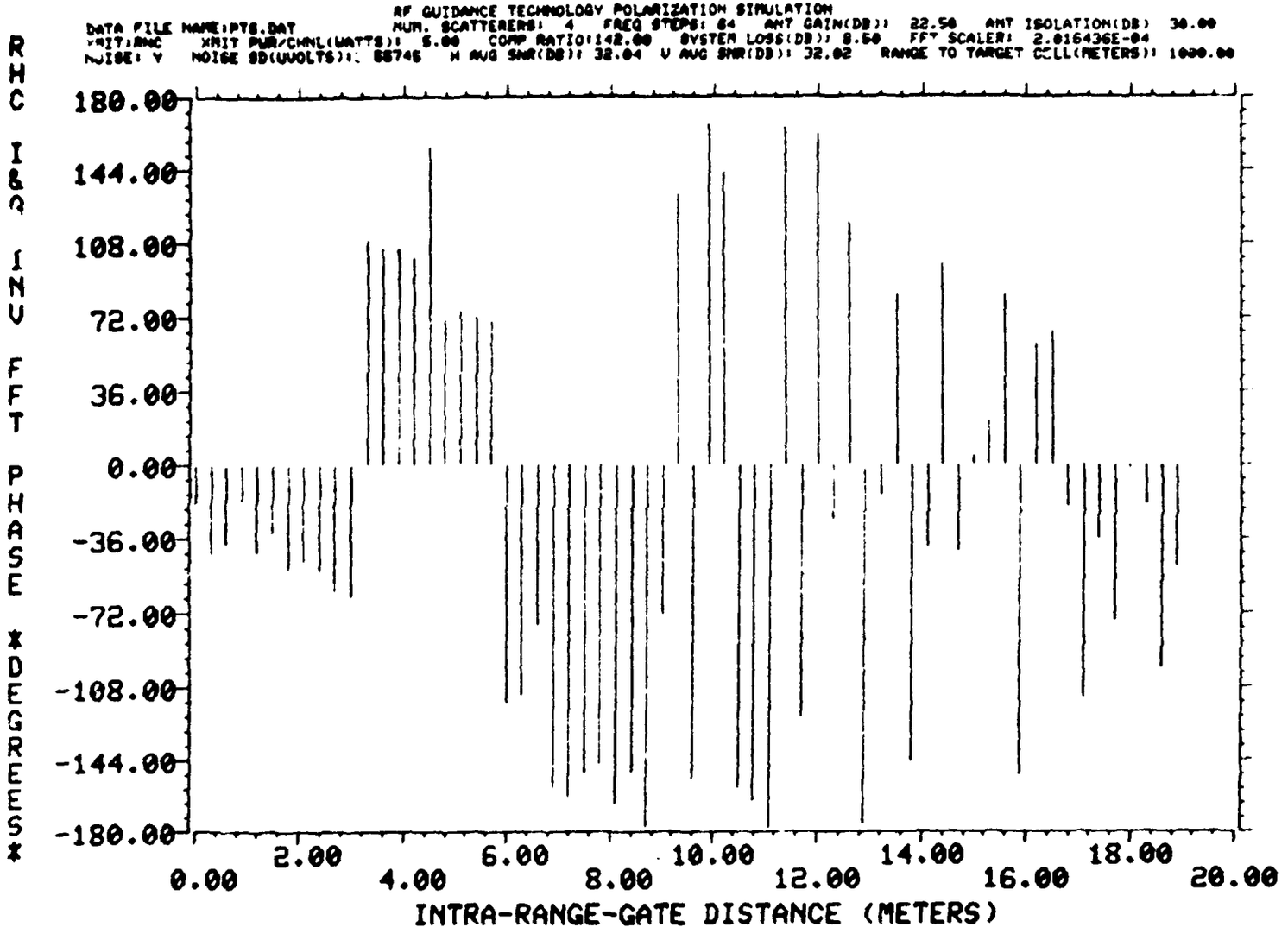


Figure 29. Inverse FFT phase angle for RHC I&Q at 30 dB antenna isolation.

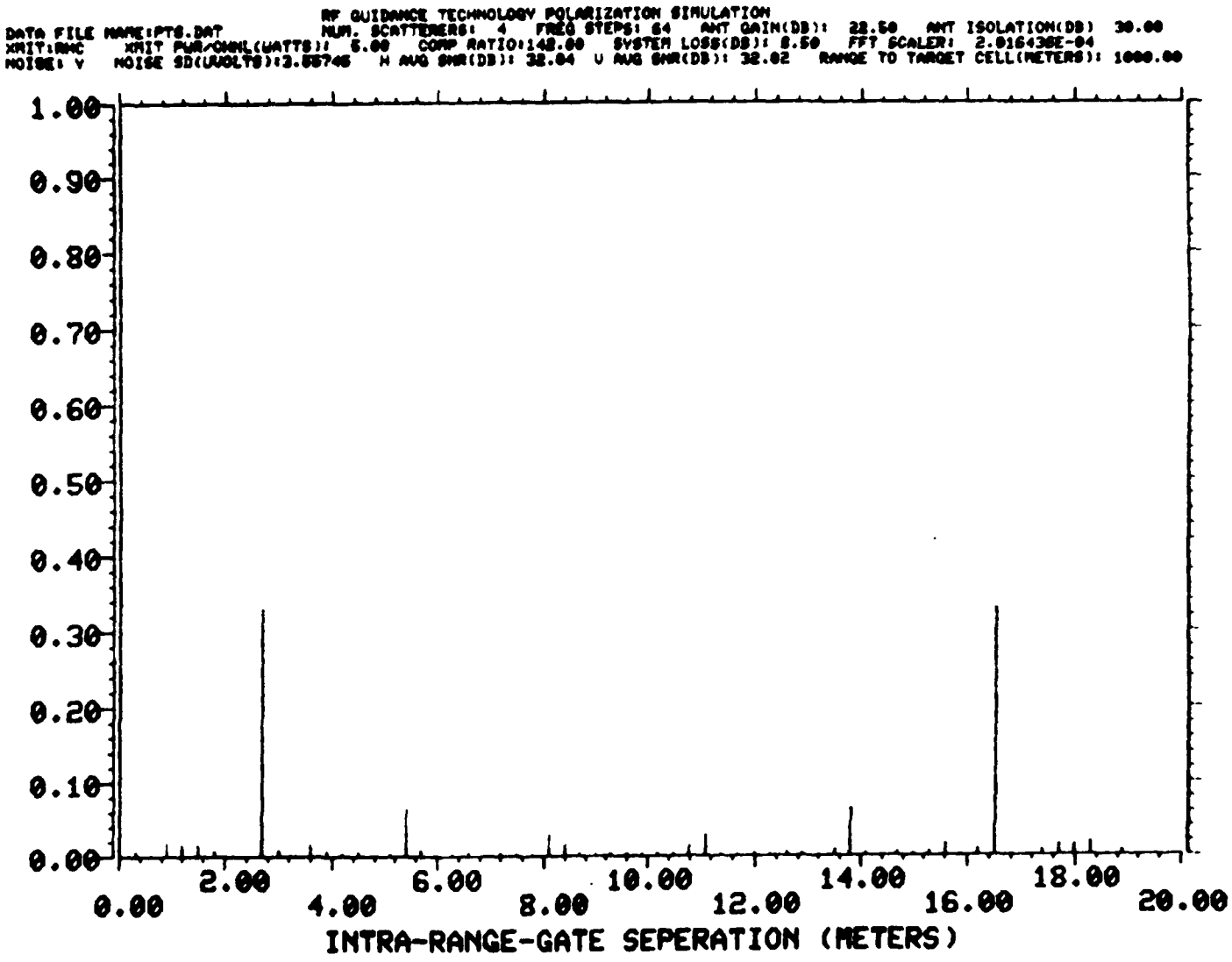


Figure 30. Inverse FFT of peak RHC amplitude at 30 dB antenna isolation.

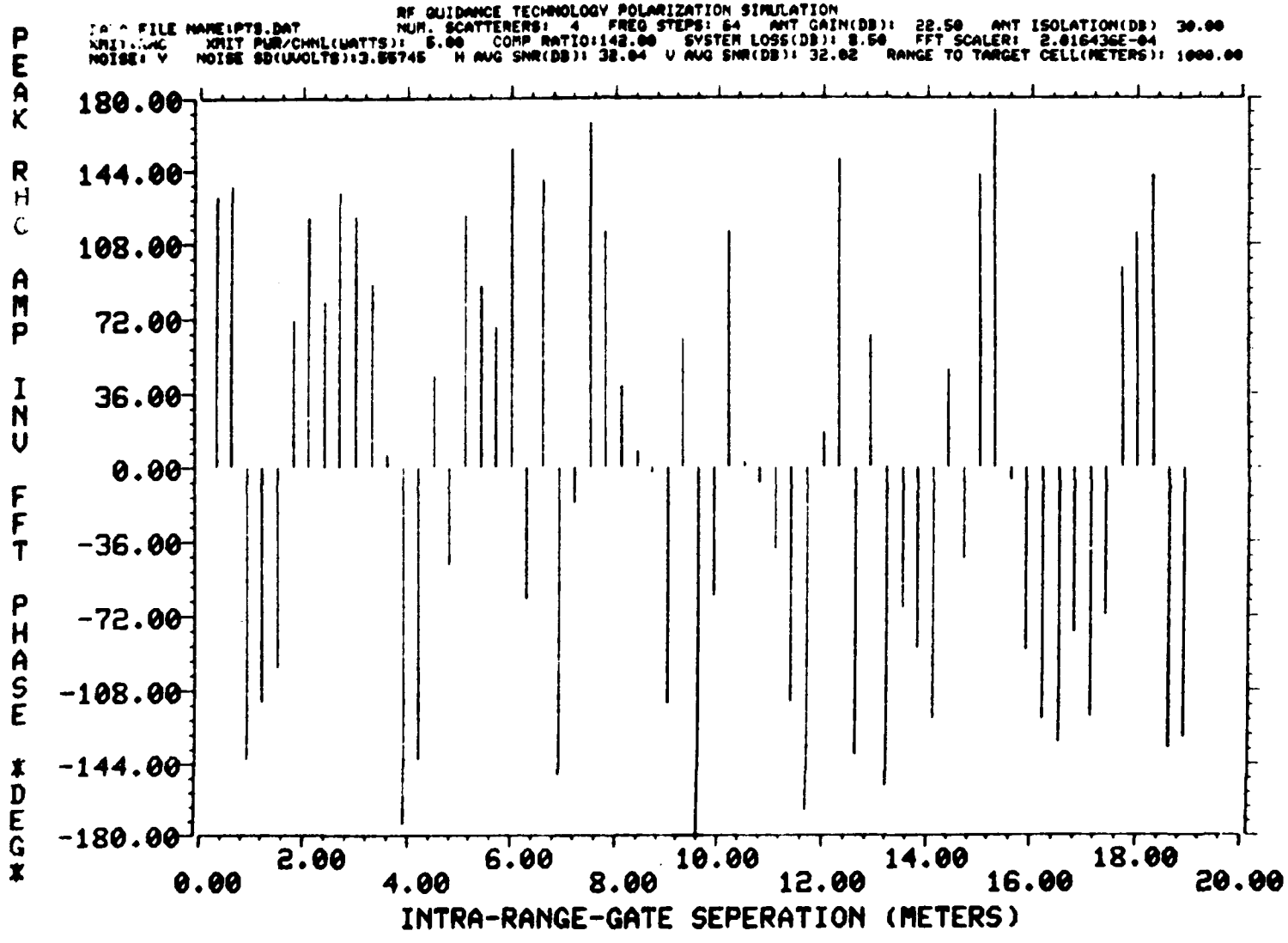


Figure 31. Inverse FFT phase angle of peak RHC amplitude at 30 dB antenna isolation.

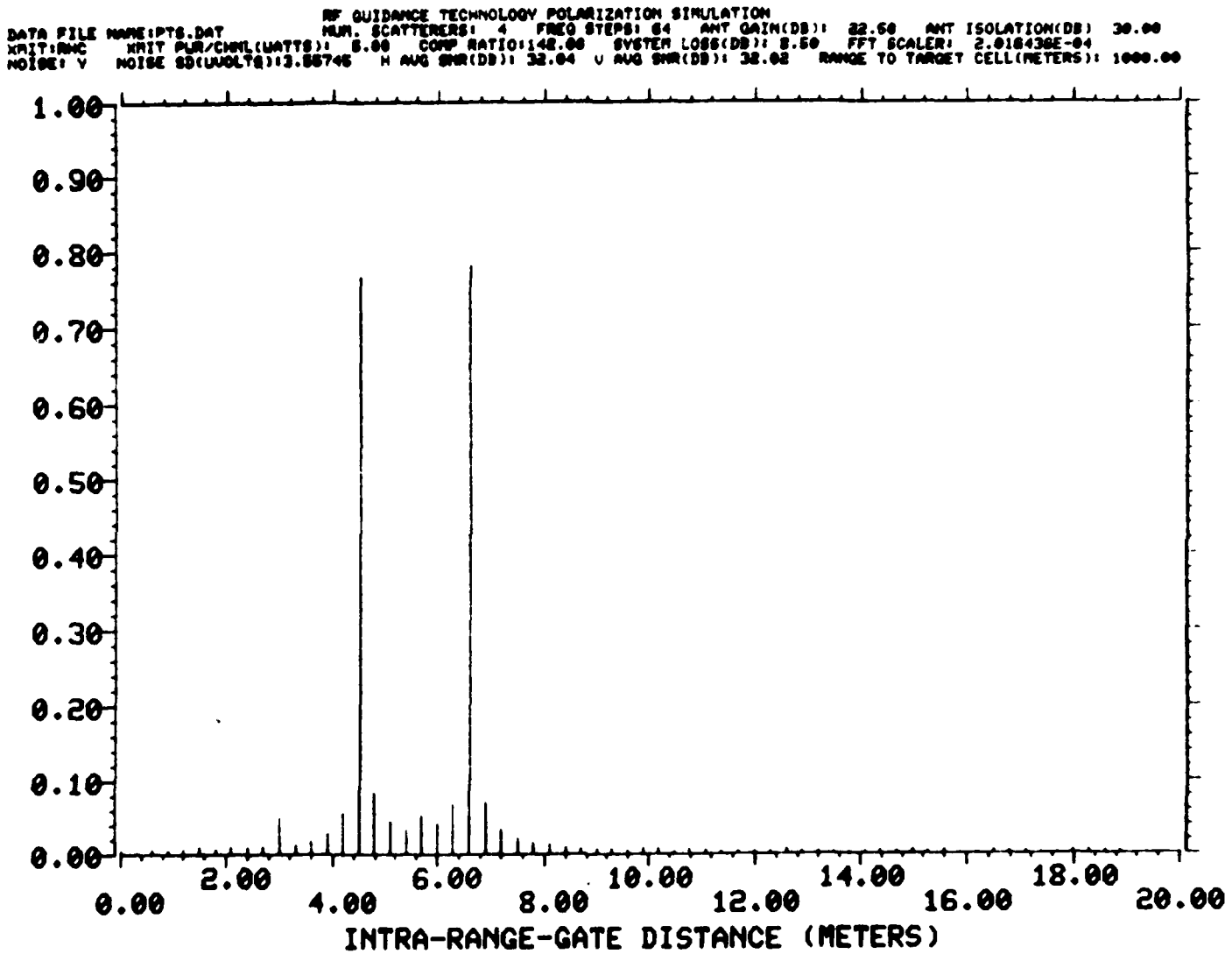


Figure 32. Inverse FFT of LHC I&Q at 30 dB antenna isolation.

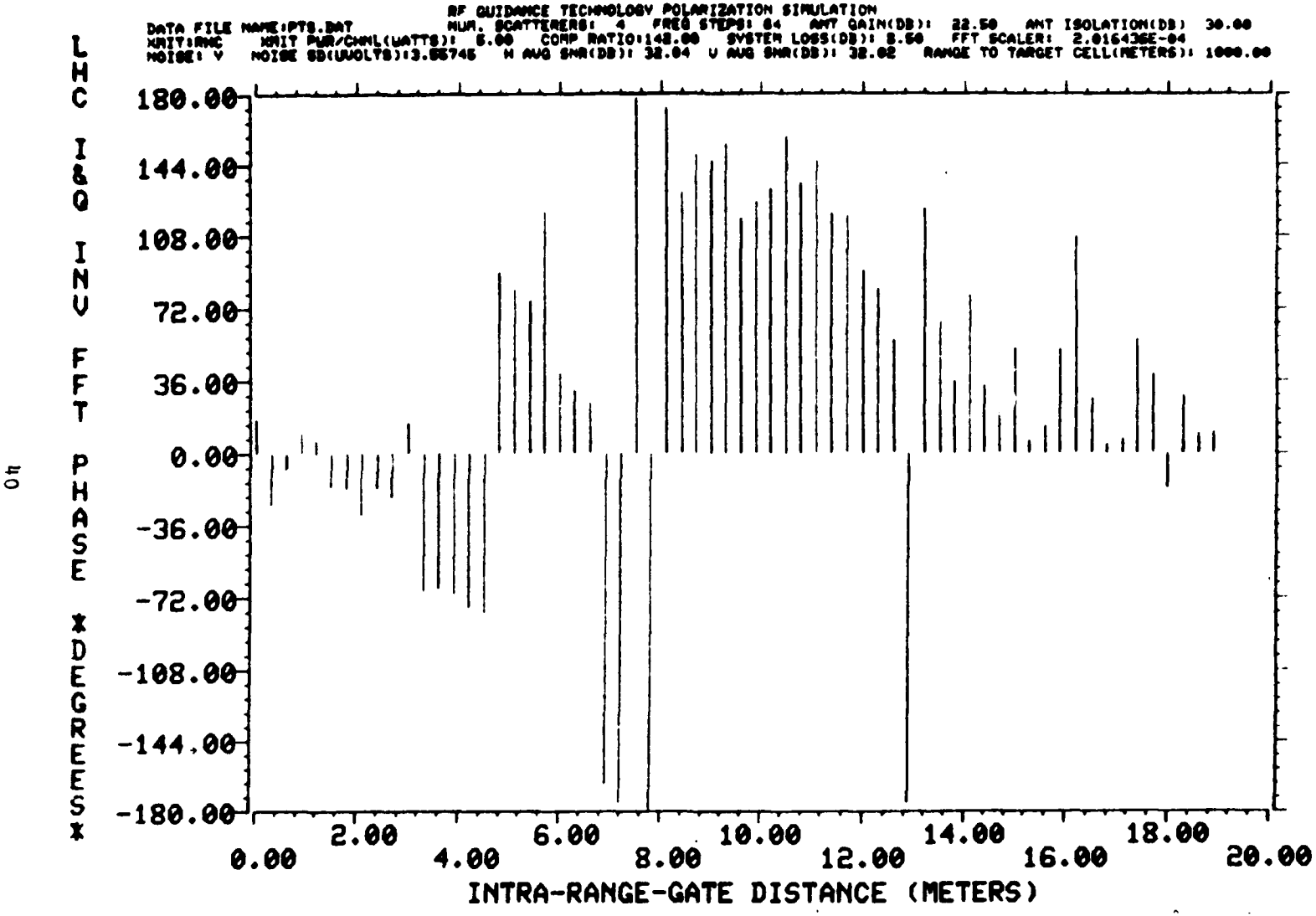


Figure 33. Inverse FFT phase angel for LHC I&Q at 30 dB antenna isolation.

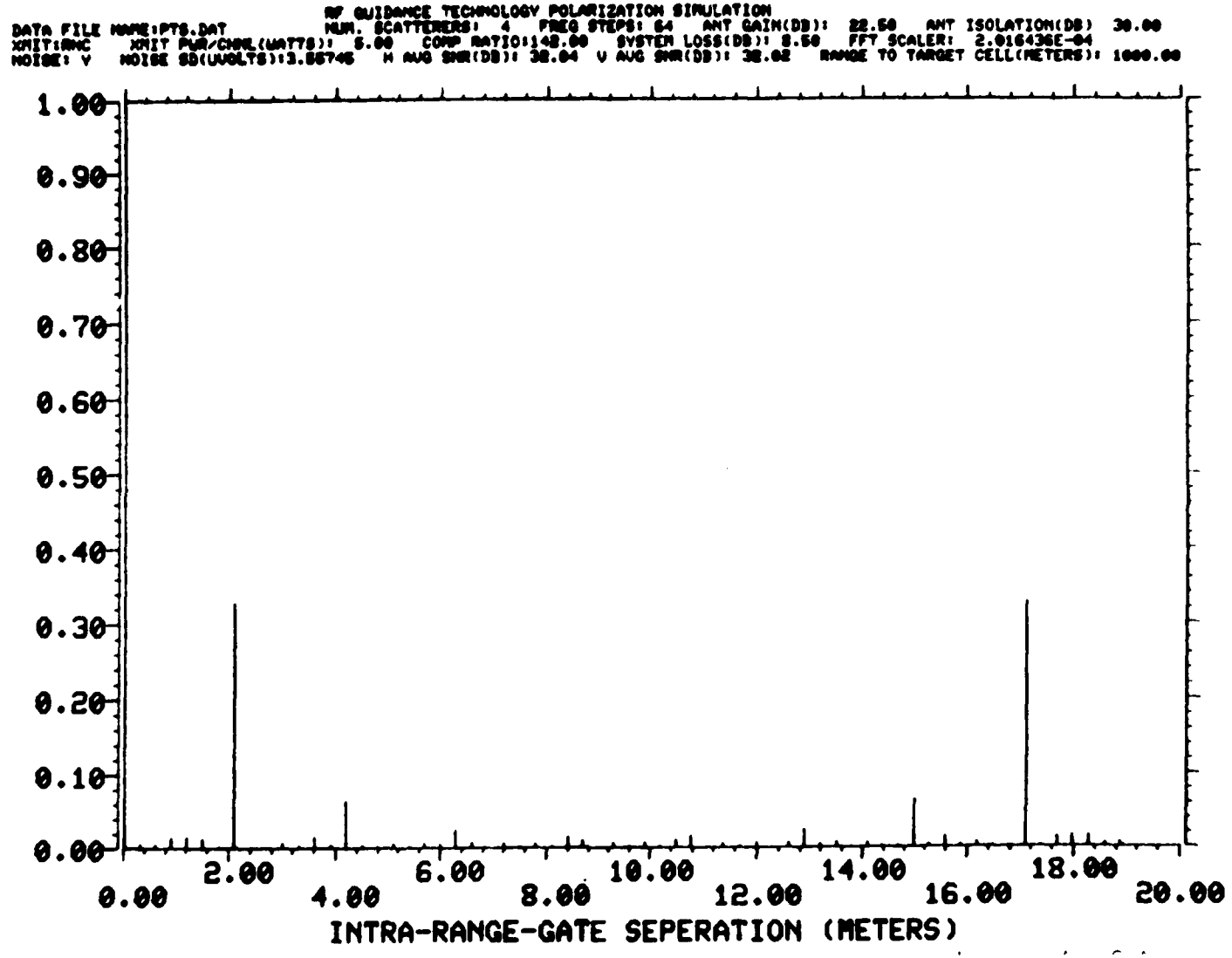


Figure 34. Inverse FFT of peak LHC amplitude at 30 dB antenna isolation.

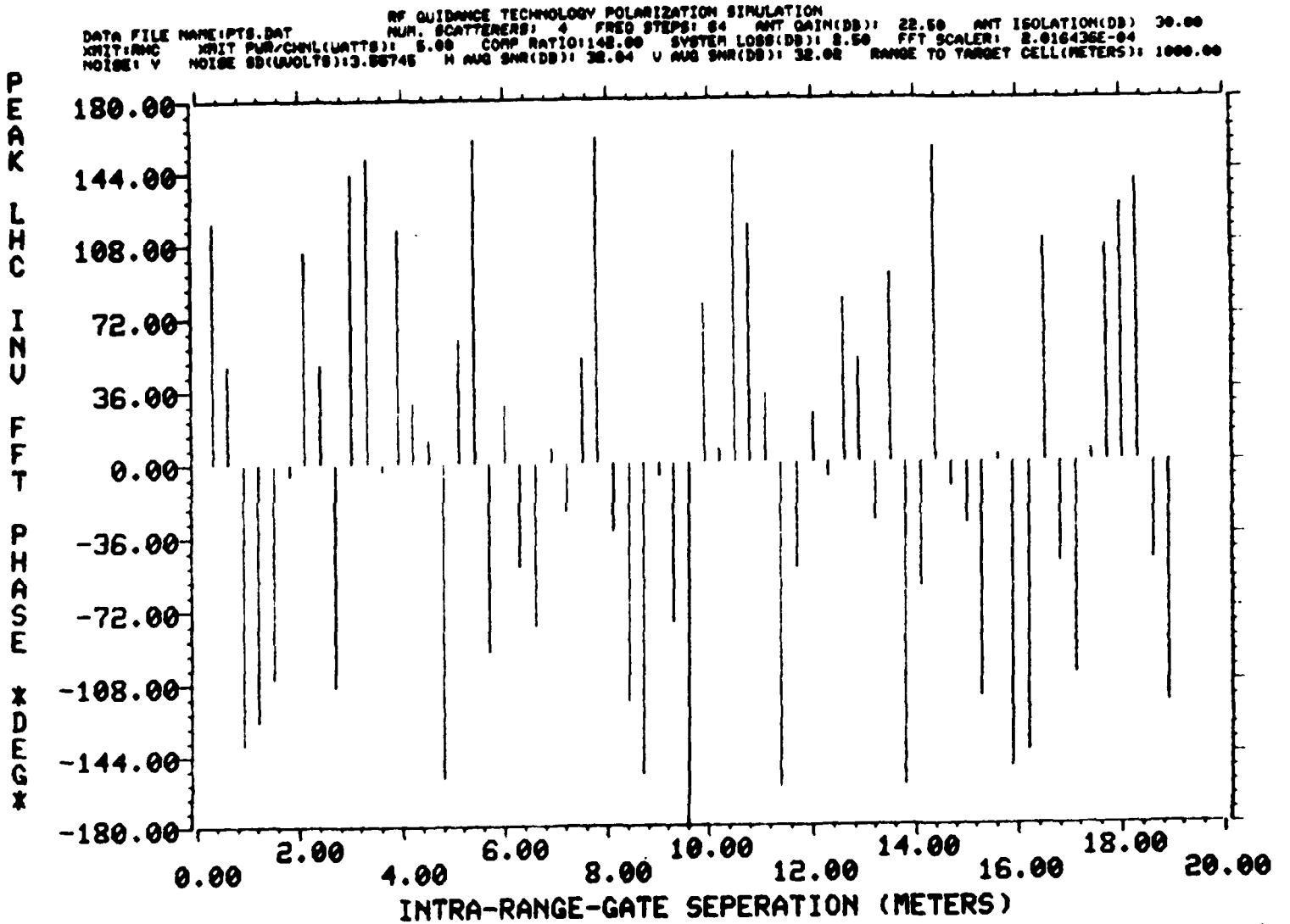


Figure 35. Inverse FFT phase angle of peak LHC amplitude at 30 dB antenna isolation.

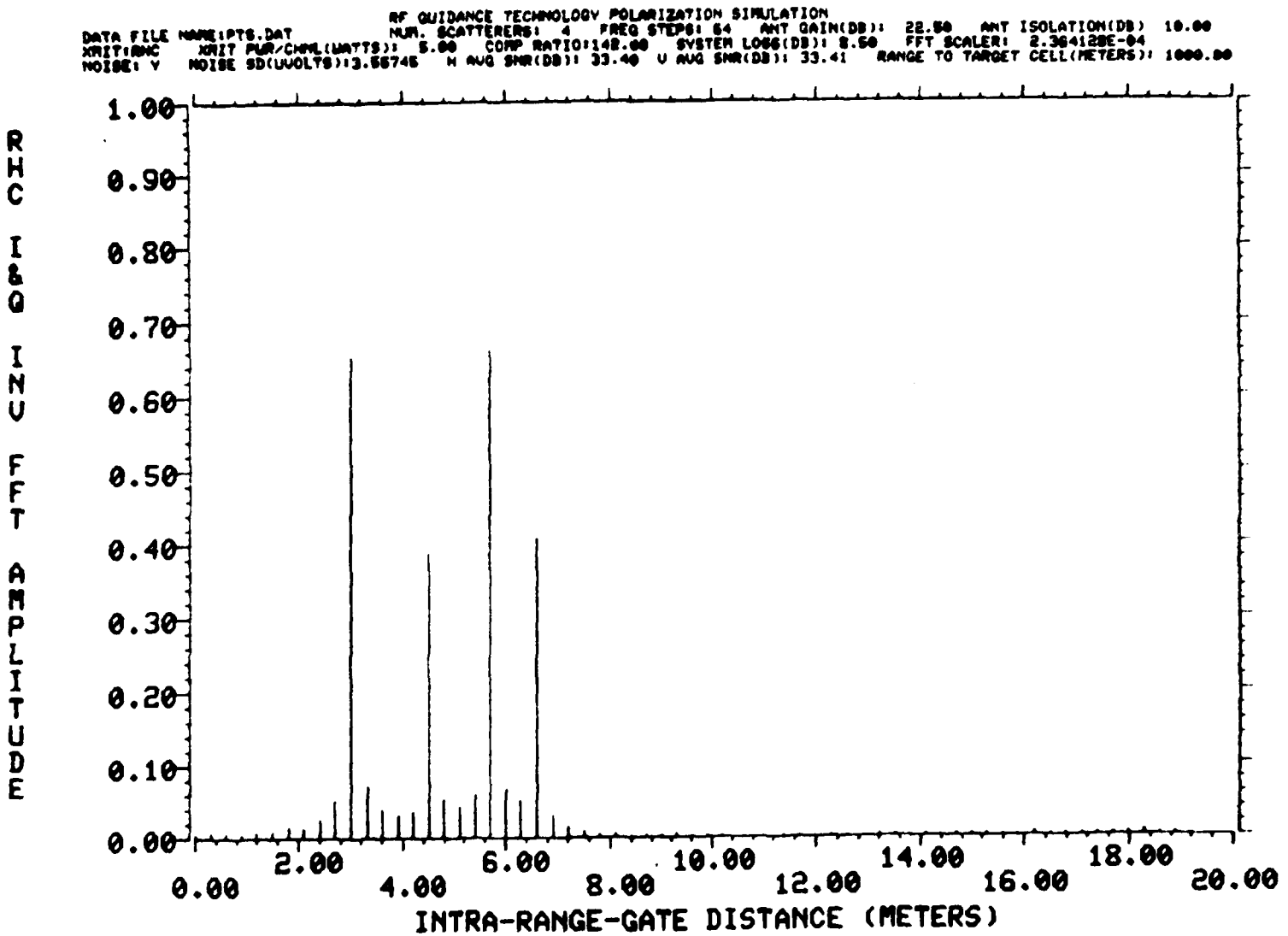


Figure 36. Inverse FFT of RHC I&Q at 10 dB antenna isolation.

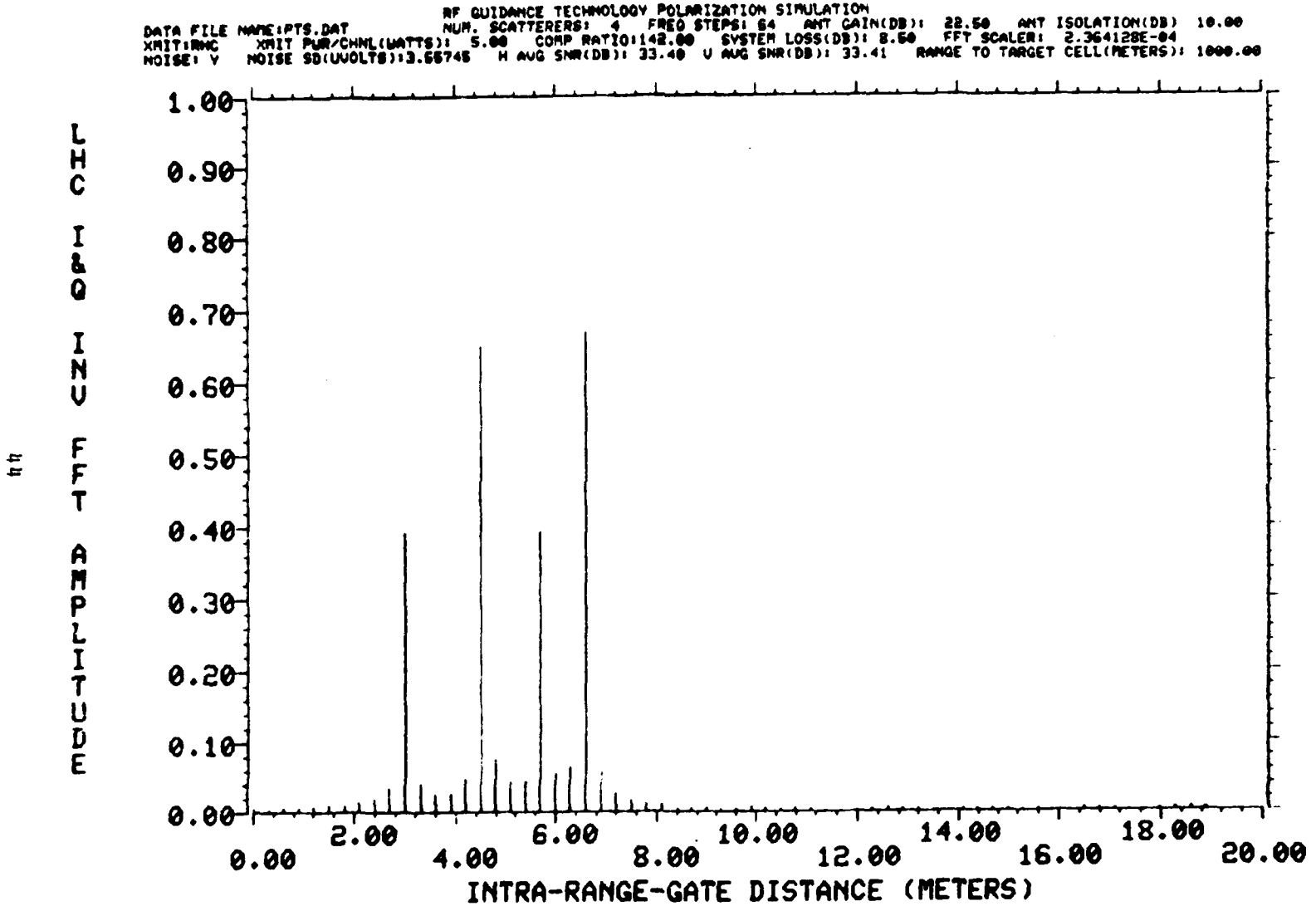


Figure 37. Inverse FFT of LHC I&Q at 10 dB antenna Isolation.

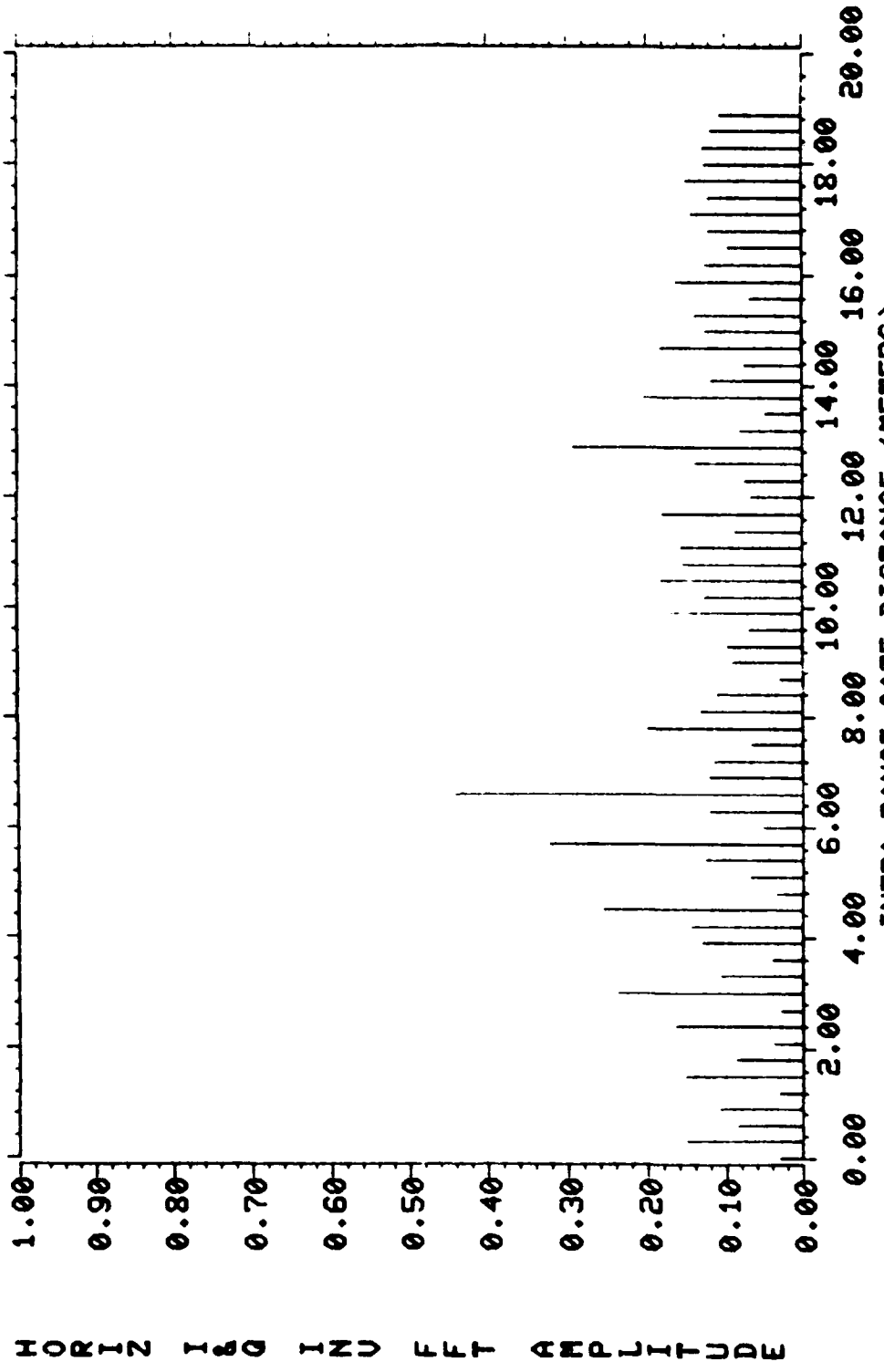


Figure 38. Inverse FFT of horizontal I&Q, single pulse S/N equal -8 dB.

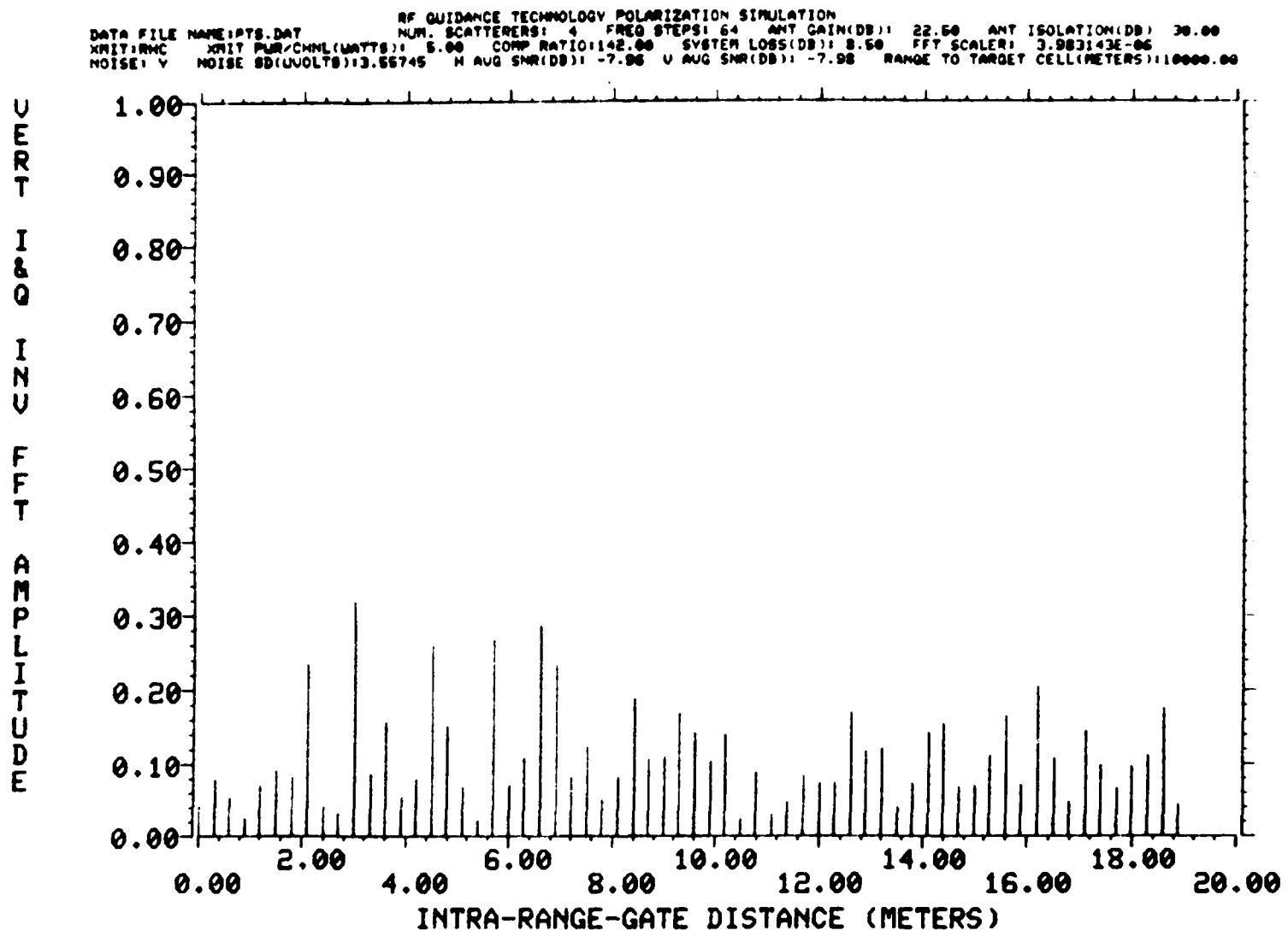


Figure 39. Inverse FFT of vertical I&Q, single pulse S/N equal -8 dB.

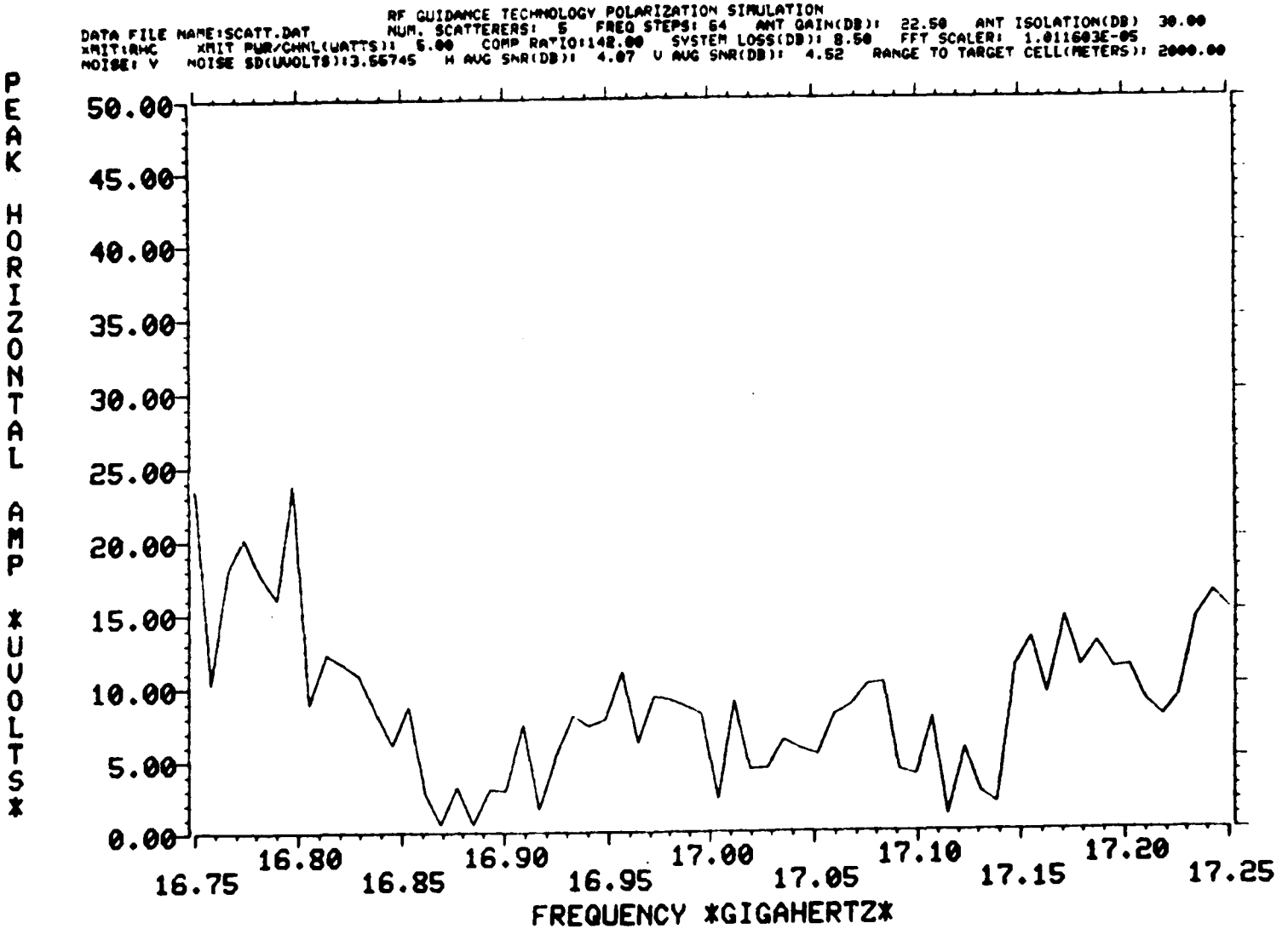


Figure 40. Peak horizontal voltage for target model only.

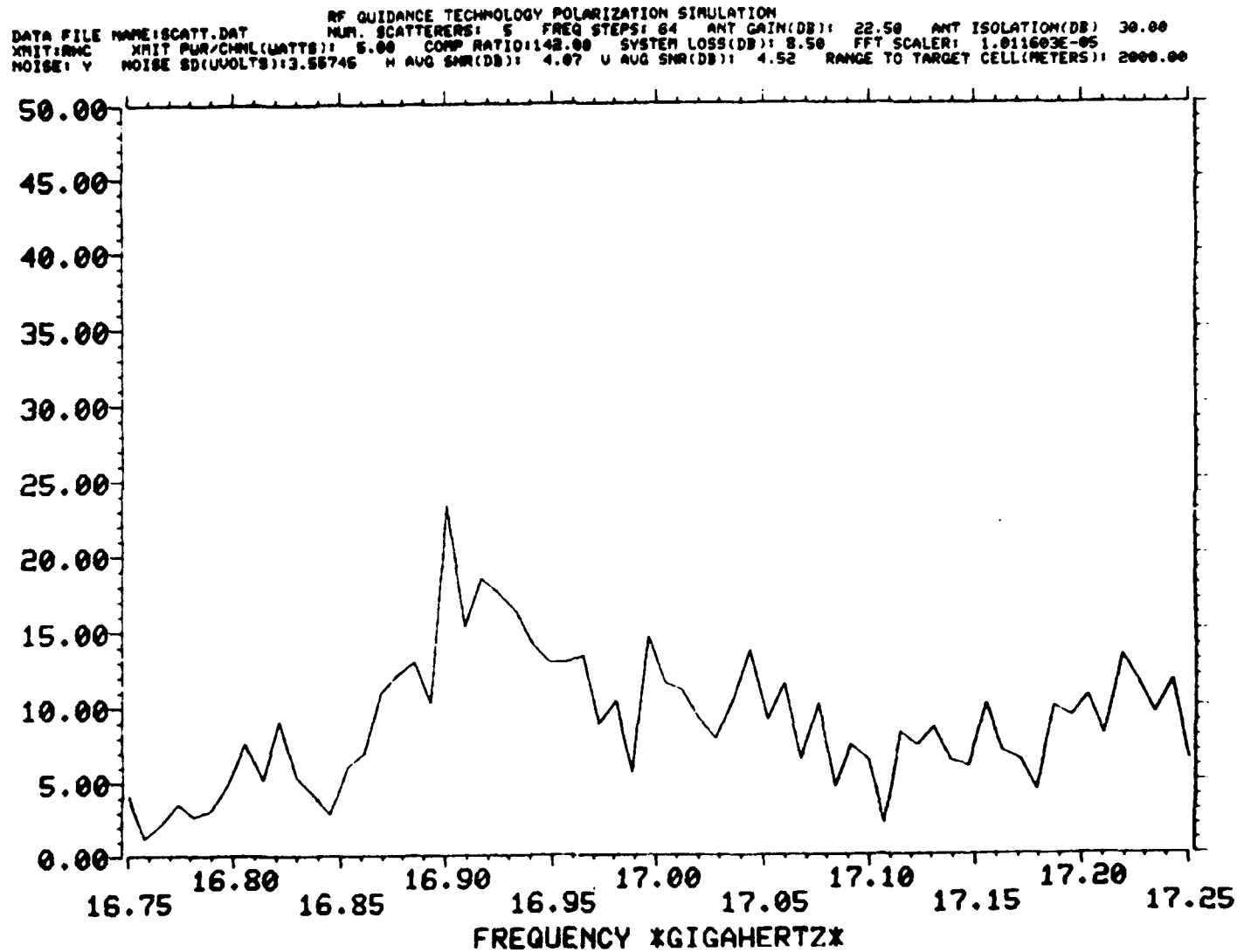


Figure 41. Peak vertical voltage for target model only.

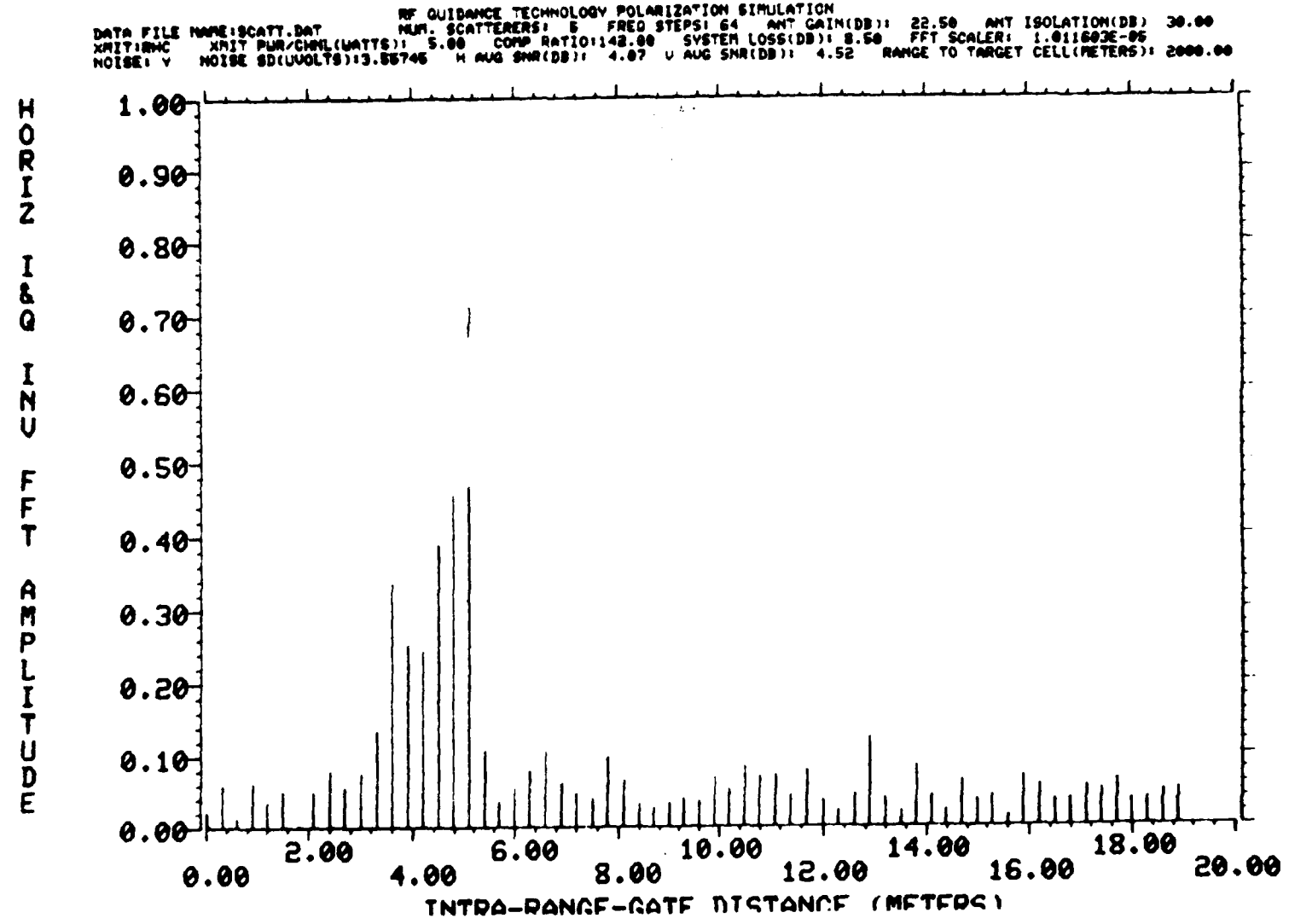


Figure 42. Inverse FFT of horizontal I&Q for tank model only.

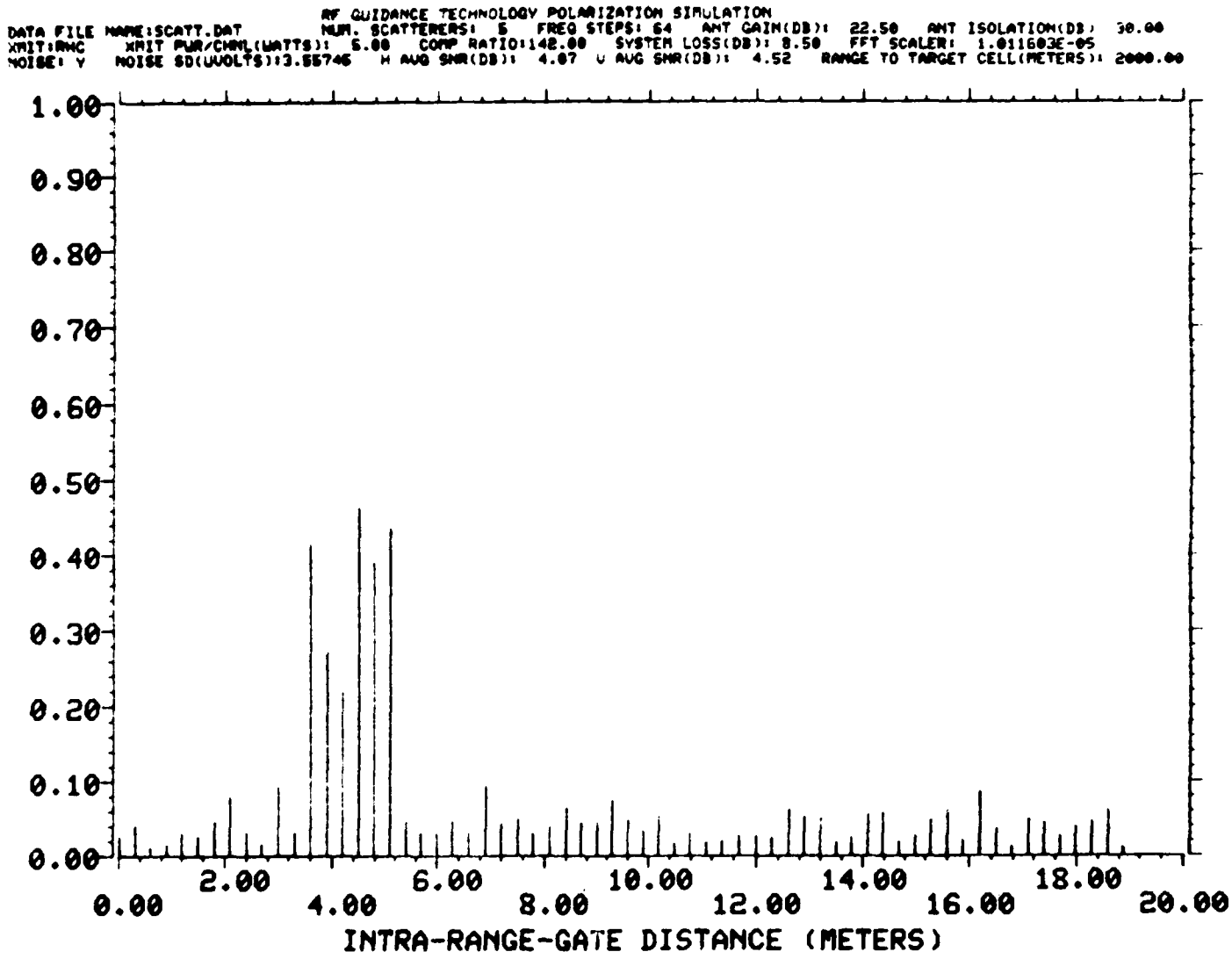


Figure 43. Inverse FFT of vertical I&Q for target model only.

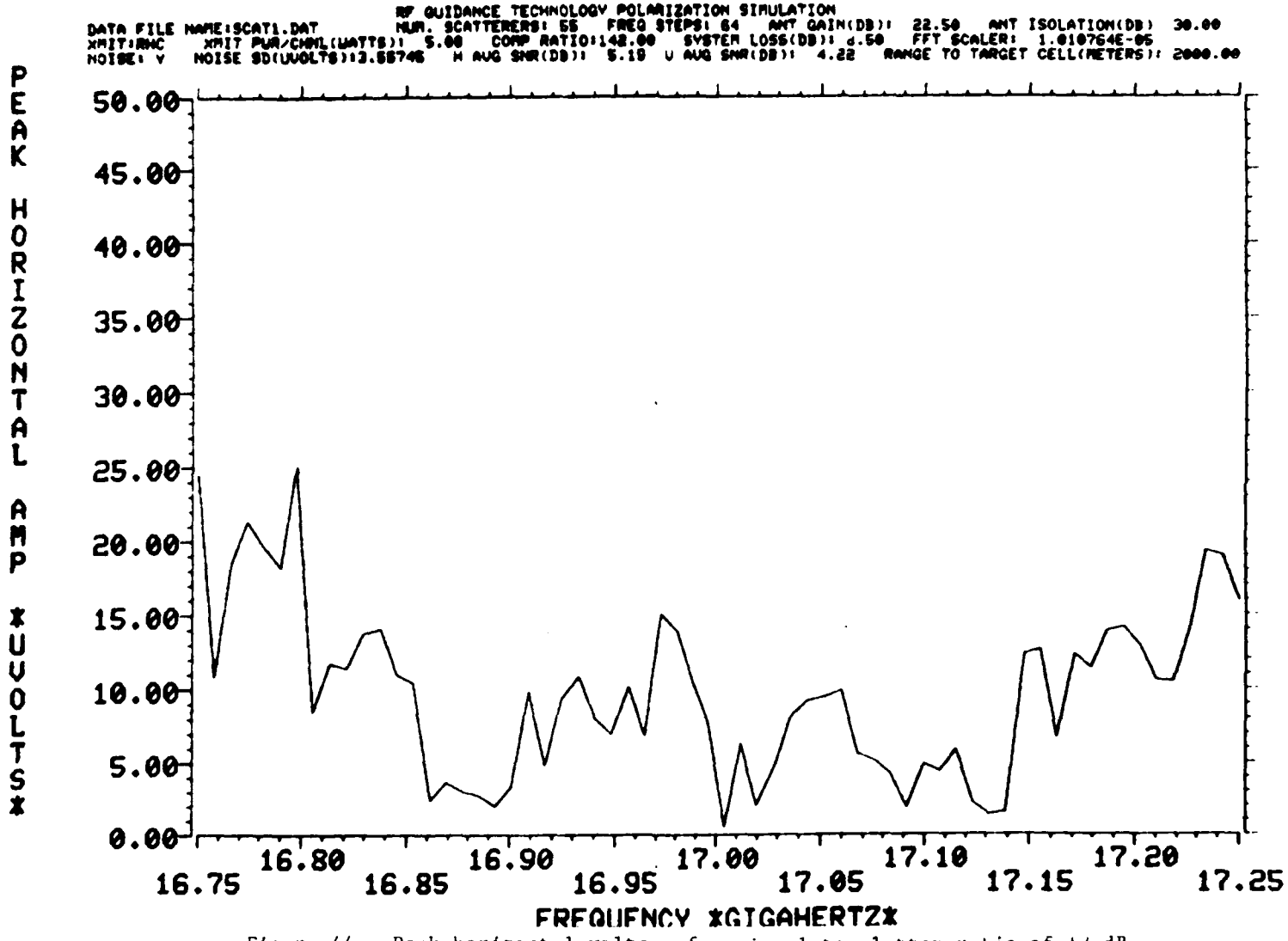


Figure 44. Peak horizontal voltage for signal to clutter ratio of +/ dB.

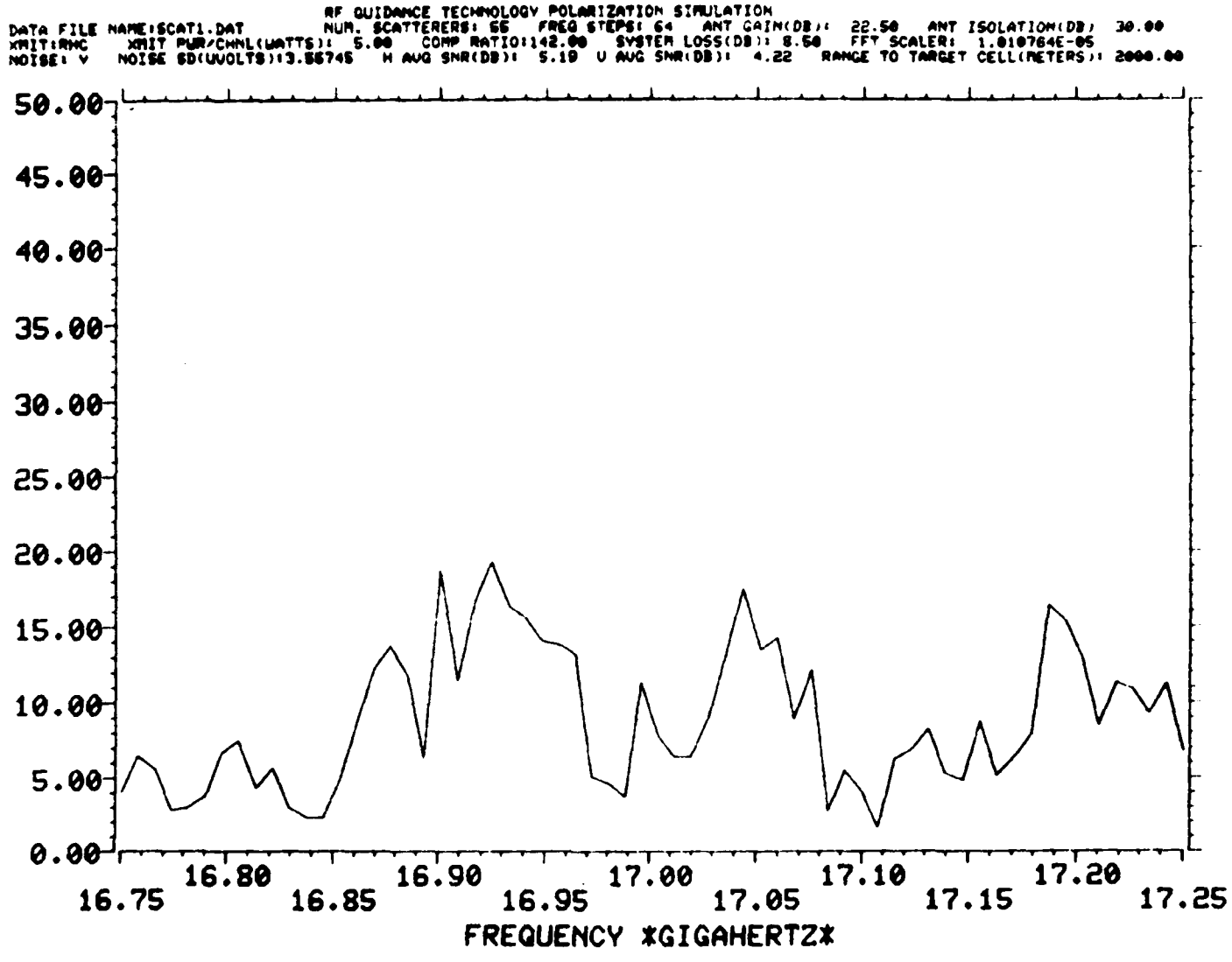


Figure 45. Peak vertical voltage for signal to clutter ratio of +7 dB.

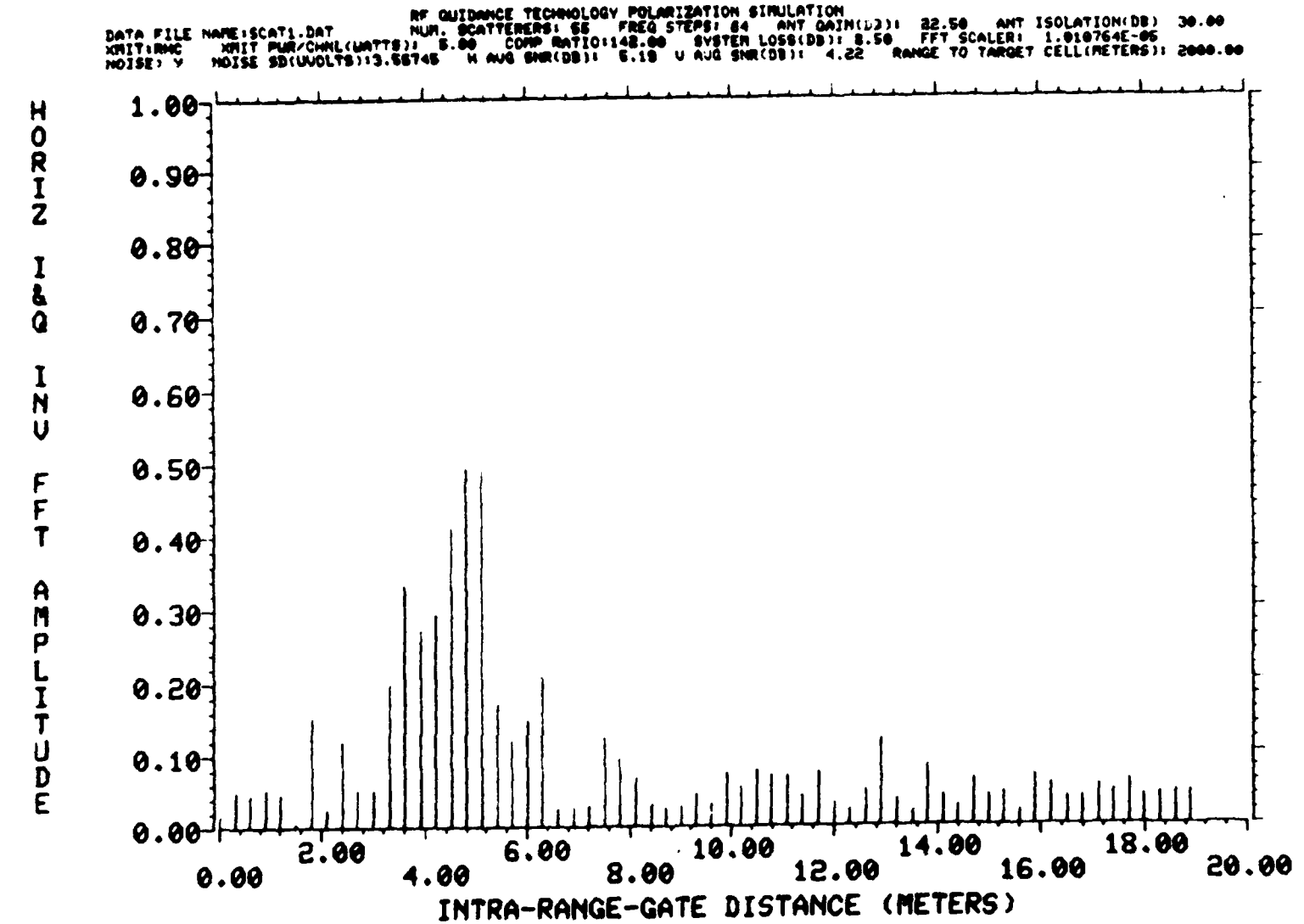


Figure 46. Inverse FFT of horizontal I&Q tor signal to clutter ratio of +7 dB.

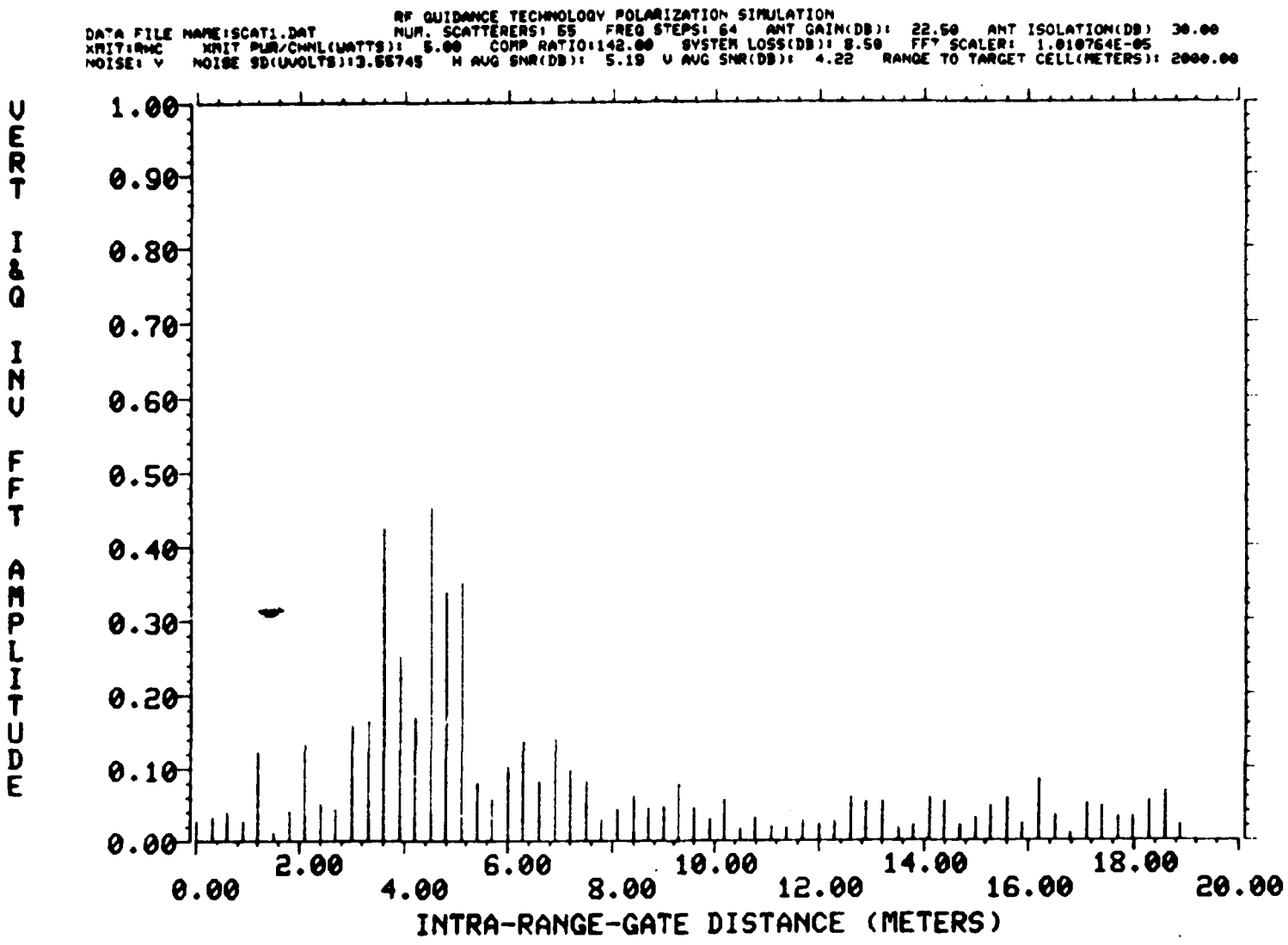


Figure 47. Inverse FFT of vertical I&Q for signal to clutter ratio of +7 dB.

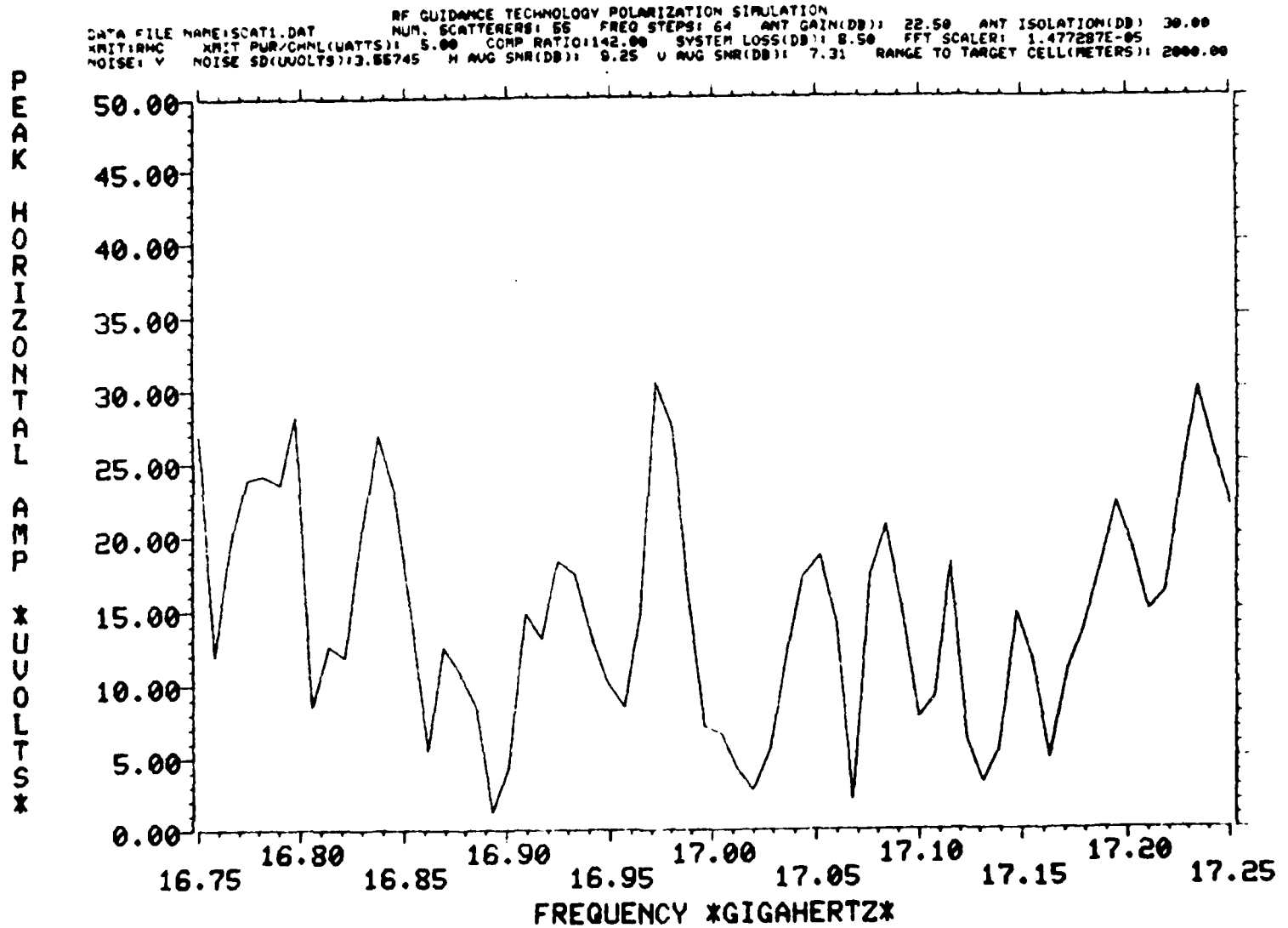


Figure 48. Peak horizontal voltage for signal to clutter ratio of -3 dB.

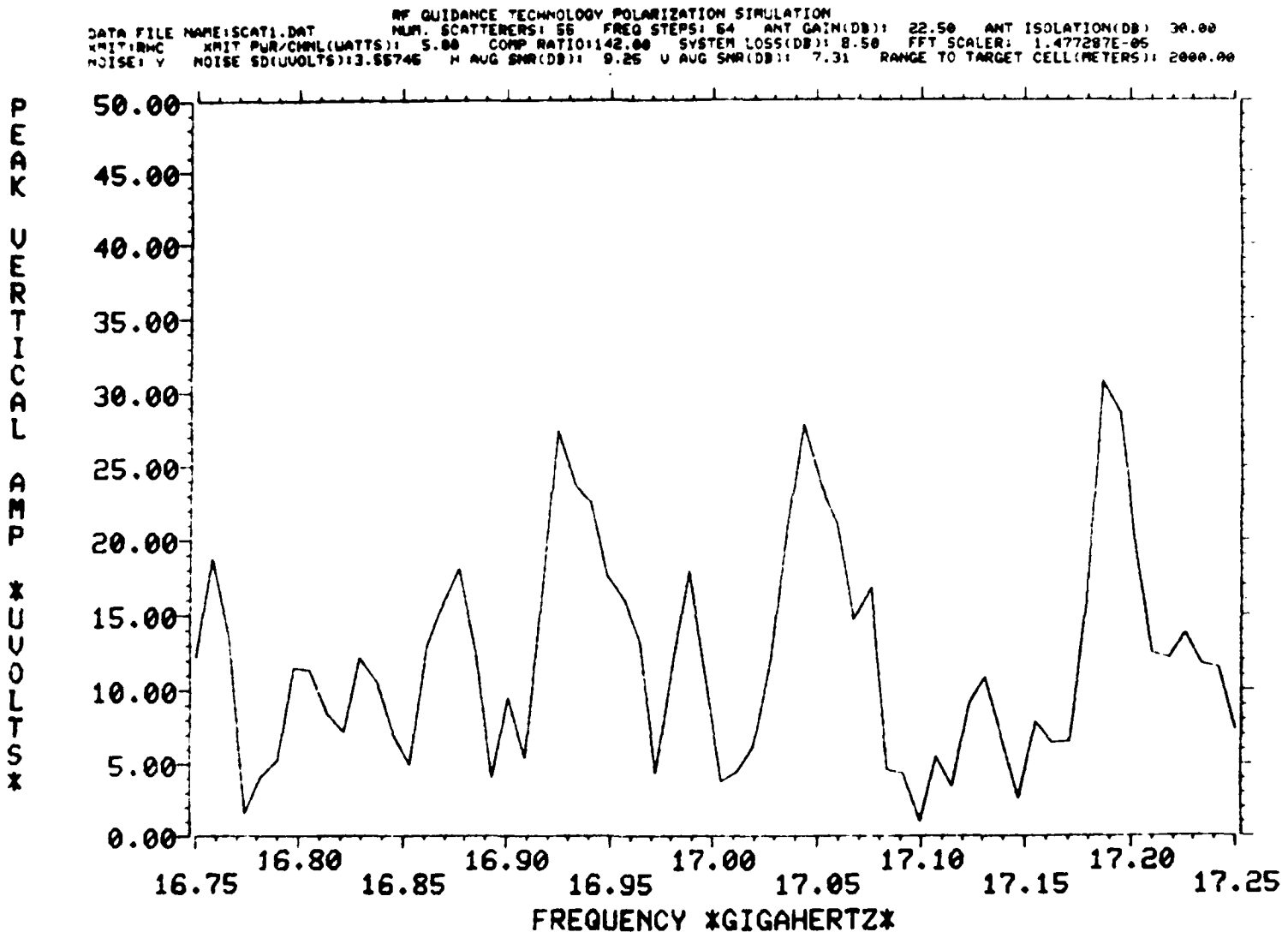


Figure 49. Peak vertical voltage for signal to clutter ratio of -3 dB.

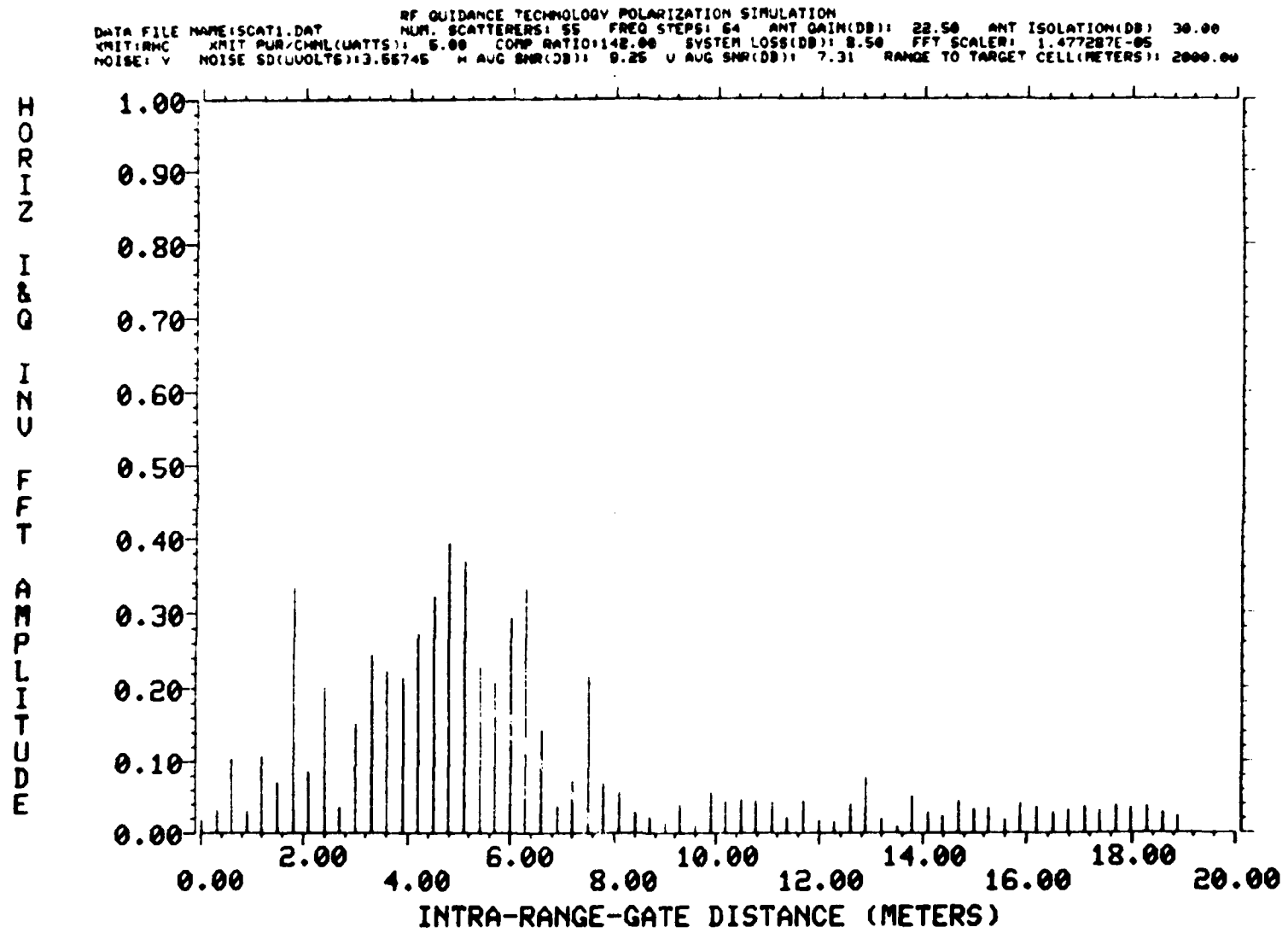


Figure 50. Inverse FFT of horizontal I&Q for signal to clutter ratio of -3 dB.

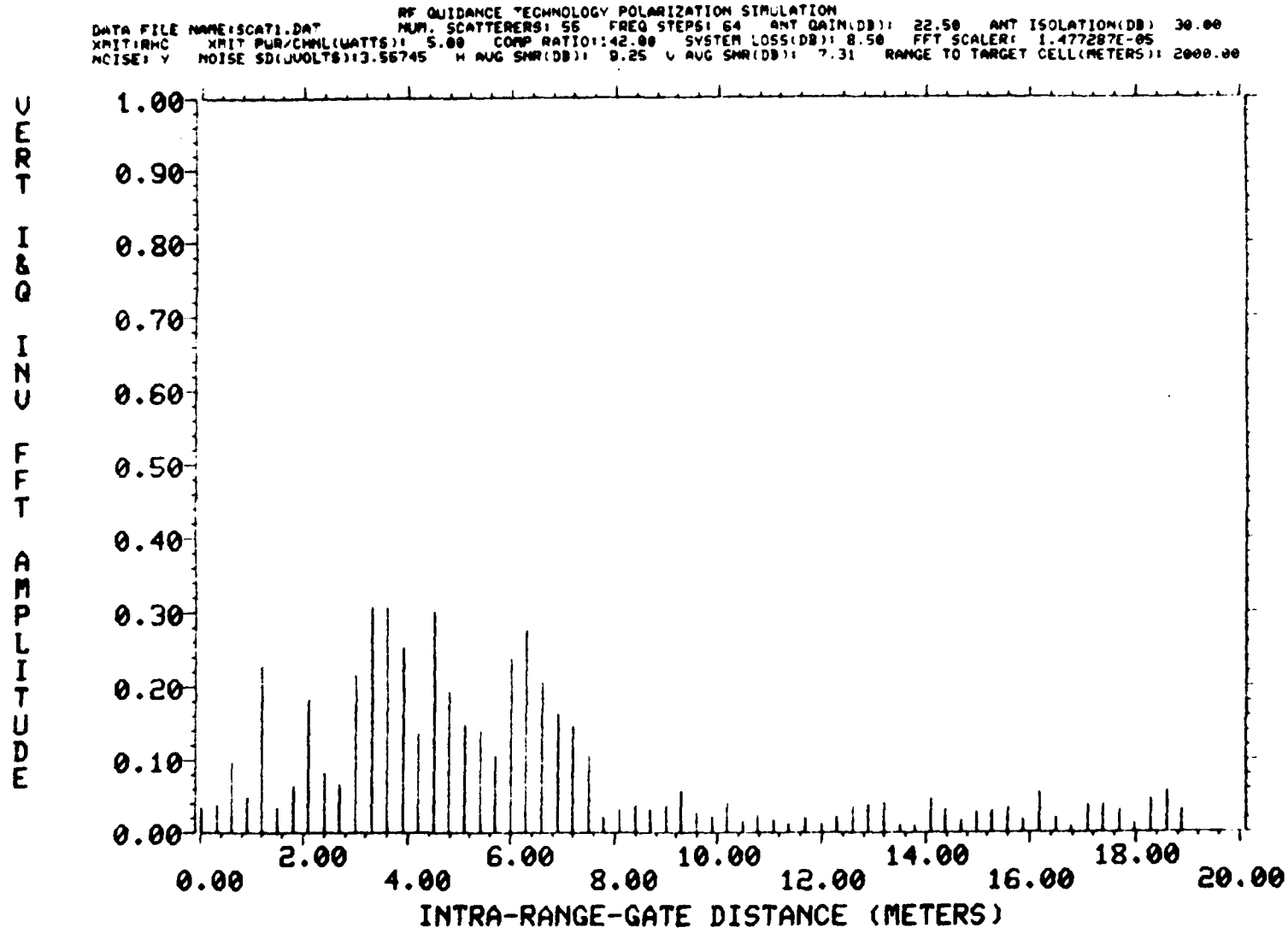


Figure 51. Inverse FFT of vertical I&Q for signal to clutter ratio of -3 dB.

APPENDIX

Simulation Flow Charts and Program Listing

The following flow charts were developed as an aid in following the mathematical development of signals. The program listing of subroutines is short enough to provide an easily followed path without flowcharts. This simulation has been developed and run on a Digital Equipment Corporation (DEC) 28K word PDP-11/10 computer running DEC's RT-11 operating system. The plots were performed utilizing a Tektronix terminal 4014 driven by in-house developed plotting software. The plotting subroutines are described by function only without software listing included in the program printout. This will provide a programming guide for tailoring plots to other systems.

DEC's RT-11 Subroutines

Call Assign - Attaches a disk file for reading or writing and assigns a logical unit number.

Call Close - Closes an attached disk file.

In-House Computer Subroutines

Call IAND - Performs logical bit anding of the two arguments.

Call SWR - Read computer switch register. Used to control line printer and hard copy functions.

Call NLOGN - Perform forward or inverse in place FFT of complex array.

In-House Plotting subroutines

Call PLOT - Erase 4014 screen.

Call V14CSZ - Select size of Alphanumeric characters typed on 4014.

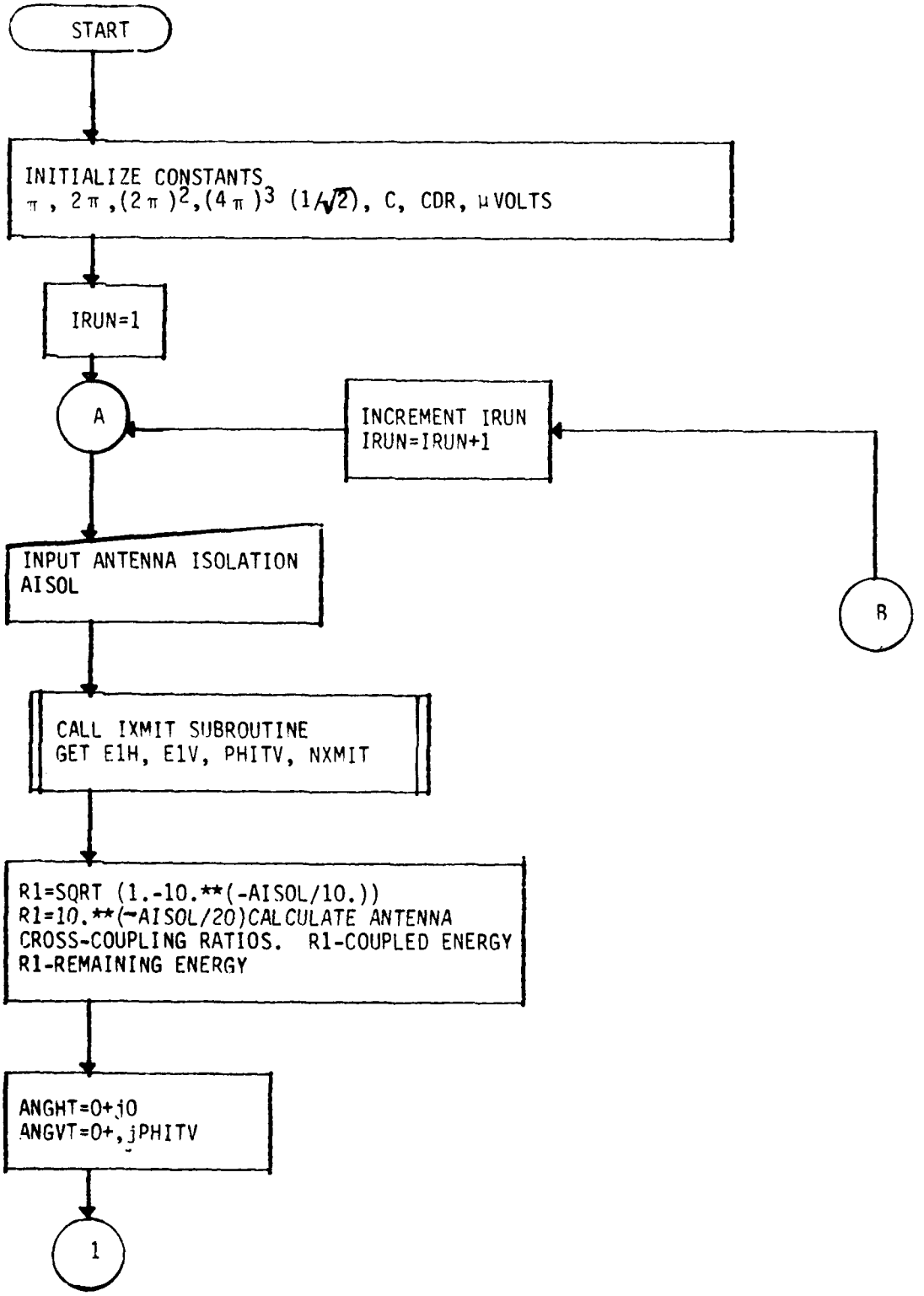
Call AXES - Draw plotting axes by screen position and tic-marks controlled by user units and store parameters for user units plotting by call LINE.

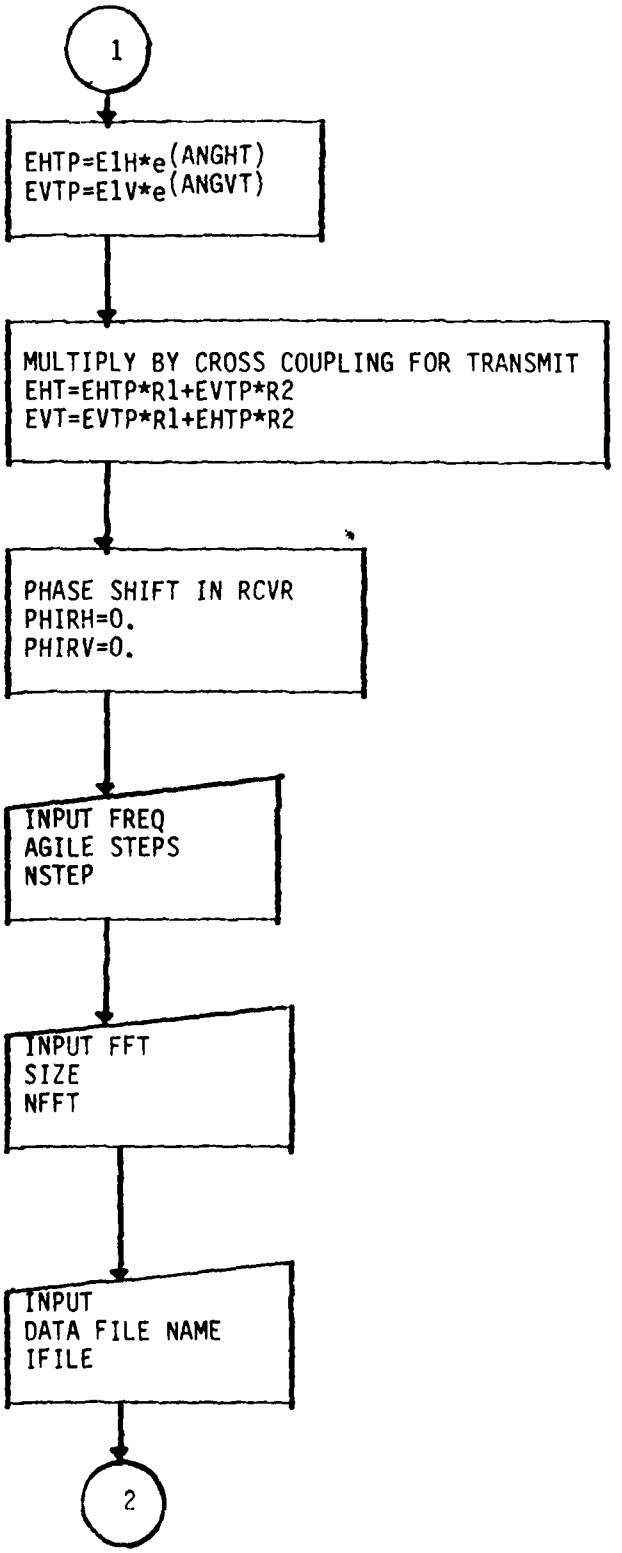
Call LINE - Draw graphic line on 4014 between two points described by user units.

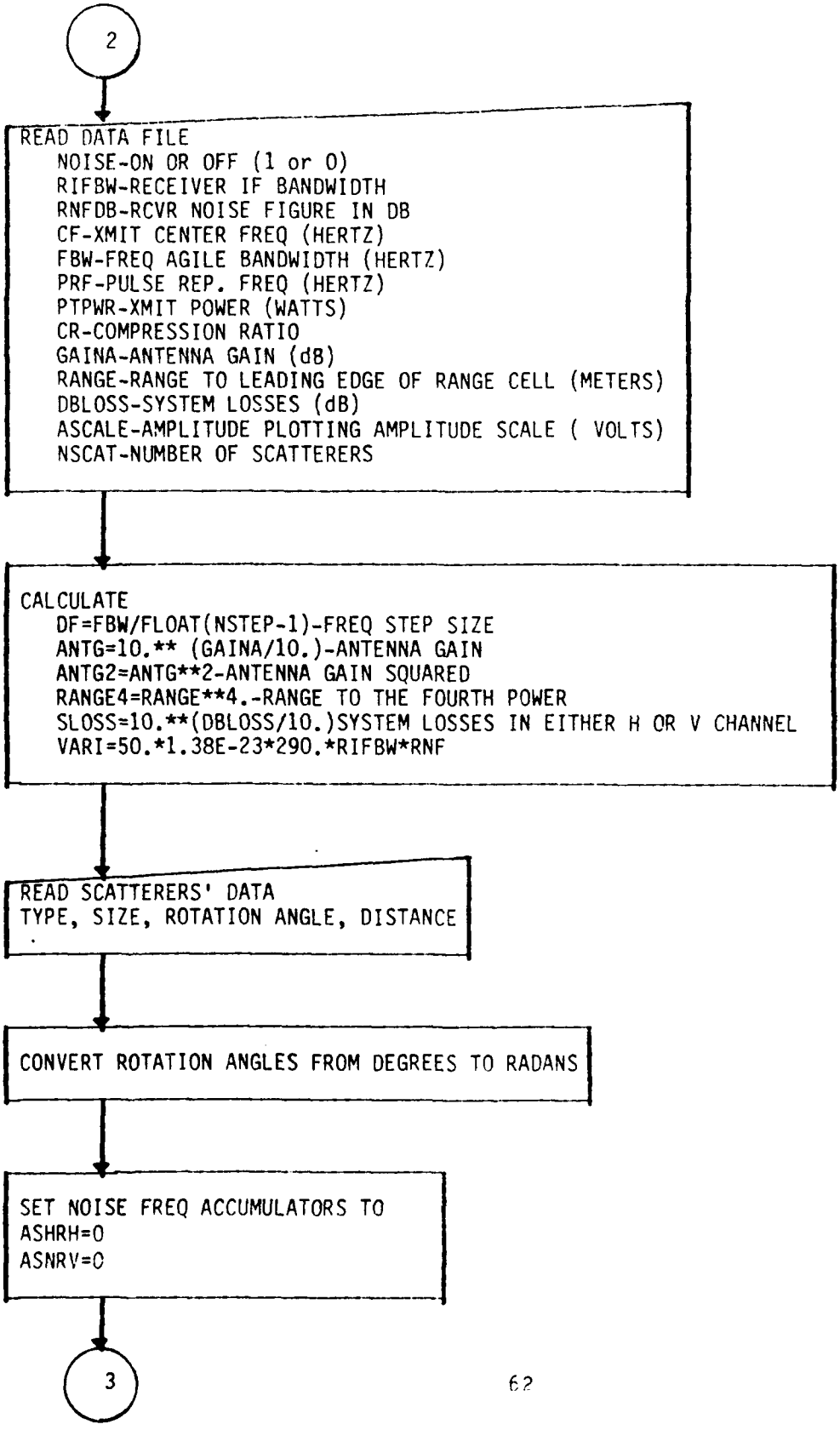
Call HRDCPY - Cause hard copy of 4014 screen to be produced.

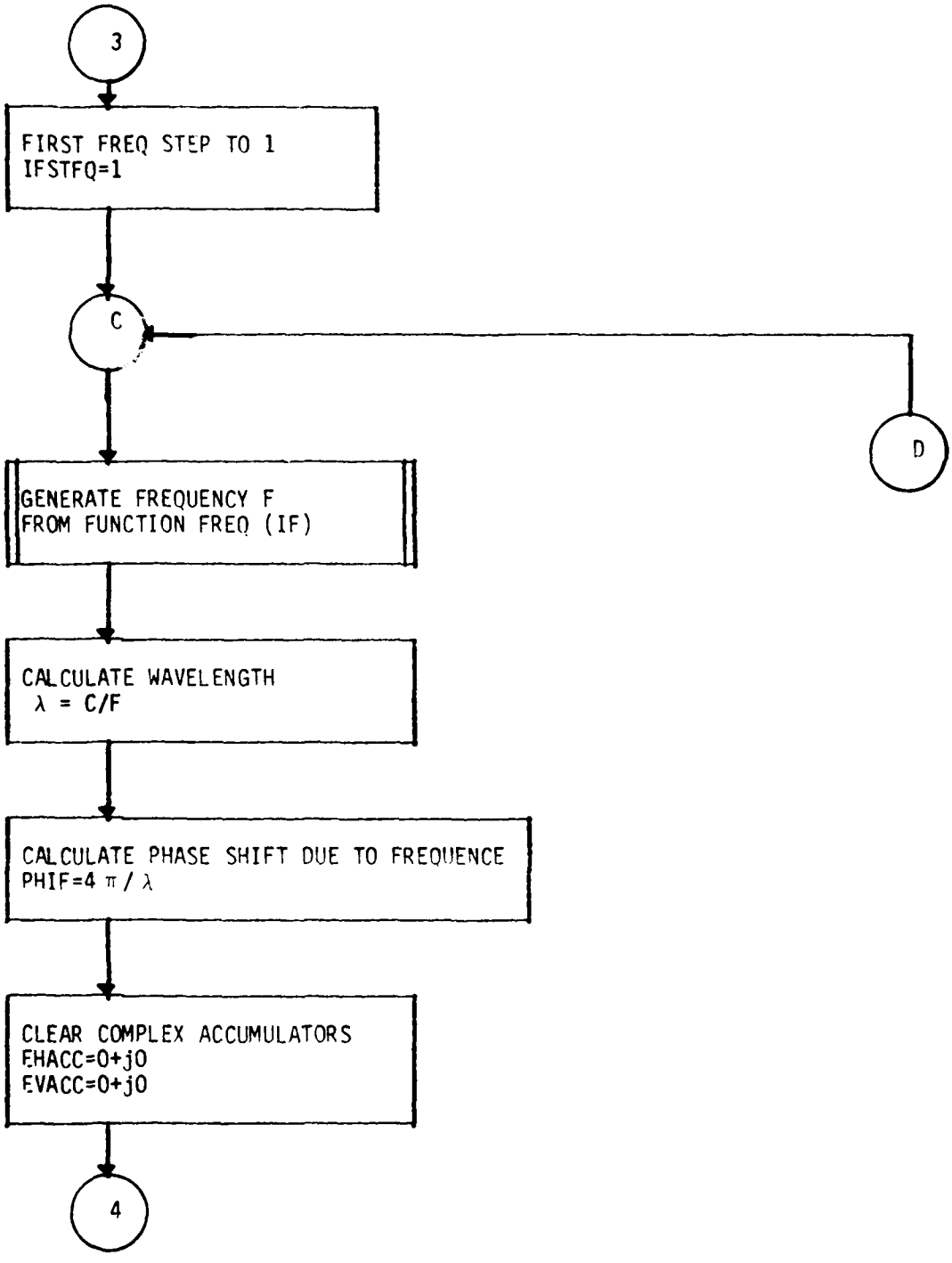
Call LABEL - Provide X and Y axes labels to be centered and typed on 4014.

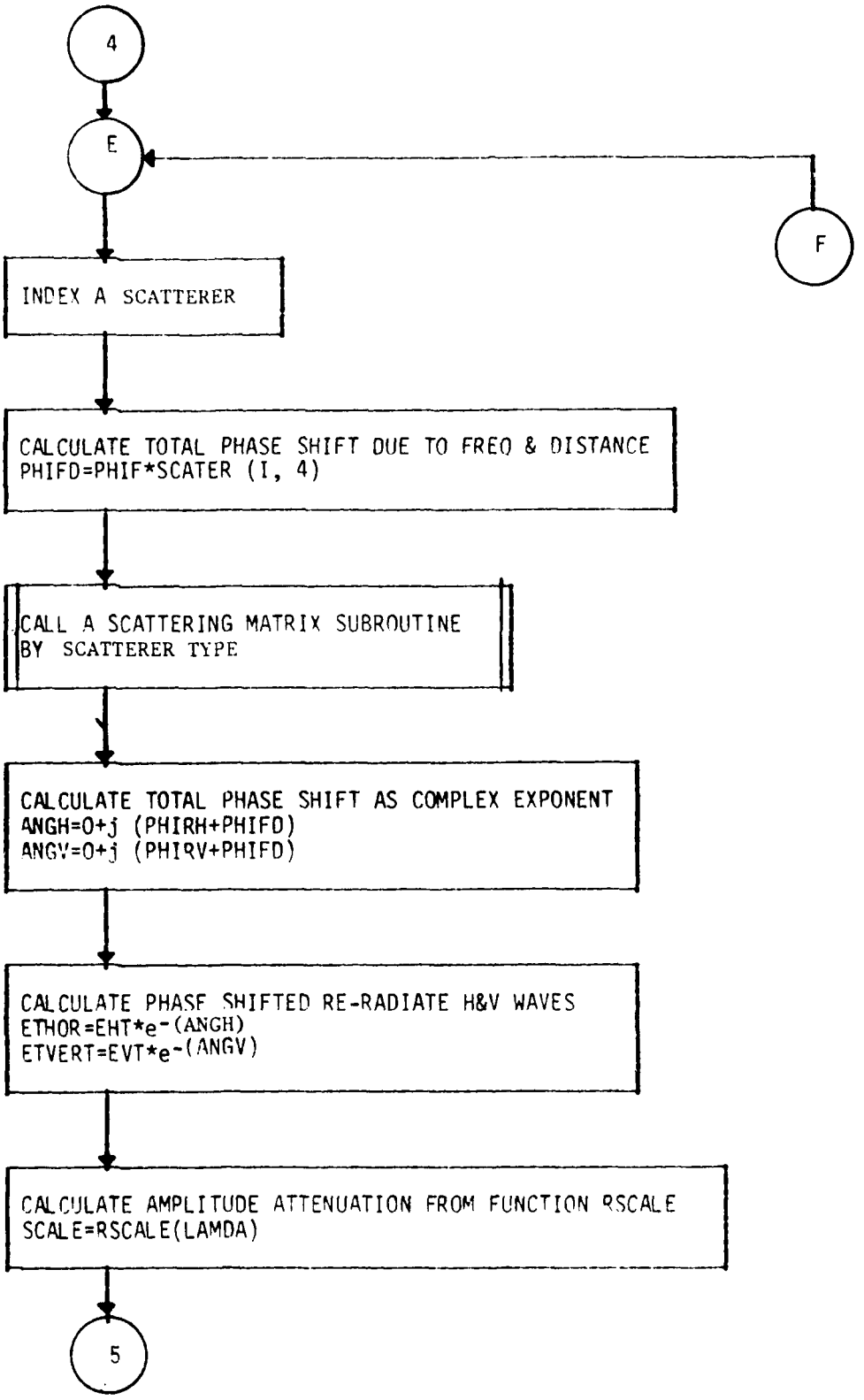
Call STALL - Cause computer to wait momentarily for HRDCPY to be executed.

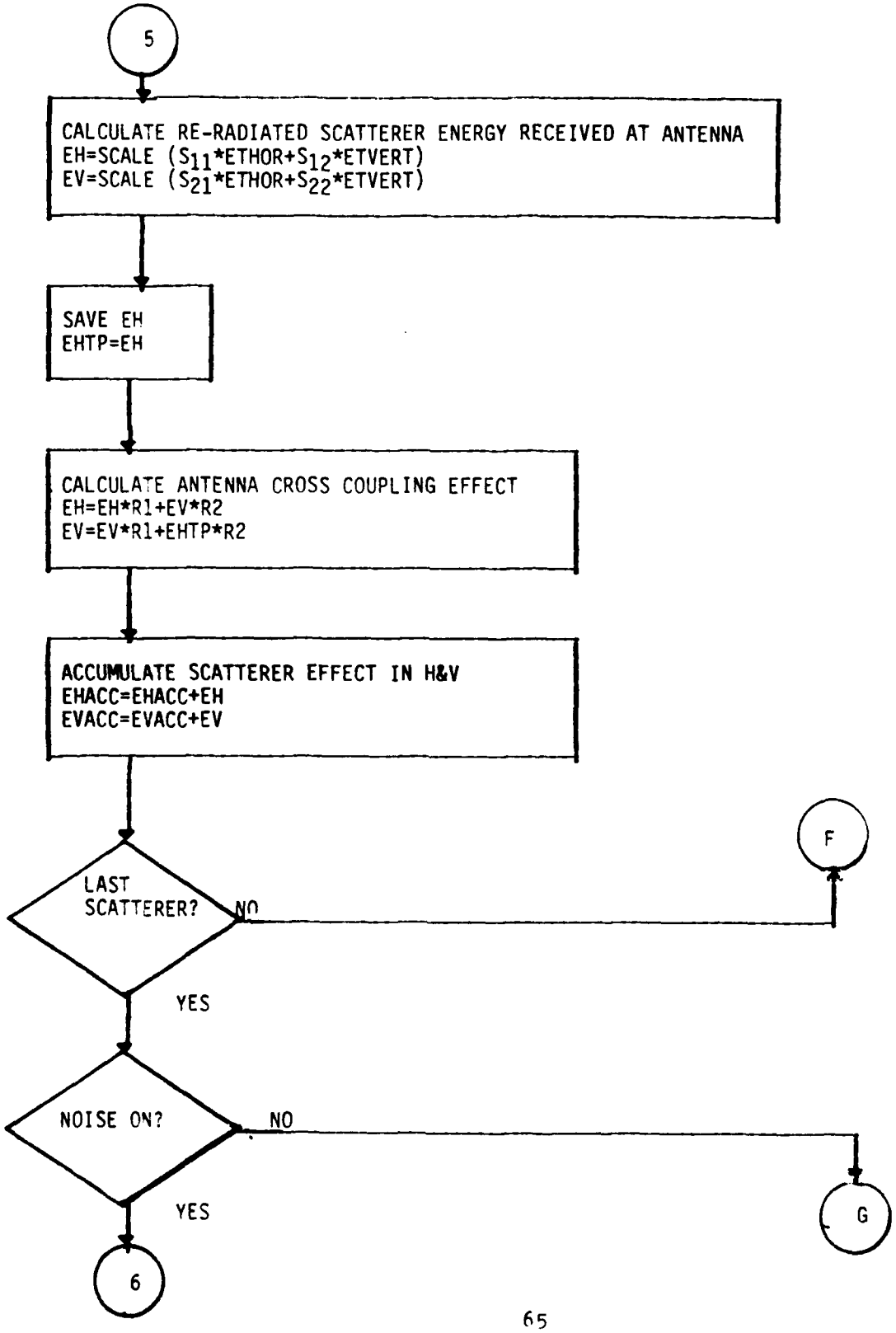

START
INITIALIZE CONSTANTS
π , 2π ,(2π)2,(4π)3 (1/√2), C, CDR, μVOLTS
IRUN=1
A
INCREMENT IRUN
IRUN=IRUN+1
INPUT ANTENNA ISOLATION
AISOL
B
CALL IXMIT SUBROUTINE
GET E1H, E1V, PHITV, NXMIT
R1=SQRT (1.-10.**(-AISOL/10.))
R1=10.**(~AISOL/20)CALCULATE ANTENNA
CROSS-COUPLING RATIOS. R1-COUPLED ENERGY
R1-REMAINING ENERGY
ANGHT=0+j0
ANGVT=0+,jPHITV
1

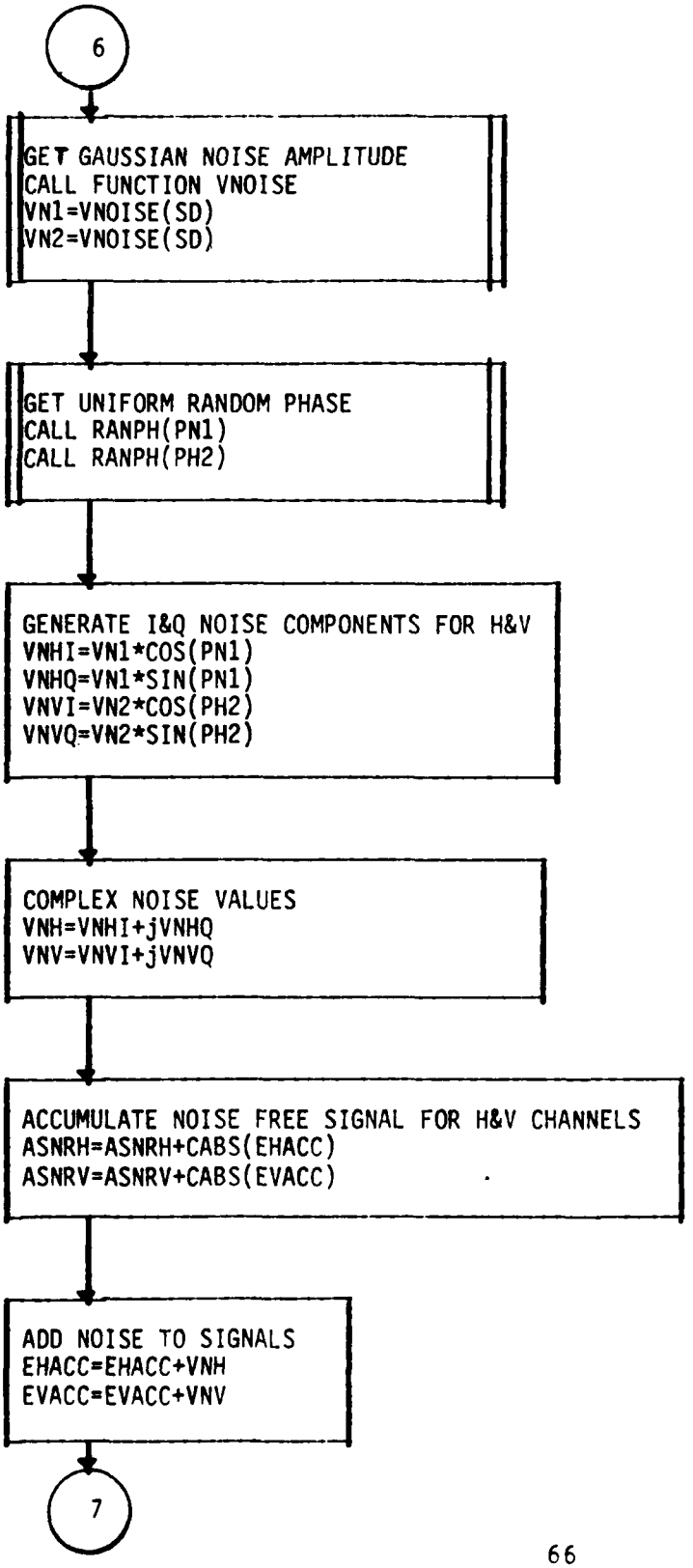

1
EHTP=E1H*e(ANGHT)
EVTP=E1V*e(ANGVT)
MULTIPLY BY CROSS COUPLING FOR TRANSMIT
EHT=EHTP*R1+EVTP*R2
EVT=EVTP*R1+EHTP*R2
PHASE SHIFT IN RCVR
PHIRH=0.
PHIRV=0.
INPUT FREQ
AGILE STEPS
NSTEP
INPUT FFT
SIZE
NFFT
INPUT
DATA FILE NAME
IFILE
2

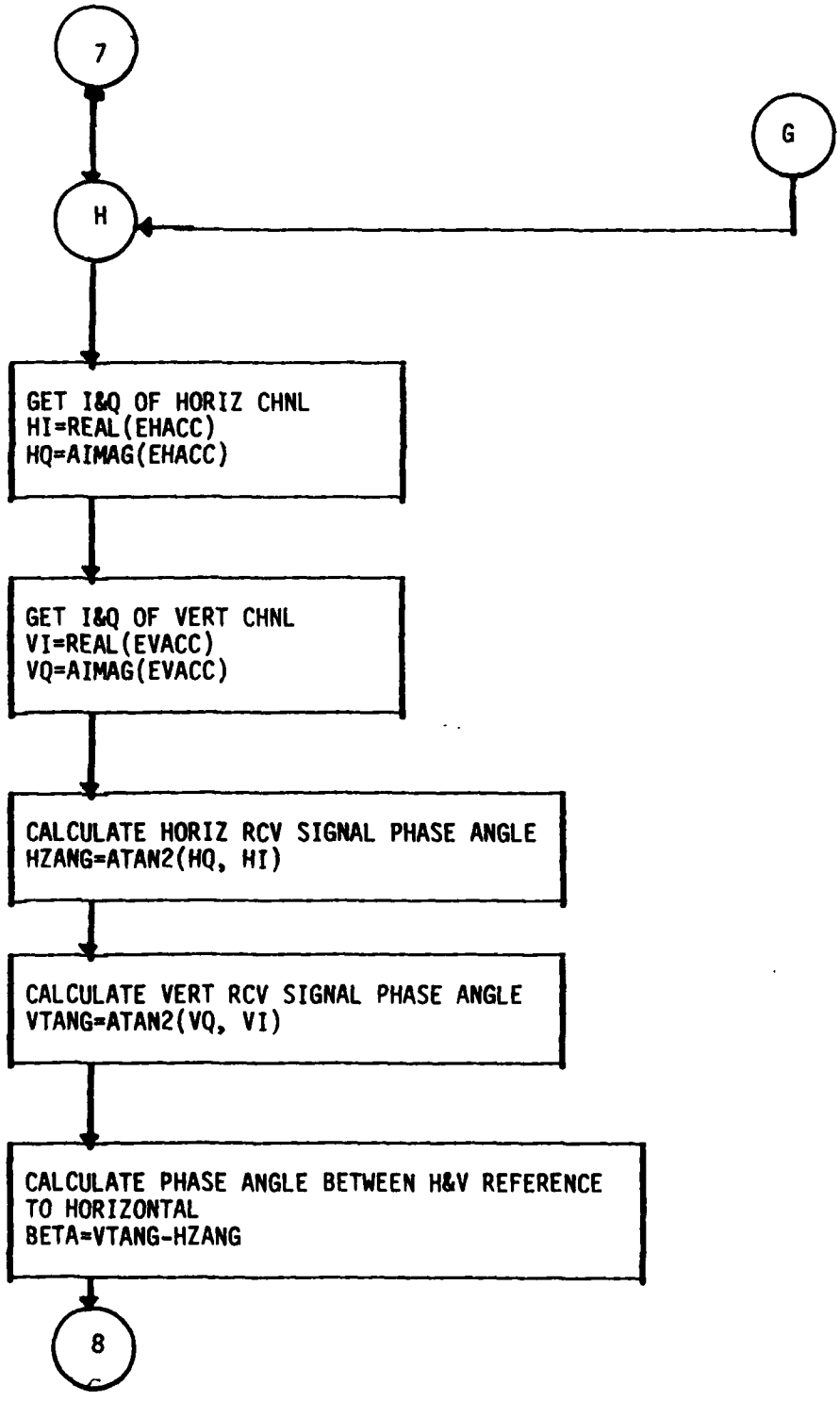

2
READ DATA FILE
NOISE-ON OR OFF (1 or 0)
RIFBW-RECEIVER IF BANDWIDTH
RNFDB-RCVR NOISE FIGURE IN DB
CF-XMIT CENTER FREQ (HERTZ)
FBW-FREQ AGILE BANDWIDTH (HERTZ)
PRF-PULSE REP. FREQ (HERTZ)
PTPWR-XMIT POWER (WATTS)
CR-COMPRESSION RATIO
GAINA-ANTENNA GAIN (dB)
RANGE-RANGE TO LEADING EDGE OF RANGE CELL (METERS)
DBLOSS-SYSTEM LOSSES (dB)
ASCALE-AMPLITUDE PLOTTING AMPLITUDE SCALE (VOLTS)
NSCAT-NUMBER OF SCATTERERS
CALCULATE
DF=FBW/FLOAT(NSTEP-1)-FREQ STEP SIZE
ANTG=10.** (GAINA/10.)-ANTENNA GAIN
ANTG2=ANTG**2-ANTENNA GAIN SQUARED
RANGE4=RANGE**4.-RANGE TO THE FOURTH POWER
SLOSS=10.**(DBLOSS/10.)SYSTEM LOSSES IN EITHER H OR V CHANNEL
VARI=50.*1.38E-23*290.*RIFBW*RNF
READ SCATTERERS' DATA
TYPE, SIZE, ROTATION ANGLE, DISTANCE
CONVERT ROTATION ANGLES FROM DEGREES TO RADANS
SET NOISE FREQ ACCUMULATORS TO
ASHRH=0
ASNRV=0
3

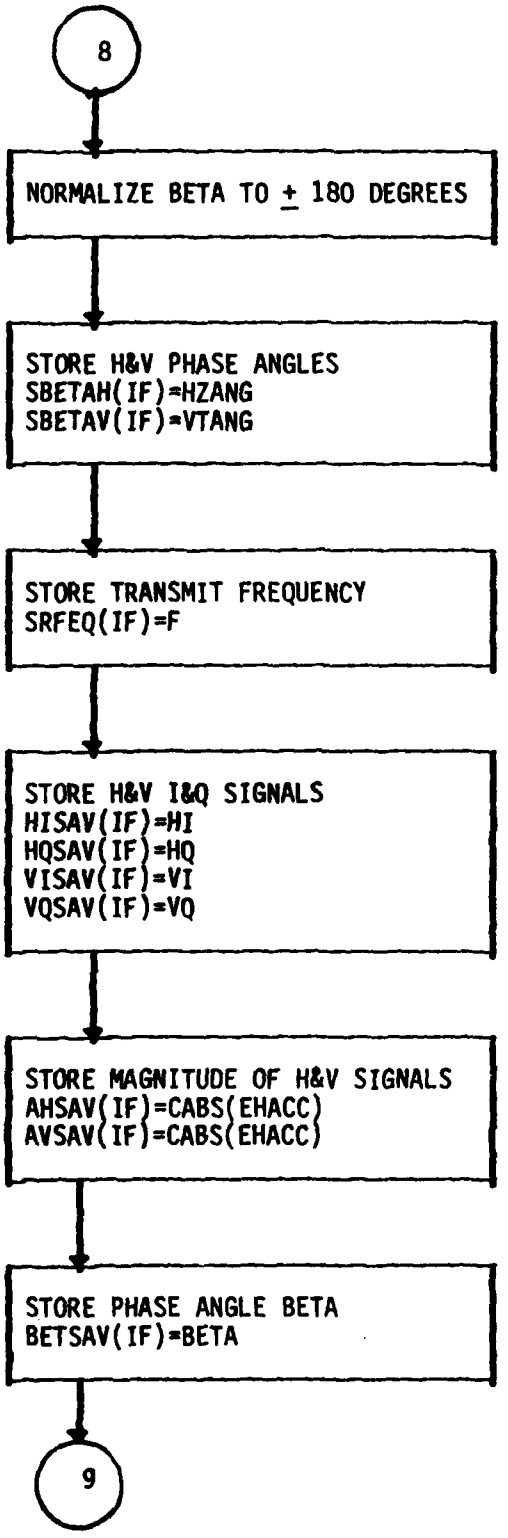

3
FIRST FREQ STEP TO 1
IFSTFQ=1
C
D
GENERATE FREQUENCY F
FROM FUNCTION FREQ (IF)
CALCULATE WAVELENGTH
λ = C/F
CALCULATE PHASE SHIFT DUE TO FREQUENCE
PHIF=4 π / λ
CLEAR COMPLEX ACCUMULATORS
EHACC=0+j0
EVACC=0+j0
4

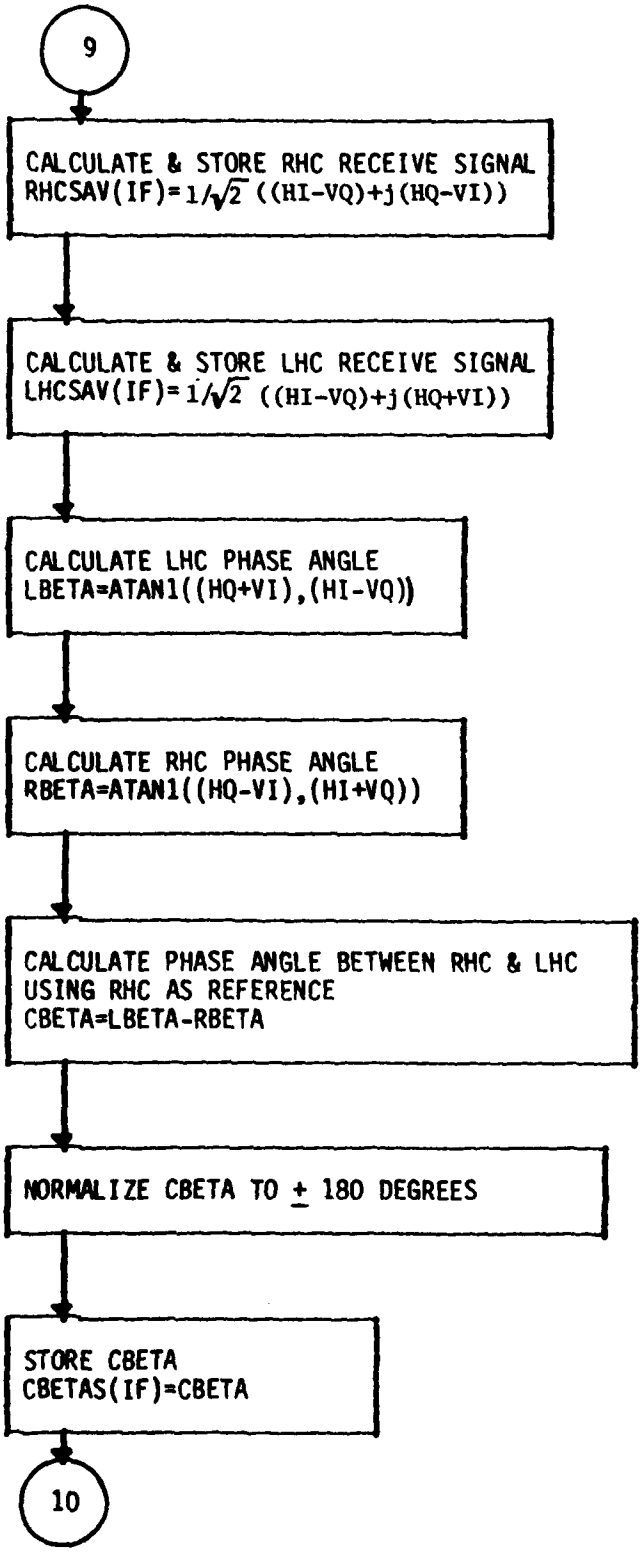

4
E
F
INDEX A SCATTERER
CALCULATE TOTAL PHASE SHIFT DUE TO FREQ & DISTANCE
PHIFD=PHIF*SCATER (I, 4)
CALL A SCATTERING MATRIX SUBROUTINE
BY SCATTERER TYPE
CALCULATE TOTAL PHASE SHIFT AS COMPLEX EXPONENT
ANGH=0+j (PHIRH+PHIFD)
ANGV=0+j (PHIRV+PHIFD)
CALCULATE PHASE SHIFTED RE-RADIATE H&V WAVES
ETHOR=EHT*e-(ANGH)
ETVERT=EVT*e-(ANGV)
CALCULATE AMPLITUDE ATTENUATION FROM FUNCTION RSCALE
SCALE=RSCALE(LAMDA)
5

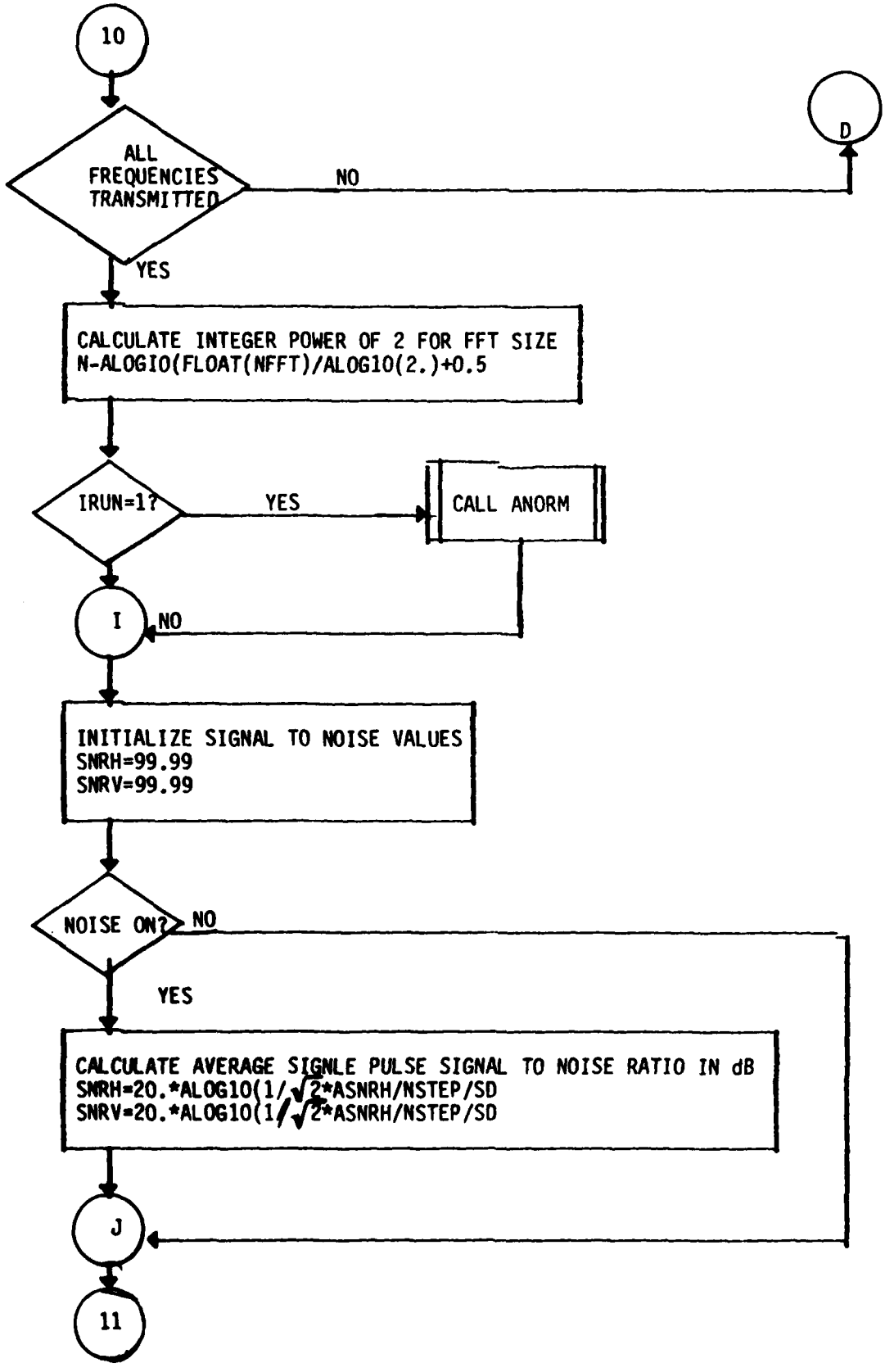

5
CALCULATE RE-RADIATED SCATTERER ENERGY RECEIVED AT ANTENNA
EH=SCALE (S_{11}*ETHOR+S_{12}*ETVERT)
EV=SCALE (S_{21}*ETHOR+S_{22}*ETVERT)
SAVE EH
EHTP=EH
CALCULATE ANTENNA CROSS COUPLING EFFECT
EH=EH*R1+EV*R2
EV=EV*R1+EHTP*R2
ACCUMULATE SCATTERER EFFECT IN H&V
EHACC=EHACC+EH
EVACC=EVACC+EV
F
LAST SCATTERER?
NO
YES
NOISE ON?
NO
YES
G
6

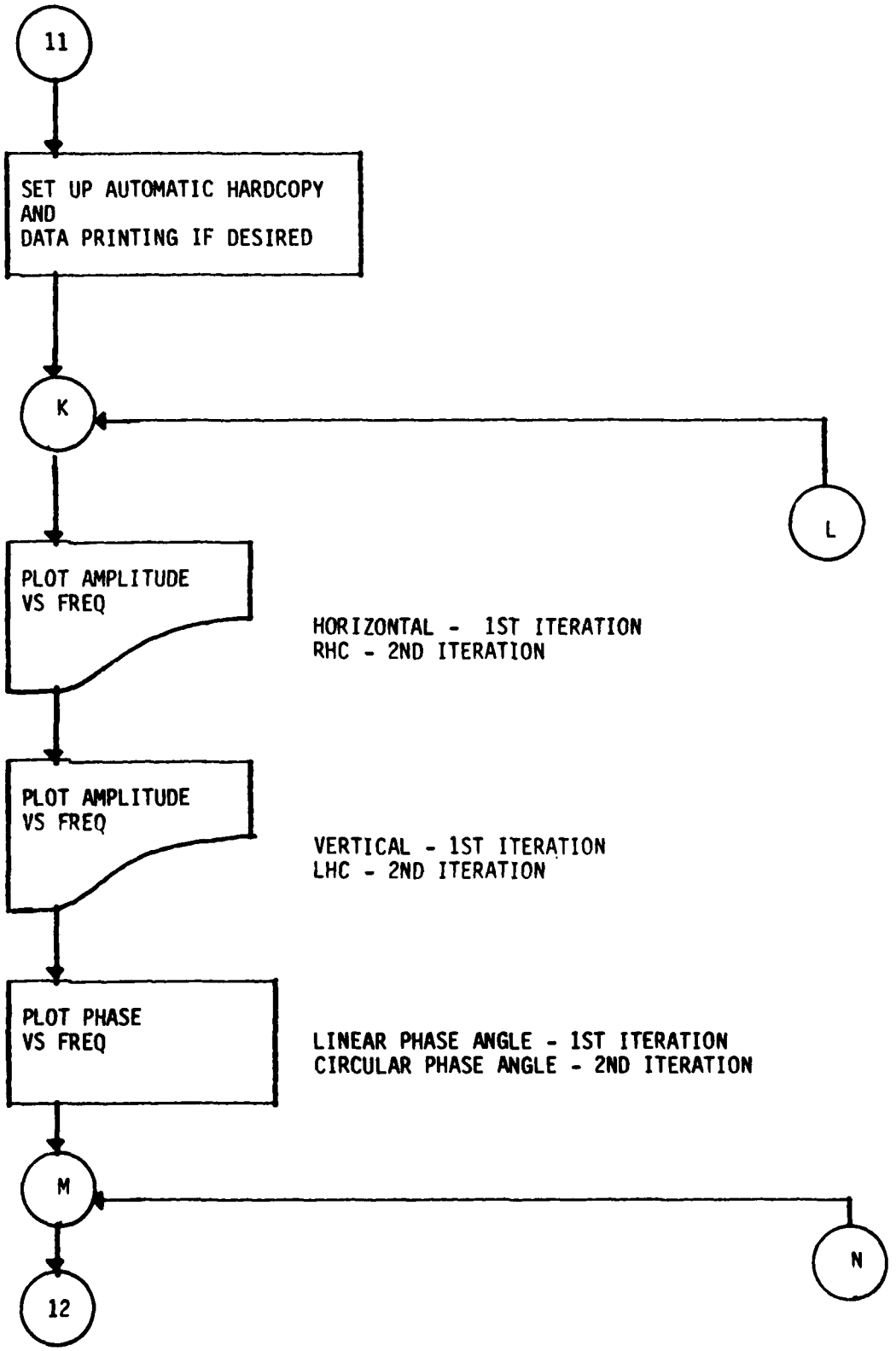

6
GET GAUSSIAN NOISE AMPLITUDE
CALL FUNCTION VNOISE
VN1=VNOISE(SD)
VN2=VNOISE(SD)
GET UNIFORM RANDOM PHASE
CALL RANPH(PN1)
CALL RANPH(PH2)
GENERATE I&Q NOISE COMPONENTS FOR H&V
VNHI=VN1*COS(PN1)
VNHQ=VN1*SIN(PN1)
VNVI=VN2*COS(PH2)
VNVQ=VN2*SIN(PH2)
COMPLEX NOISE VALUES
VNH=VNHI+jVNHQ
VNV=VNVI+jVNVQ
ACCUMULATE NOISE FREE SIGNAL FOR H&V CHANNELS
ASNRH=ASNRH+CABS(EHACC)
ASNRV=ASNRV+CABS(EVACC)
ADD NOISE TO SIGNALS
EHACC=EHACC+VNH
EVACC=EVACC+VNV
7

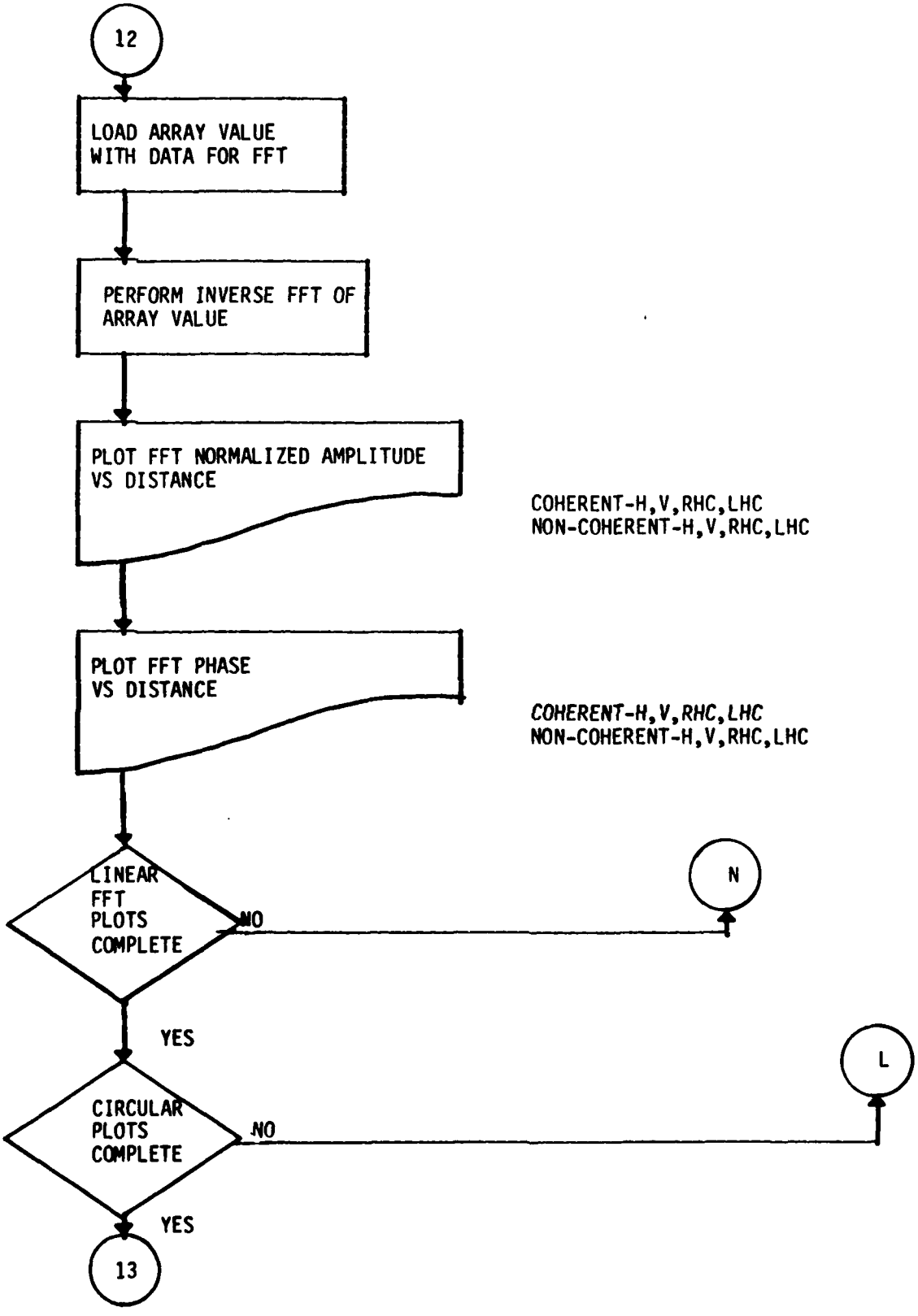

7
G
H
GET I&Q OF HORIZ CHNL
HI=REAL(EHACC)
HQ=AIMAG(EHACC)
GET I&Q OF VERT CHNL
VI=REAL(EVACC)
VQ=AIMAG(EVACC)
CALCULATE HORIZ RCV SIGNAL PHASE ANGLE
HZANG=ATAN2(HQ, HI)
CALCULATE VERT RCV SIGNAL PHASE ANGLE
VTANG=ATAN2(VQ, VI)
CALCULATE PHASE ANGLE BETWEEN H&V REFERENCE
TO HORIZONTAL
BETA=VTANG-HZANG
8

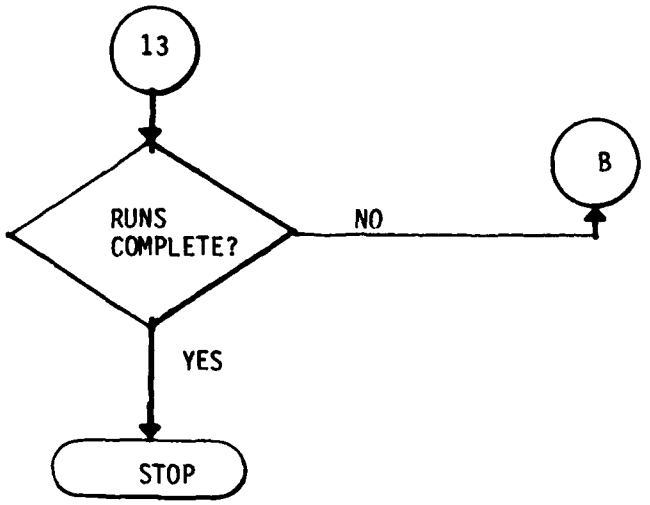

8
NORMALIZE BETA TO ± 180 DEGREES
STORE H&V PHASE ANGLES
SBETAH(IF)=HZANG
SBETAV(IF)=VTANG
STORE TRANSMIT FREQUENCY
SRFEQ(IF)=F
STORE H&V I&Q SIGNALS
HISAV(IF)=HI
HQSAV(IF)=HQ
VISAV(IF)=VI
VQSAV(IF)=VQ
STORE MAGNITUDE OF H&V SIGNALS
AHSAV(IF)=CABS(EHACC)
AVSAV(IF)=CABS(EHACC)
STORE PHASE ANGLE BETA
BETSAV(IF)=BETA
9


9
CALCULATE & STORE RHC RECEIVE SIGNAL
RHCSAV(IF)= 1/√2 ((HI-VQ)+j(HQ-VI))
CALCULATE & STORE LHC RECEIVE SIGNAL
LHCSAV(IF)= 1/√2 ((HI-VQ)+j(HQ+VI))
CALCULATE LHC PHASE ANGLE
LBETA=ATAN1((HQ+VI),(HI-VQ))
CALCULATE RHC PHASE ANGLE
RBETA=ATAN1((HQ-VI),(HI+VQ))
CALCULATE PHASE ANGLE BETWEEN RHC & LHC
USING RHC AS REFERENCE
CBETA=LBETA-RBETA
NORMALIZE CBETA TO ± 180 DEGREES
STORE CBETA
CBETAS(IF)=CBETA
10


10
ALL FREQUENCIES TRANSMITTED
NO
D
YES
CALCULATE INTEGER POWER OF 2 FOR FFT SIZE
N-ALOG10(FLOAT(NFFT)/ALOG10(2.)+0.5
IRUN=1?
YES
CALL ANORM
I
NO
INITIALIZE SIGNAL TO NOISE VALUES
SNRH=99.99
SNRV=99.99
NOISE ON?
NO
YES
CALCULATE AVERAGE SIGNLE PULSE SIGNAL TO NOISE RATIO IN dB
SNRH=20.*ALOG10(1/√2*ASNRH/NSTEP/SD
SNRV=20.*ALOG10(1/√2*ASNRH/NSTEP/SD
J
11


11
SET UP AUTOMATIC HARDCOPY
AND
DATA PRINTING IF DESIRED
K
L
PLOT AMPLITUDE
VS FREQ
HORIZONTAL - 1ST ITERATION
RHC - 2ND ITERATION
PLOT AMPLITUDE
VS FREQ
VERTICAL - 1ST ITERATION
LHC - 2ND ITERATION
PLOT PHASE
VS FREQ
LINEAR PHASE ANGLE - 1ST ITERATION
CIRCULAR PHASE ANGLE - 2ND ITERATION
M
N
12


12
LOAD ARRAY VALUE WITH DATA FOR FFT
PERFORM INVERSE FFT OF ARRAY VALUE
PLOT FFT NORMALIZED AMPLITUDE VS DISTANCE
COHERENT-H,V,RHC,LHC
NON-COHERENT-H,V,RHC,LHC
PLOT FFT PHASE VS DISTANCE
COHERENT-H,V,RHC,LHC
NON-COHERENT-H,V,RHC,LHC
LINEAR FFT PLOTS COMPLETE
NO
N
YES
CIRCULAR PLOTS COMPLETE
NO
L
YES
13


13
B
RUNS
COMPLETE?
NO
YES
STOP

```
0001          PROGRAM SIMPTS
      C
      C   PROGRAM TO SIMULATE THE RF GUIDANCE TECHNOLOGY'S POLARIMETRIC
      C     TECHNOLOGY SEEKER (PTS).
      C
      C     WRITTEN BY: F. W. SEVENQUIST AND R. F. RUSSELL  4-FEB-82
      C
      C     LATEST UPDATE: 28-OCT-82
      C
0002          INTEGER NXAXIS(20),NYAXIS(20)
0003          INTEGER IFILE(8),NXMIT(2)
0004          REAL LOBETA,LAMDA
0005          COMPLEX VALUE(256)
0006          COMPLEX AVGVAL
0007          DIMENSION SFREQ(256)
0008          DIMENSION SBETAM(256),SBETAV(256)
0009          DIMENSION SCATER(100,4),AMSAV(256),AVSAV(256),BETSAV(256)
0010          DIMENSION HMSAV(256),HISAV(256),VMSAV(256),VISAV(256)
0011          DIMENSION CBETAS(256)
0012          DIMENSION A(20)
0013          COMPLEX RHCSAV(256),LHCSAV(256)
0014          COMPLEX SMATRX(2,2),EH,EV,ETHOR,ETVERT,ANGH,ANGV,EHTP,EVTP,EHT,EVT
0015          COMPLEX ANGHT,ANGVT
0016          COMPLEX EHACC,EVACC,VNH,VNV
0017          COMMON /WORKF/IFSTFQ,IUP,LSTEP,NSTEP,DF,CF,FBW
0018          COMMON /MKSCT/ SCATER,SMATRX
0019          COMMON /HEAD/AISOL,NSCAT,GAINA,NOISE,
             1    RANGE,OBLOSS,NXMIT,IFILE,SD,BIG,
             1    SNRH,SNRV,SNRHI,SNRHQ,SNRVI,SNRVQ,SNR
0020          COMMON /WORK/HMSAV,HISAV,VMSAV,VISAV,SFREQ,SBETAM,SBETAV,BETSAV,
             1             RHCSAV,LHCSAV,CBETAS,AMSAV,AVSAV,
             1             VALUE
0021          COMMON /SIGVAL/PTPAR,RANGE4,CK,ANTG2,SLOSS,PI4C
      C
      C  INITIALIZATION VALUES
0022          V1OSR2=1./SQRT(2.)
0023          UVOLTS=1.E-6          !MICRO-VOLTS SCALER
0024          C=2.99793E8
0025          ICPY=0
0026          IRUN=1
0027          PI=3.14159
0028          PI2=2.*PI
0029          PI4=2.*PI2
0030          PI4C=PI4**3.
0031          CON=PI/180.  !CONVERT DEGREES TO RADIANS
      C
0032          CALL PLOT(0)
0033          CALL V14CS2(1)
0034          TYPE 6005
0035          ACCEPT 6004,AISOL
0036          CALL XMIT(EIH,EIV,PHITV,NXMIT)
0037          R1=SQRT(1.-10.**(-AISOL/10.))        !REMAINING VOLTAGE RATIO
0038          R2=10.**(-AISOL/20.)          !TRANSFERED VOLTAGE RATIO
      C       NORMIZ TRANSMIT POWE
```

```
0039          ANGHT=CMPLX(0.,0.)
0040          ANGVT=CMPLX(0.,PHITV)
0041          EHTP=E1H*CEXP(ANGHT)
0042          EVTP=E1V*CEXP(ANGVT)
      C       HORIZONTAL TRANSMIT COMPONENT WITH ANTENNA X-COUPLING
0043          EHT=EHTP*R1+EVTP*R2
      C       VERTICAL TRANSMIT COMPONENT WITH ANTENNA X-COUPLING
0044          EVT=EVTP*R1+EHTP*R2
0045          PHIRH=0.      !PHASE SHIFT TO RECEIVED HOR SIG
0046          PHIRV=0.      !PHASE SHIFT TO RECEIVED VERT SIG
0047          TYPE 6003
0048          ACCEPT 6002,NSTEP
0049          TYPE 6001
0050          ACCEPT 6002,NFFT
0051    11111 CALL PLOT(0)
0052          TYPE 6000,IRUN
      C
      C   *** ALL INPUT DATA IS READ FROM FILE INPUT TO IFILE
      C
0053          ACCEPT 6012,IFILE
0054          CALL ASSIGN (22,IFILE,0,'ROO')
0055          READ(22,6002)NOISE          !ENTER 0 FOR NOISE OFF, 1 FOR NOISE ON
0056          READ(22,6004)RIFBW  !RECEIVER IF BANDWIDTH IN HERTZ
0057          READ(22,6004)RNFDB          !RECEIVER NOISE FIGURE IN DB
0058          RNF=10.**(RNFDB/20.)        !RECEIVER NOISE FIGURE
      C   VARIANCE = RKTBNF
0059          VARI=(50.*1.38E-23*290.*RIFBW*RNF)  !50 OHM IMPEDANCE
0060          SD=SQRT(VARI)       !STANDARD DEVIATION = SQRT(VARIANCE)
0061          READ(22,6004)CF             !TRANSMITTER CENTER FREQUENCY
0062          READ(22,6004)FBW            !FREQUENCY AGILITY BANDWIDTH
0063          READ(22,6004)PRF            !TRANSMIT PULSE REP FREQ
0064          DF=FBW/FLOAT(NSTEP-1)
0065          READ(22,6004)PTPAV  !AVERAGE XMIT PWR HORIZ OR VERT CHANNEL
      C        WHEN XMITTER TURNED ON (I.E. XMIT POWER/2)
0066          READ(22,6004)CR             !COMPRESSION RATIO
0067          READ(22,6004)GAINA  !ANTENNA GAIN IN DB
0068          ANTG=10.**(GAINA/10.)       !ANTENNA GAIN
0069          ANTG2=ANTG**2.
0070          READ(22,6004)RANGE  !RANGE IN METERS TO CELL OF INTEREST
0071          RANGE4=RANGE**4.
0072          READ(22,6004)DBLOSS !SYSTEM LOSSES IN DB FOR EITHER H OR V CHANNEL
0073          SLOSS=10.**(DBLOSS/10.)     !SYSTEM LOSSES EITHER CHANNEL H OR V
0074          READ (22,6004)ASCALE        !AMPLITUDE SCALE MAX SCALE
      C
      C       THE NEXT LINE MUST BE NUMBER OF SCATTERS TO BE READ
      C        FROM INPUT FILE
      C
      C     THEN EACH SUCCEEDING LINE WILL CHARACTERIZE THE SCATTERERS AS FOLLOWS:
      C
      C
      C   SCATER(I,1)=TYPE
      C    ENTER 1 FOR FLAT PLATE
      C          2 FOR DIHEDRAL
      C          3 FOR TRIHEDRAL
```

```
C          4 FOR DIPOLE
C
C  SCATER(I,2)= SIZE (SQ METERS)
C  SCATER(I,3) = ORIENTATION ANGLE IN DEGREES
C  SCATER(I,4)=ONE WAY DISTANCE FROM LEADING EDGE OF RANGE CELL (METERS)
C
0075          READ (22,6002)NSCAT
0076          CALL PLOT(0)
0077          CALL V14CSZ(4)
0078          TYPE 6010,IFILE,ASCALE,NSCAT
0079          DO 50 I=1,NSCAT
0080          READ (22,6011)(SCATER(I,K),K=1,4)
0081    50    TYPE 6009,(SCATER(I,K),K=1,4)
0082          CALL V14CSZ(1)
0083          CALL CLOSE(22)
0084          DO 60 I=1,NSCAT
0085    60    SCATER(I,3)=SCATER(I,3)*CDR
0086          IFSTEP=1      !START FREQ @ STEP 1 OF UP RAMP
0087          ASNRH=0.      !INITIATE ACCUMULATOR FOR NOISE-FREE H CHANNEL SIGNAL
0088          ASNRV=0.
0089          DO 200 IF=1,NSTEP
0090          F=FREQ(IF)
0091          LAMDA=C/F
0092          PHIF=PI4/LAMDA
0093          EHACC=CMPLX(0.,0.)  !INITAITE H ACCUMULATOR
0094          EVACC=CMPLX(0.,0.)  !INITIATE V ACCUMULATOR
0095          DO 100 I=1,NSCAT
0096          PHIFD=PHIF*SCATER(I,4)
0097          CALL GETSM(I)
0098          ANGH=CMPLX(0.,PHIRH+PHIFD)  !EFFECTIVE HOR PHASE SHIFT
0099          ANGV=CMPLX(0.,PHIRV+PHIFD)  !EFFECTIVE VERT PHASE SHIFT
0100          ETHOR=EHT*CEXP(-ANGH)
0101          ETVERT=EVT*CEXP(-ANGV)
0102          SCALE=RSCALE(LAMDA)
0103          EH=SCALE*(SMATRX(1,1)*ETHOR+SMATRX(1,2)*ETVERT)
0104          EV=SCALE*(SMATRX(2,1)*ETHOR+SMATRX(2,2)*ETVERT)
0105          EHTP=EH      !SAVE PURE RECEIVE HOR SIGNAL
0106          EH=EH*R1+EV*R2
0107          EV=EV*R1+EHTP*R2     !USE PURE HOR SIG
0108          EHACC=EHACC+EH
0109   100    EVACC=EVACC+EV
0110          IF(NOISE.EQ.0)GOTO 120
0112          VN1=VNOISE(SD)
0113          CALL RANPH(PH1)
0114          VN2=VNOISE(SD)
0115          CALL RANPH(PH2)
0116          VNHI=VN1*COS(PH1)
0117          VNHQ=VN1*SIN(PH1)
0118          VNVI=VN2*COS(PH2)
0119          VNVQ=VN2*SIN(PH2)
0120          VNH=CMPLX(VNHI,VNHQ)
0121          VNV=CMPLX(VNVI,VNVQ)
0122          ASNRH=ASNRH+CABS(EHACC)              !ACCUMULATE H NOISE-FREE SIGNAL
0123          ASNRV=ASNRV+CABS(EVACC)              !ACCUMULATE V NOISE-FREE SIGNAL
```

```
0124          EMACC=EMACC+VNM              !ADD NOISE TO M SIGNAL
0125          EVACC=EVACC+VNV              !ADD NOISE TO V SIGNAL
0126    120   CONTINUE
0127          MI=REAL(EMACC)
0128          MQ=AIMAG(EMACC)
0129          VI=REAL(EVACC)
0130          VQ=AIMAG(EVACC)
0131          MZANG=0.
0132          VTANG=0.
0133          IF(MI.EQ.0.0.AND.MQ.EQ.0.0)GOTO 130
0135          MZANG=ATAN2(MQ,MI)
0136    130   IF(VI.EQ.0.0.AND.VQ.EQ.0.0)GOTO 140
0138          VTANG=ATAN2(VQ,VI)
0139    140   DETA=VTANG-MZANG
      C
      C  BY DEFINITION: DETA IS ZERO(0) IF EITHER HORIZONTAL
      C     OR VERTICAL ANGLE IS ZERO.
      C
0140          IF(MZANG.EQ.0.0.OR.VTANG.EQ.0.0)DETA=0.
0142          DETA=AMOD(DETA,PI2)
0143          IF(DETA.GT.PI)DETA=DETA-PI2
0145          IF(DETA.LT.-PI)DETA=PI2+DETA
0147          SDETAM(IF)=MZANG
0148          SDETAV(IF)=VTANG
0149          SFREQ(IF)=F
0150          MQSAV(IF)=MQ
0151          MISAV(IF)=MI
0152          VQSAV(IF)=VQ
0153          VISAV(IF)=VI
0154          AMSAV(IF)=CABS(EMACC)        !CALCULATING PEAK HORIZ AMPLITUDE
0155          AVSAV(IF)=CABS(EVACC)        !CALCULATING PEAK VERT AMPLITUDE
0156          DETSAV(IF)=DETA
0157          RMCSAV(IF)=V10SR2*CMPLX((MI+VQ),(MQ-VI))
0158          LMCSAV(IF)=V10SR2*CMPLX((MI-VQ),(MQ+VI))
0159          LDETA=ATAN2((MQ+VI),(MI-VQ))
0160          RDETA=ATAN2((MQ-VI),(MI+VQ))
0161          CDETA=LDETA-RDETA
0162          CDETA=AMOD(CDETA,PI2)
0163          IF(CDETA.GT.PI)CDETA=CDETA-PI
0165          IF(CDETA.LT.-PI)CDETA=CDETA+PI
0167          CDETAS(IF)=CDETA
0168    200   CONTINUE
0169          N=ALOG10(FLOAT(NFFT))/ALOG10(2.)+0.5
0170          IF(IRUN.EQ.1)CALL ANORM(N,NFFT,NSTEP,BIG)
0172          SNRH=99.99   !VALUE IF NOISE IS TURNED OFF
0173          SNRV=99.99
0174          IF(NOISE.EQ.0)GOTO 210
      C
      C  CALCULATE AVERAGE RMS NOISE-FREE SINGLE PULSE SNR FOR EACH CHANNEL
      C
0176          SNRH=20.*ALOG10(V10SR2*ASNRH/(FLOAT(NSTEP)*SD))
0177          SNRV=20.*ALOG10(V10SR2*ASNRV/(FLOAT(NSTEP)*SD))
      C
0178    210   CALL SNR(ISW)
```

```
0179          IF(IAND(ISW,2).EQ.2)ICPY=1
0181          IF(IAND(ISW,1).EQ.0)GOTO 250
0183          CALL ASSIGN (6,'LP: ',0)
0184          WRITE (6,6006)
0185          DO 220 I=1,NSTEP
0186    220   WRITE (6,6007)SFREQ(I),HISAV(I),HQSAV(I),VISAV(I),VQSAV(I),
             1    AMSAV(I),AVSAV(I),dETSAV(I),SdETAM(I),SBETAV(I)
0187          CALL CLOSE (6)
0188    250   CALL ASSIGN(22,'DK:PTSSIM.NAM ',0,'RUO')
0189          ICIR=0
0190    300   CALL PLOT(0)
0191          IFIRST=1
0192          XMIN=(CF-FBW/2.)/1.E9
0193          XMAX=(CF+FBW/2.)/1.E9
0194          YMIN=0.
0195          YMAX=ASCALE
0196          CALL AXES(A,XMIN,XMAX,(XMAX-XMIN)/50.,YMIN,YMAX,
             1    (YMAX-YMIN)/50.,XMIN,YMIN,840.,600.,150.,100.)
0197          READ(22,6006)NXAXIS
0198          READ(22,6006)NYAXIS
0199          CALL LABEL(A,NXAXIS,NYAXIS)
0200          CALL HEADER
      C
      C  PLOTTING HORIZ OR RHC AMPLITUDE
      C
0201          DO 400 IF=1,NSTEP
0202          X=SFREQ(IF)/1.E9
0203          IF(ICIR.EQ.0)Y=AMSAV(IF)     !GETTING PEAK HORIZ AMPLITUDE
0205          IF(ICIR.EQ.1)Y=CABS(RHCSAV(IF))     !GETTING PEAK RHC AMPLITUDE
0207          IF(IFIRST.EQ.1)CALL LINE(A,X,Y/UVOLTS,0)
0209          IFIRST=0
0210    400   CALL LINE(A,X,Y/UVOLTS,1)
0211          IF(ICPY.EQ.0) GOTO 410
0213          CALL HRDCPY
0214          CALL STALL
0215          GOTO 420
0216    410   ACCEPT 6006,IANS
0217    420   CALL PLOT(0)
0218          IFIRST=1
0219          XMIN=(CF-FBW/2.)/1.E9
0220          XMAX=(CF+FBW/2.)/1.E9
0221          YMIN=0.
0222          YMAX=ASCALE
0223          CALL AXES(A,XMIN,XMAX,(XMAX-XMIN)/50.,YMIN,YMAX,
             1    (YMAX-YMIN)/50.,XMIN,YMIN,840.,600.,150.,100.)
0224          READ(22,6006)NXAXIS
0225          READ(22,6006)NYAXIS
0226          CALL LABEL(A,NXAXIS,NYAXIS)
0227          CALL HEADER
      C
      C  PLOTTING VERT OR LHC AMPLITUDE
      C
0228          DO 500 IF=1,NSTEP
0229          X=SFREQ(IF)/1.E9
```

```
0250          IF(ICIN.EQ.0)Y=AVSAV(IF)     !GETTING PEAK VERT AMPLITUDE
0252          IF(ICIN.EQ.1)Y=CABS(LHCSAV(IF))      !LOADING PEAK LHC AMPLITUDE
0254          IF(IFIRST.EQ.1)CALL LINE(A,X,Y/UVOLTS,0)
0256          IFIRST=0
0257   500    CALL LINE(A,X,Y/UVOLTS,1)
0258          IF(ICPY.EQ.0) GOTO 510
0260          CALL HRDCPY
0261          CALL STALL
0262          GOTO 520
0263   510    ACCEPT 6006,IANS
0264   520    IFIRST=1
0265          XMIN=(CF-FBW/2.)/1.E9
0266          XMAX=(CF+FBW/2.)/1.E9
0267          YMIN=-180.
0268          YMAX=180.
0269          CALL PLOT(0)
0270          CALL AXES(A,XMIN,XMAX,(XMAX-XMIN)/50.,YMIN,YMAX,
             1    (YMAX-YMIN)/50.,XMIN,YMIN,840.,600.,150.,100.)
0271          READ(22,6006)NXAXIS
0272          READ(22,6006)NYAXIS
0273          CALL LABEL(A,NXAXIS,NYAXIS)
0274          CALL HEADER
      C
      C  PLOTTING PHASE BETWEEN H & V OR RHC & LHC
      C
0275          DO 600 IF=1,NSTEP
0276          X=SFREQ(IF)/1.E9
0277          IF(ICIN.EQ.0)Y=HETSAV(IF)*180./PI
0279          IF(ICIN.EQ.1)Y=COETAS(IF)*180./PI
0281          IF(IFIRST.EQ.1)CALL LINE(A,X,Y,0)
0283          IFIRST=0
0284   600    CALL LINE(A,X,Y,1)
0285          IF(ICPY.EQ.0) GOTO 610
0287          CALL HRDCPY
0288          CALL STALL
0289          GOTO 620
0290   610    ACCEPT 6006,IANS
0291   620    CONTINUE
0292          DO 1500 IFFT=1,2
0293          DO 1010 I=1,NFFT
0294   1010   VALUE(I)=CMPLX(0.,0.)        !ZERO OUT COMPLEX BUFFER VALUE
0295          DO 1020 I=1,NSTEP
0296          IF(ICIN.EQ.0.AND.IFFT.EQ.1)VALUE(I)=
             1     CMPLX(HISAV(I),HQSAV(I))          !LOAD BUFFER WITH HORIZ I&Q
0298          IF(ICIN.EQ.0.AND.IFFT.EQ.2)VALUE(I)=
             1    CMPLX(AMSAV(I),0.)          !LOAD BUFFER REAL PART WITH HORIZ AMP
0300          IF(ICIN.EQ.1.AND.IFFT.EQ.1)VALUE(I)=RHCSAV(I)        !LOAD BUFFER WITH RH
0302          IF(ICIN.EQ.1.AND.IFFT.EQ.2)VALUE(I)=
             1    CMPLX(CABS(RHCSAV(I)),0.)          !LOAD BUFFER REAL PART WITH RHC PEAK
0304   1020   CONTINUE
0305          AVGVAL=CMPLX(0.,0.)
0306          GOTO 1070
      C
      C  REMOVE DC VALUE FROM FFT INPUT (IF IMPLEMENTED)
```

```
      C
      C     DO 1050 I=1,NSTEP
      C1050 AVGVAL=AVGVAL+VALUE(I)
      C     AVGVAL=AVGVAL/FLOAT(NSTEP)
      C     DO 1060 I=1,NSTEP
      C1060 VALUE(I)=VALUE(I)-AVGVAL
0287  1070  CONTINUE
      C     CALL SMMATE(VALUE,N)          !FFT INPUT WEIGHTING IF CALLED
0288        CALL ALOGN(N,VALUE,+1.)
0289        CALL PLOT (0)
0290        DELX=C/(2.*FSW)
0291        IXMAX=IFIX(DELX*(NFFT-1)+0.5)
0292        IREMAN=MOD(IXMAX,5)
0293        IF(IREMAN.EQ.0)GOTO 1080
0295        IXMAX=IXMAX+(5-IREMAN)
0296  1080  XMIN=0.
0297        XMAX=FLOAT(IXMAX)
0298        YMIN=0.
0299        YMAX=1.
0300        CALL AXES (A,XMIN,XMAX,(XMAX-XMIN)/50.,
     1    YMIN,YMAX,(YMAX-YMIN)/50.,XMIN,YMIN,840.,600.,150.,100.)
0301        READ (22,6006)NXAXIS
0302        READ (22,6006)NYAXIS
0303        CALL LABEL(A,NXAXIS,NYAXIS)
0304        CALL HEADER
      C
      C  PLOTTING FFT OF HORIZ OR RHC CHANNEL
      C
0305        DO 1100 I=1,NFFT
0306        X=(I-1)*DELX
0307        Y=CABS(VALUE(I))/BIG
0308        CALL LINE(A,X,0.,0)
0309  1100  CALL LINE(A,X,Y,1)
0310        IF(ICPY.EQ.0) GOTO 1110
0312        CALL HRDCPY
0313        CALL STALL
0314        GOTO 1120
0315  1110  ACCEPT 6006,IANS
0316  1120  CALL PLOT (0)
0317        XMIN=0.
0318        XMAX=FLOAT(IXMAX)
0319        YMIN=-180.
0320        YMAX=180.
0321        CALL AXES (A,XMIN,XMAX,(XMAX-XMIN)/50.,
     1    YMIN,YMAX,(YMAX-YMIN)/50.,XMIN,YMIN,840.,600.,150.,100.)
0322        READ (22,6006)NXAXIS
0323        READ (22,6006)NYAXIS
0324        CALL LABEL(A,NXAXIS,NYAXIS)
0325        CALL HEADER
      C
      C  PLOTTING HORIZ OR RHC FFT PHASE ANGLE DATA
      C
0326        DO 1200 I=1,NFFT
0327        X=(I-1)*DELX
```

```
0328          Y=ATAN2(AIMAG(VALUE(I)),REAL(VALUE(I)))
0329          Y=AMOD(Y,PI2)
0330          IF(Y.GT.PI)Y=Y-PI2
0332          IF(Y.LT.-PI)Y=PI2+Y
0334          Y=Y*180./PI
0335          CALL LINE(A,X,0.,0)
0336   1200   CALL LINE(A,X,Y,1)
0337          IF(ICPY.EQ.0) GOTO 1210
0339          CALL HKOCPY
0340          CALL SIALL
0341          GOTO 1500
0342   1210   ACCEPT 6006,IANS
0343   1500   CONTINUE
0344          DO 2500 IFFT=1,2
0345          DO 2010 I=1,NFFT
0346   2010   VALUE(I)=CMPLX(0.,0.)        !ZERO OUT COMPLEX BUFFER VALUE
0347          DO 2020 I=1,NSTEP
0348          IF(ICIN.EQ.0.AND.IFFT.EQ.1)VALUE(I)=
              1    CMPLX(VISAV(I),VQSAV(I)) !LOAD BUFFER WITH VERT I&Q
0350          IF(ICIN.EQ.0.AND.IFFT.EQ.2)VALUE(I)=
              1    CMPLX(AVSAV(I),0.)       !LOAD BUFFER REAL PART WITH VERT AMP
0352          IF(ICIN.EQ.1.AND.IFFT.EQ.1)VALUE(I)=LMCSAV(I)       !LOAD BUFFER WITH LI
0354          IF(ICIN.EQ.1.AND.IFFT.EQ.2)VALUE(I)=
              1    CMPLX(CABS(LMCSAV(I)),0.)        !LOAD BUFFER REAL PART WITH LMC AMP
0356   2020   CONTINUE
0357          AVGVAL=CMPLX(0.,0.)
0358          GOTO 2070
              C
              C   REMOVE DC VALUE FROM FFT INPUT (IF IMPLEMENTED)
              C
              C     DO 2050 I=1,NSTEP
              C2050 AVGVAL=AVGVAL+VALUE(I)
              C     AVGVAL=AVGVAL/FLOAT(NSTEP)
              C     DO 2060 I=1,NSTEP
              C2060 VALUE(I)=VALUE(I)-AVGVAL
0359   2070   CONTINUE
              C     CALL GWGATE(VALUE,N)         !FFT INPUT WEIGHTING IF CALLED
0360          CALL NLOGN(N,VALUE,+1.)
0361          CALL PLOT (0)
0362          XMIN=0.
0363          XMAX=FLOAT(IXMAX)
0364          YMIN=0.
0365          YMAX=1.
0366          CALL AXES (A,XMIN,XMAX,(XMAX-XMIN)/50.,
              1    YMIN,YMAX,(YMAX-YMIN)/50.,XMIN,YMIN,840.,600.,150.,100.)
0367          READ (22,6006)NXAXIS
0368          READ (22,6006)NYAXIS
0369          CALL LABEL(A,NXAXIS,NYAXIS)
0370          CALL HEADER
              C
              C  PLOTTING FFT OF VERT OR LMC CHNL
              C
0371          DO 2100 I=1,NFFT
0372          X=(I-1)*DELX
```

```
0373          Y=CABS(VALUE(I))/BIG
0374          CALL LINE(A,X,0.,0)
0375    2100  CALL LINE(A,X,Y,1)
0376          IF(ICPY.EQ.0) GOTO 2110
0378          CALL HRDCPY
0379          CALL STALL
0380          GOTO 2120
0381    2110  ACCEPT 6006,IANS
0382    2120  CALL PLOT (0)
0383          XMIN=0.
0384          XMAX=FLOAT(IXMAX)
0385          YMIN=-180.
0386          YMAX=180
0387          CALL AXES (A,XMIN,XMAX,(XMAX-XMIN)/50.,
             1    YMIN,YMAX,(YMAX-YMIN)/50.,XMIN,YMIN,840.,600.,150.,100.)
0388          READ (22,6006)NXAXIS
0389          READ (22,6006)NYAXIS
0390          CALL LABEL(A,NXAXIS,NYAXIS)
0391          CALL HEADER
      C
      C  PLOTTING VERT OR LHC FFT PHASE ANGLE
      C
0392          DO 2200 I=1,NFFT
0393          X=(I-1)*DELX
0394          Y=ATAN2(AIMAG(VALUE(I)),REAL(VALUE(I)))
0395          Y=AMOD(Y,PI2)
0396          IF(Y.GT.PI)Y=Y-PI2
0398          IF(Y.LT.-PI)Y=PI2+Y
0400          Y=Y*180./PI
0401          CALL LINE(A,X,0.,0)
0402    2200  CALL LINE(A,X,Y,1)
0403          IF(ICPY.EQ.0) GOTO 2210
0405          CALL HRDCPY
0406          CALL STALL
0407          GOTO 2500
0408    2210  ACCEPT 6006,IANS
0409    2500  CONTINUE
0410          IF(ICIR.EQ.1)GOTO 2600
0412          ICIR=1
0413          GOTO 300
0414    2600  CALL CLOSE (22)
0415          IRUN=IRUN+1
0416          IF(IRUN.EQ.4)GOTO 3000
0418          GOTO 11111
0419    3000  CALL V14CSZ(4)
      C
      C
0420    6012  FORMAT(8A2)
0421    6011  FORMAT(4F20.0)
0422    6010  FORMAT(1X,'FILE NAME: '8A2/
             1    1X,'SCALE: ',F7.0/1X,'NUMBER SCATTERS: ',I6/)
0425    6009  FORMAT(1X,3(F7.2,','),F7.2)
0424    6008  FORMAT(1X,T5,'FREQ',
             1    T18,'HORZ I',T31,'HORZ Q',T44,'VERT I',T57,'VERT Q',
```

```
     1   T70,'HORZ AMP',T83,'VERT AMP',T96,'BETA',
     1   T109,'H BETA',T122,'V BETA')
0425  6007  FORMAT(10(2X,1PE11.4))
0426  6006  FORMAT(20A2)
0427  6005  FORMAT(' ANTENNA ISOLATION IN DB')
0428  6004  FORMAT(F10.0)
0429  6003  FORMAT(' NUMBER OF FREQUENCY RAMP STEPS?'/)
0430  6002  FORMAT(I6)
0431  6001  FORMAT(/' NUMBER OF FFT POINTS (LESS OR EQUAL 256'/)
0432  6000  FORMAT(' RUN NUMBER ',I3//' DATA FILE FOR SCATTERERS'/)
0433        END
```

FORTRAN IV STORAGE MAP

NAME	OFFSET	ATTRIBUTES
NXAXIS	000006	INTEGER*2 ARRAY (20)
NYAXIS	000056	INTEGER*2 ARRAY (20)
A	000126	REAL*4 ARRAY (20)
LBETA	001200	REAL*4 VARIABLE
LAMDA	001204	REAL*4 VARIABLE
AVGVAL	001210	COMPLEX*8 VARIABLE
EH	001220	COMPLEX*8 VARIABLE
EV	001230	COMPLEX*8 VARIABLE
ETHOR	001240	COMPLEX*8 VARIABLE
ETVERT	001250	COMPLEX*8 VARIABLE
ANGH	001260	COMPLEX*8 VARIABLE
ANGV	001270	COMPLEX*8 VARIABLE
EHTP	001300	COMPLEX*8 VARIABLE
EVTP	001310	COMPLEX*8 VARIABLE
EHT	001320	COMPLEX*8 VARIABLE
EVT	001330	COMPLEX*8 VARIABLE
ANGHT	001340	COMPLEX*8 VARIABLE
ANGVT	001350	COMPLEX*8 VARIABLE
EHACC	001360	COMPLEX*8 VARIABLE
EVACC	001370	COMPLEX*8 VARIABLE
VNH	001400	COMPLEX*8 VARIABLE
VNV	001410	COMPLEX*8 VARIABLE
V1USR2	001420	REAL*4 VARIABLE
SQRT	000000	REAL*4 PROCEDURE
UVOLTS	001424	REAL*4 VARIABLE
C	001430	REAL*4 VARIABLE
ICPY	001434	INTEGER*2 VARIABLE
IRUN	001436	INTEGER*2 VARIABLE
PI	001440	REAL*4 VARIABLE
PI2	001444	REAL*4 VARIABLE
PI4	001450	REAL*4 VARIABLE
CDR	001454	REAL*4 VARIABLE
PLOT	000000	REAL*4 PROCEDURE
V14CSZ	000000	REAL*4 PROCEDURE
XMIT	000000	REAL*4 PROCEDURE
E1H	001460	REAL*4 VARIABLE
E1V	001464	REAL*4 VARIABLE
PHITV	001470	REAL*4 VARIABLE
R1	001474	REAL*4 VARIABLE
R2	001500	REAL*4 VARIABLE
CMPLX	000000	COMPLEX*8 PROCEDURE
CEXP	000000	COMPLEX*8 PROCEDURE
PHIRH	001504	REAL*4 VARIABLE
PHIRV	001510	REAL*4 VARIABLE
NFFT	001514	INTEGER*2 VARIABLE
ASSIGN	000000	REAL*4 PROCEDURE
RIFOR	001516	REAL*4 VARIABLE
RNFOR	001522	REAL*4 VARIABLE
RNF	001526	REAL*4 VARIABLE
VARI	001532	REAL*4 VARIABLE
PRF	001536	REAL*4 VARIABLE
FLOAT	000000	REAL*4 PROCEDURE

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FORTRAN IV      STORAGE MAP

NAME    OFFSET  ATTRIBUTES

ANTG    001542  REAL*4      VARIABLE
ASCALE  001546  REAL*4      VARIABLE
I       001552  INTEGER*2   VARIABLE
K       001554  INTEGER*2   VARIABLE
CLOSE   000000  REAL*4      PROCEDURE
ASNRH   001556  REAL*4      VARIABLE
ASNRV   001562  REAL*4      VARIABLE
IF      001566  INTEGER*2   VARIABLE
F       001570  REAL*4      VARIABLE
FREQ    000000  REAL*4      PROCEDURE
PHIF    001574  REAL*4      VARIABLE
PHIFD   001600  REAL*4      VARIABLE
GETSM   000000  REAL*4      PROCEDURE
SCALE   001604  REAL*4      VARIABLE
RSCALE  000000  REAL*4      PROCEDURE
VN1     001610  REAL*4      VARIABLE
VNOISE  000000  REAL*4      PROCEDURE
RANPH   000000  REAL*4      PROCEDURE
PN1     001614  REAL*4      VARIABLE
VN2     001620  REAL*4      VARIABLE
PN2     001624  REAL*4      VARIABLE
VNMI    001630  REAL*4      VARIABLE
COS     000000  REAL*4      PROCEDURE
VNMQ    001634  REAL*4      VARIABLE
SIN     000000  REAL*4      PROCEDURE
VNVI    001640  REAL*4      VARIABLE
VNVQ    001644  REAL*4      VARIABLE
CABS    000000  REAL*4      PROCEDURE
MI      001650  REAL*4      VARIABLE
REAL    000000  REAL*4      PROCEDURE
MQ      001654  REAL*4      VARIABLE
AIMAG   000000  REAL*4      PROCEDURE
VI      001660  REAL*4      VARIABLE
VQ      001664  REAL*4      VARIABLE
MZANG   001670  REAL*4      VARIABLE
VTANG   001674  REAL*4      VARIABLE
ATAN2   000000  REAL*4      PROCEDURE
BETA    001700  REAL*4      VARIABLE
AMOD    000000  REAL*4      PROCEDURE
RBETA   001704  REAL*4      VARIABLE
CBETA   001710  REAL*4      VARIABLE
N       001714  INTEGER*2   VARIABLE
ALOG10  000000  REAL*4      PROCEDURE
ANORM   000000  REAL*4      PROCEDURE
SAM     000000  REAL*4      PROCEDURE
ISW     001716  INTEGER*2   VARIABLE
IAND    000000  INTEGER*2   PROCEDURE
ICIN    001720  INTEGER*2   VARIABLE
IFINST  001722  INTEGER*2   VARIABLE
XMIN    001724  REAL*4      VARIABLE
XMAX    001730  REAL*4      VARIABLE
YMIN    001734  REAL*4      VARIABLE
```

```
FORTRAN IV       STORAGE MAP

NAME     OFFSET  ATTRIBUTES

YMAX     001740  REAL*4     VARIABLE
AXES     000000  REAL*4     PROCEDURE
LABEL    000000  INTEGER*2  PROCEDURE
HEADER   000000  REAL*4     PROCEDURE
X        001744  REAL*4     VARIABLE
Y        001750  REAL*4     VARIABLE
LINE     000000  INTEGER*2  PROCEDURE
HRDCPY   000000  REAL*4     PROCEDURE
STALL    000000  REAL*4     PROCEDURE
IANS     001754  INTEGER*2  VARIABLE
IFFT     001756  INTEGER*2  VARIABLE
XLOGN    000000  INTEGER*2  PROCEDURE
DELX     001760  REAL*4     VARIABLE
IXMAX    001764  INTEGER*2  VARIABLE
IFIX     000000  INTEGER*2  PROCEDURE
IREMAN   001766  INTEGER*2  VARIABLE
MOD      000000  INTEGER*2  PROCEDURE


COMMON BLOCK /WORKF/     LENGTH 000024

IFSTFG   000000  INTEGER*2  VARIABLE
IUP      000002  INTEGER*2  VARIABLE
LSTEP    000004  INTEGER*2  VARIABLE
NSTEP    000006  INTEGER*2  VARIABLE
DF       000010  REAL*4     VARIABLE
CF       000014  REAL*4     VARIABLE
FSM      000020  REAL*4     VARIABLE

COMMON BLOCK /WKSCT/     LENGTH 003140

SCATER   000000  REAL*4     ARRAY (100,4) VECTORED
SMATRX   003100  COMPLEX*8  ARRAY (2,2) VECTORED

COMMON BLOCK /HEAD/      LENGTH 000114

AISOL    000000  REAL*4     VARIABLE
NSCAT    000004  INTEGER*2  VARIABLE
GAINA    000006  REAL*4     VARIABLE
NOISE    000012  INTEGER*2  VARIABLE
RANGE    000014  REAL*4     VARIABLE
OBLOSS   000020  REAL*4     VARIABLE
VXMIT    000024  INTEGER*2  ARRAY (2)
IFILE    000030  INTEGER*2  ARRAY (8)
SU       000050  REAL*4     VARIABLE
DIG      000054  REAL*4     VARIABLE
SNRH     000060  REAL*4     VARIABLE
SNRV     000064  REAL*4     VARIABLE
SNRHI    000070  REAL*4     VARIABLE
SNRHQ    000074  REAL*4     VARIABLE
SNRVI    000100  REAL*4     VARIABLE
SNRVQ    000104  REAL*4     VARIABLE
```

```
FORTRAN IV       STORAGE MAP

NAME     OFFSET   ATTRIBUTES

SNR      000110   REAL*4      VARIABLE

COMMON BLOCK /WORK/      LENGTH 042000

MQSAV    000000   REAL*4      ARRAY (256)
MISAV    002000   REAL*4      ARRAY (256)
VQSAV    004000   REAL*4      ARRAY (256)
VISAV    006000   REAL*4      ARRAY (256)
SFREQ    010000   REAL*4      ARRAY (256)
SBETAM   012000   REAL*4      ARRAY (256)
SBETAV   014000   REAL*4      ARRAY (256)
QETSAV   016000   REAL*4      ARRAY (256)
RHCSAV   020000   COMPLEX*8   ARRAY (256)
LHCSAV   024000   COMPLEX*8   ARRAY (256)
CBETAS   030000   REAL*4      ARRAY (256)
AMSAV    032000   REAL*4      ARRAY (256)
AVSAV    034000   REAL*4      ARRAY (256)
VALUE    036000   COMPLEX*8   ARRAY (256)

COMMON BLOCK /SIGNAL/    LENGTH 000030

PTPWR    000000   REAL*4      VARIABLE
RANGE4   000004   REAL*4      VARIABLE
CR       000010   REAL*4      VARIABLE
ANTG2    000014   REAL*4      VARIABLE
SLOSS    000020   REAL*4      VARIABLE
PI4C     000024   REAL*4      VARIABLE
```

```
FORTRAN IV       V01C-03F+  THU 28-OCT-82 00:06:48                PAGE 001

0001          SUBROUTINE XMIT (E1H,E1V,PHITV,NAME)
              C
              C  THIS SUBROUTINE DETERMINES THE TRANSMITTED SIGNAL POLARIZATION
              C
0002          INTEGER NAME (2)
0003          INTEGER RHC(2),LHC(2),HOR(2),VER(2),HV(2)
0004          DATA RHC,LHC,HOR,VER,HV/'RH','C ','LH','C ','HO','R ',
             1   'VE','R ','H-','V '/
0005          PI=3.14159
0006    1     CALL PLOT(0)
0007          TYPE 10
0008    10    FORMAT(' 1 - RHC'/' 2 - LHC'/' 3 - HORIZONTAL'/' 4 - VERTICAL'/
             1  ' 5 - HORIZONTAL & VERTICAL'/)
0009          ACCEPT 15,IXMIT
0010    15    FORMAT(I6)
0011          IF(IXMIT.LT.1.OR.IXMIT.GT.5)GOTO 1
0013          GOTO (100,200,300,400,500)IXMIT
0014    100   E1H=1.
0015          E1V=1.
0016          PHITV=-PI/2
0017          NAME(1)=RHC(1)
0018          NAME(2)=RHC(2)
0019          RETURN
0020    200   E1H=1.
0021          E1V=1.
0022          PHITV=+PI/2
0023          NAME(1)=LHC(1)
0024          NAME(2)=LHC(2)
0025          RETURN
0026    300   E1H=1.
0027          E1V=0.
0028          PHITV=0.
0029          NAME(1)=HOR(1)
0030          NAME(2)=HOR(2)
0031          RETURN
0032    400   E1H=0.
0033          E1V=1.
0034          PHITV=0.
0035          NAME(1)=VER(1)
0036          NAME(2)=VER(2)
0037          RETURN
0038    500   E1H=1.
0039          E1V=1.
0040          PHITV=0.
0041          NAME(1)=HV(1)
0042          NAME(2)=HV(2)
0043          RETURN
0044          END
```

```
FORTRAN IV       STORAGE MAP

NAME     OFFSET   ATTRIBUTES

NAME     000022   INTEGER*2 PARAMETER ARRAY (2)
RHC      000024   INTEGER*2 ARRAY (2)
LHC      000030   INTEGER*2 ARRAY (2)
HOR      000034   INTEGER*2 ARRAY (2)
VER      000040   INTEGER*2 ARRAY (2)
HV       000044   INTEGER*2 ARRAY (2)
E1H      000014   REAL*4     PARAMETER VARIABLE
E1V      000016   REAL*4     PARAMETER VARIABLE
PHITV    000020   REAL*4     PARAMETER VARIABLE
PI       000214   REAL*4     VARIABLE
PLOT     000000   REAL*4     PROCEDURE
IXMIT    000220   INTEGER*2 VARIABLE
```

```
FORTRAN IV      V01C-03F+  THU 28-OCT-82 00:06:57                PAGE 001

0001          SUBROUTINE BHWATE(A,N)
      C
      C  FFT INPUT WEIGHTING
      C
0002          COMPLEX A(N)
0003          DATA PI2/6.283185/
0004          DO 100 I=1,N
0005          WATE=0.42323-0.49755*COS(PI2/N*(I-1))+0.07922*COS(PI2/N*2*(I-1))
0006    100   A(I)=A(I)*WATE
0007          RETURN
0008          END
```

```
FORTRAN IV      STORAGE MAP

NAME     OFFSET   ATTRIBUTES

A        000014   COMPLEX*8 PARAMETER ARRAY (N)
N        000016   INTEGER*2 PARAMETER VARIABLE
PI2      000020   REAL*4    VARIABLE
I        000040   INTEGER*2 VARIABLE
WATE     000042   REAL*4    VARIABLE
COS      000000   REAL*4    PROCEDURE
```

```
FORTRAN IV      V01C-03F+  THU 28-OCT-82 00:07:02                PAGE 001

0001          FUNCTION VNOISE(SD)
      C
      C  THIS FUNCTION GENERATES GAUSSIAN DISTRIBUTED NOISE VOLTAGE
      C
0002          SUM=0.
0003          DO 10 I=1,12
0004    10    SUM=SUM+RANF(U)
0005          VNOISE=(SUM-6.)*SD
0006          RETURN
0007          END
```

```
FORTRAN IV      STORAGE MAP

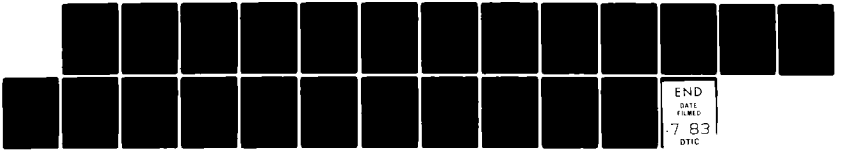
NAME     OFFSET   ATTRIBUTES

VNOISE   000016   REAL*4    VARIABLE
SD       000014   REAL*4    PARAMETER VARIABLE
SUM      000022   REAL*4    VARIABLE
I        000026   INTEGER*2 VARIABLE
RANF     000000   REAL*4    PROCEDURE
U        000030   REAL*4    VARIABLE
```

AD-A129 502 ANALYTICAL RESEARCH BY COMPUTER SIMULATION OF DEVELOPMENTAL POLARIMETRIC/..(U) ARMY MISSILE COMMAND REDSTONE ARSENAL AL ADVANCED SENSORS DIR..

2/2

UNCLASSIFIED R F RUSSELL ET AL. DEC 82 DRSMI-RE-83-7-TR F/G 17/9 NL



END
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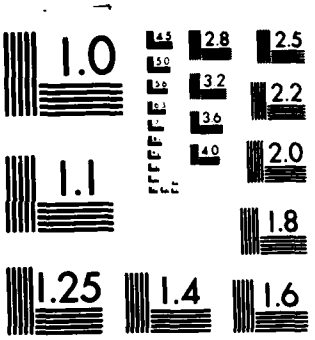

MICROCOPY RESOLUTION TEST CHART
NATIONAL BUREAU OF STANDARDS-1963-A

```
FORTRAN IV      V01C-03F+  THU 28-OCT-82 00:07:06                PAGE 001

      C
0001          SUBROUTINE RANPH(PHASE)
      C
      C  THIS SUBROUTINE GENERATES UNIFORMALLY DISTRIBUTED PHASE NOISE
      C
0002          DATA PI /3.14159/
0003          PHASE=(RANF(U)-.5)*PI
0004          RETURN
0005          END


FORTRAN IV      STORAGE MAP

NAME     OFFSET  ATTRIBUTES

PHASE    000014  REAL*4     PARAMETER VARIABLE
PI       000016  REAL*4     VARIABLE
RANF     000000  REAL*4     PROCEDURE
U        000022  REAL*4     VARIABLE


FORTRAN IV      V01C-03F+  THU 28-OCT-82 00:07:10                PAGE 001

      C
0001          FUNCTION RANF(U)
      C
      C  UNIFORM NUMBER GENERATOR
      C
0002          DATA I,J/0,0/
0003          RANF=RAN(I,J)
0004          RETURN
0005          END


FORTRAN IV      STORAGE MAP

NAME     OFFSET  ATTRIBUTES

RANF     000016  REAL*4     VARIABLE
U        000014  REAL*4     PARAMETER VARIABLE
I        000022  INTEGER*2  VARIABLE
J        000024  INTEGER*2  VARIABLE
RAN      000000  REAL*4     PROCEDURE
```

```
0001          SUBROUTINE ANORM(N,NFFT,NSTEP,BIG)
      C
      C  THIS SUBROUTINE DETERMINES THE BIGGEST FFT OUTPUT FOR NORMALIZATION
      C    OF FFT PLOTS
      C
0002          COMPLEX VALUE(256),SMATRX(2,2)
0003          DIMENSION SFREQ(256)
0004          DIMENSION SBETAH(256),SBETAV(256)
0005          DIMENSION SCATER(100,4),AHSAV(256),AVSAV(256),BETSAV(256)
0006          DIMENSION HQSAV(256),HISAV(256),VQSAV(256),VISAV(256)
0007          DIMENSION CBETAS(256)
0008          COMPLEX RHCSAV(256),LHCSAV(256)
0009          COMMON /WORK/HQSAV,HISAV,VQSAV,VISAV,SFREQ,SBETAH,SBETAV,BETSAV,
             1             RHCSAV,LHCSAV,CBETAS,AHSAV,AVSAV,
             1             VALUE
0010          COMMON /WKSCT/SCATER,SMATRX
      C
0011          BIG=0.
0012          DO 3 I=1,NFFT
0013    3     VALUE(I)=CMPLX(0.,0.)
0014          DO 5 I=1,NSTEP
0015    5     VALUE(I)=CMPLX(CABS(RHCSAV(I)),0.0) !REAL PART WITH RHC AMP
0016          CALL NLOGN (N,VALUE,+1.)
0017          CALL BIGEST(NFFT,BIG)
0018          DO 13 I=1,NFFT
0019   13     VALUE(I)=CMPLX(0.,0.)
0020          DO 15 I=1,NSTEP
0021   15     VALUE(I)=CMPLX(CABS(LHCSAV(I)),0.0) !REAL PART WITH LHC AMP
0022          CALL NLOGN (N,VALUE,+1.)
0023          CALL BIGEST(NFFT,BIG)
0024          DO 18 I=1,NFFT
0025   18     VALUE(I)=CMPLX(0.,0.)
0026          DO 20 I=1,NSTEP
0027   20     VALUE(I)=CMPLX(AHSAV(I),0.) !REAL PART WITH HORIZ AMP
0028          CALL NLOGN(N,VALUE,+1.)
0029          CALL BIGEST(NFFT,BIG)
0030          DO 23 I=1,NFFT
0031   23     VALUE(I)=CMPLX(0.,0.)
0032          DO 25 I=1,NSTEP
0033   25     VALUE(I)=CMPLX(AVSAV(I),0.) !REAL PART WITH VERT AMP
0034          CALL NLOGN(N,VALUE,+1.)
0035          CALL BIGEST(NFFT,BIG)
0036          RETURN
0037          END
```

```
FORTRAN IV       STORAGE MAP

NAME     OFFSET  ATTRIBUTES

N        000014  INTEGER*2 PARAMETER VARIABLE
NFFT     000016  INTEGER*2 PARAMETER VARIABLE
NSTEP    000020  INTEGER*2 PARAMETER VARIABLE
BIG      000022  REAL*4    PARAMETER VARIABLE
I        000050  INTEGER*2 VARIABLE
CMPLX    000000  COMPLEX*8 PROCEDURE
CABS     000000  REAL*4    PROCEDURE
NLOGN    000000  INTEGER*2 PROCEDURE
BIGEST   000000  REAL*4    PROCEDURE


COMMON BLOCK /WORK/      LENGTH 042000

HQSAV    000000  REAL*4    ARRAY (256)
HISAV    002000  REAL*4    ARRAY (256)
VQSAV    004000  REAL*4    ARRAY (256)
VISAV    006000  REAL*4    ARRAY (256)
SFREQ    010000  REAL*4    ARRAY (256)
SBETAH   012000  REAL*4    ARRAY (256)
SBETAV   014000  REAL*4    ARRAY (256)
BETSAV   016000  REAL*4    ARRAY (256)
HHCSAV   020000  COMPLEX*8 ARRAY (256)
LHCSAV   024000  COMPLEX*8 ARRAY (256)
CBETAS   030000  REAL*4    ARRAY (256)
AHSAV    032000  REAL*4    ARRAY (256)
AVSAV    034000  REAL*4    ARRAY (256)
VALUE    036000  COMPLEX*8 ARRAY (256)

COMMON BLOCK /WKSCT/     LENGTH 003140

SCATER   000000  REAL*4    ARRAY (100,4) VECTORED
SMATRX   003100  COMPLEX*8 ARRAY (2,2) VECTORED
```

```
FORTRAN IV      V01C-03F+  THU 28-OCT-82 00:07:25                PAGE 001

      C
      C
0001          SUBROUTINE BIGEST(N,BIG)
0002          COMPLEX VALUE(256),SMATRX(2,2)
0003          DIMENSION SFREQ(256)
0004          DIMENSION SBETAH(256),SBETAV(256)
0005          DIMENSION SCATER(100,4),AHSAV(256),AVSAV(256),BETSAV(256)
0006          DIMENSION HQSAV(256),HISAV(256),VQSAV(256),VISAV(256)
0007          DIMENSION CBETAS(256)
0008          COMPLEX RHCSAV(256),LHCSAV(256)
0009          COMMON /WORK/HQSAV,HISAV,VQSAV,VISAV,SFREQ,SBETAH,SBETAV,BETSAV,
             1             RHCSAV,LHCSAV,CBETAS,AHSAV,AVSAV,
             1             VALUE
0010          COMMON /WKSCT/SCATER,SMATRX
      C
0011          DO 10 I=1,N
0012          AVAL=CABS(VALUE(I))
0013    10    IF(AVAL.GT.BIG)BIG=AVAL
0015          RETURN
0016          END


FORTRAN IV      STORAGE MAP

NAME     OFFSET  ATTRIBUTES

N        000014  INTEGER*2 PARAMETER VARIABLE
BIG      000016  REAL*4    PARAMETER VARIABLE
I        000034  INTEGER*2 VARIABLE
AVAL     000036  REAL*4    VARIABLE
CABS     000000  REAL*4    PROCEDURE


COMMON BLOCK /WORK/     LENGTH 042000

HQSAV    000000  REAL*4    ARRAY (256)
HISAV    002000  REAL*4    ARRAY (256)
VQSAV    004000  REAL*4    ARRAY (256)
VISAV    006000  REAL*4    ARRAY (256)
SFREQ    010000  REAL*4    ARRAY (256)
SBETAH   012000  REAL*4    ARRAY (256)
SBETAV   014000  REAL*4    ARRAY (256)
BETSAV   016000  REAL*4    ARRAY (256)
RHCSAV   020000  COMPLEX*8 ARRAY (256)
LHCSAV   024000  COMPLEX*8 ARRAY (256)
CBETAS   030000  REAL*4    ARRAY (256)
AHSAV    032000  REAL*4    ARRAY (256)
AVSAV    034000  REAL*4    ARRAY (256)
VALUE    036000  COMPLEX*8 ARRAY (256)

COMMON BLOCK /WKSCT/    LENGTH 003140

SCATER   000000  REAL*4    ARRAY (100,4) VECTORED
SMATRX   003100  COMPLEX*8 ARRAY (2,2) VECTORED
```

```
FORTRAN IV      V01C-03F+  THU 28-OCT-82 00:07:33                PAGE 001

0001          FUNCTION RSCALE (LAMDA)
      C
      C    AMPLITUDE SCALE FUNCTION
      C
0002          REAL LAMDA,IMPED
0003          COMMON /SIGNAL/PTPWR,RANGE4,CR,ANTG2,SLOSS,PI4C
0004          DATA IMPED/50./
0005          PR=PTPWR*ANTG2*(LAMDA**2.)*CR/(PI4C*RANGE4*SLOSS)
      C
      C  PEAK OUTPUT VOLTAGE IS RELATED TO AVERAGE TRANSMITTER
      C    POWER OUTPUT (WHEN TRANSMITTER IS SWITCHED ON) BY SQRT(2.)
      C
0006          RSCALE=SQRT(PR*IMPED*2.)
0007          RETURN
0008          END


FORTRAN IV      STORAGE MAP

NAME     OFFSET   ATTRIBUTES

RSCALE   000016   REAL*4     VARIABLE
LAMDA    000014   REAL*4     PARAMETER VARIABLE
IMPED    000022   REAL*4     VARIABLE
PR       000026   REAL*4     VARIABLE
SQRT     000000   REAL*4     PROCEDURE


COMMON BLOCK /SIGNAL/   LENGTH 000030

PTPWR    000000   REAL*4     VARIABLE
RANGE4   000004   REAL*4     VARIABLE
CR       000010   REAL*4     VARIABLE
ANTG2    000014   REAL*4     VARIABLE
SLOSS    000020   REAL*4     VARIABLE
PI4C     000024   REAL*4     VARIABLE
```

```
FORTRAN IV      V01C-03F+  THU 28-OCT-82 00:07:38                PAGE 001

0001          FUNCTION FREQ(I)
      C
      C  GENERATE FREQUENCY OUTPUT STEP AS A FUNCTION OF
      C     THE LAST FREQUENCY RAMP STEP TRANSMITTED
      C
0002          COMMON /WORKF/IFSTFQ,IUP,LSTEP,NSTEP,DF,CF,FBW
      C
0003          IF(IFSTFQ.NE.1)GOTO 5
0005          IFSTFQ=0
0006          FREQ=CF-FBW/2.
0007          IUP=1
0008          LSTEP=1
0009          RETURN
0010    5     IF(IUP.NE.1)GOTO 100
0012          IF(LSTEP.NE.NSTEP)GOTO 10
0014          IUP=0
0015          RETURN
0016    10    LSTEP=LSTEP+1
0017          FREQ=FREQ+DF
0018          RETURN
0019    100   IF(LSTEP.NE.1)GOTO 110
0021          IUP=1
0022          RETURN
0023    110   LSTEP=LSTEP-1
0024          FREQ=FREQ-DF
0025          RETURN
0026          END


FORTRAN IV      STORAGE MAP

NAME     OFFSET   ATTRIBUTES

FREQ     000016   REAL*4     VARIABLE
I        000014   INTEGER*2 PARAMETER VARIABLE


COMMON BLOCK /WORKF/     LENGTH 000024

IFSTFQ   000000   INTEGER*2 VARIABLE
IUP      000002   INTEGER*2 VARIABLE
LSTEP    000004   INTEGER*2 VARIABLE
NSTEP    000006   INTEGER*2 VARIABLE
DF       000010   REAL*4     VARIABLE
CF       000014   REAL*4     VARIABLE
FBW      000020   REAL*4     VARIABLE
```

```
FORTRAN IV      V01C-03F+  THU 28-OCT-82 00:07:45                PAGE 001

0001          SUBROUTINE GETSM(I)
              C
              C  DETERMINE SCATTERER TYPE AND CALL IT'S MATRIX
              C
0002          COMPLEX SMATRX(2,2)
0003          DIMENSION SCATER(100,4)
0004          COMMON /WKSCT/SCATER,SMATRX
0005          GOTO(100,200,300,400)IFIX(SCATER(I,1))
0006    100   CALL PLATE(I)
0007          RETURN
0008    200   CALL DIHED(I)
0009          RETURN
0010    300   CALL TRIHED(I)
0011          RETURN
0012    400   CALL DIPOLE(I)
0013          RETURN
0014          END



FORTRAN IV      STORAGE MAP

NAME     OFFSET  ATTRIBUTES

I        000014  INTEGER*2 PARAMETER VARIABLE
IFIX     000000  INTEGER*2 PROCEDURE
PLATE    000000  REAL*4    PROCEDURE
DIHED    000000  REAL*4    PROCEDURE
TRIHED   000000  REAL*4    PROCEDURE
DIPOLE   000000  REAL*4    PROCEDURE


COMMON BLOCK /WKSCT/    LENGTH 003140

SCATER   000000  REAL*4    ARRAY (100,4) VECTORED
SMATRX   003100  COMPLEX*8 ARRAY (2,2) VECTORED
```

```
FORTRAN IV      V01C-03F+  THU 28-OCT-82 00:07:51                PAGE 001

0001          SUBROUTINE PLATE(I)
      C
      C  FLAT PLATE SCATTERING MATRIX
      C
0002          DIMENSION SCATER(100,4)
0003          COMPLEX SMATRX(2,2)
0004          COMMON /WKSCT/SCATER,SMATRX
0005          SRSIGM=SQRT(SCATER(I,2))
0006          SMATRX(1,1)=CMPLX(-1.,0.)*SRSIGM
0007          SMATRX(1,2)=CMPLX(0.,0.)*SRSIGM
0008          SMATRX(2,1)=SMATRX(1,2)
0009          SMATRX(2,2)=SMATRX(1,1)
0010          RETURN
0011          END
```

```
FORTRAN IV      STORAGE MAP

NAME     OFFSET  ATTRIBUTES

I        000014  INTEGER*2 PARAMETER VARIABLE
SRSIGM   000036  REAL*4     VARIABLE
SQRT     000000  REAL*4     PROCEDURE
CMPLX    000000  COMPLEX*8  PROCEDURE


COMMON BLOCK /WKSCT/    LENGTH 003140

SCATER   000000  REAL*4     ARRAY (100,4) VECTORED
SMATRX   003100  COMPLEX*8  ARRAY (2,2) VECTORED
```

```
FORTRAN IV      V01C-03F+  THU 28-OCT-82 00:07:56                PAGE 001

      C
0001          SUBROUTINE DIMED(I)
      C
      C  DIHEDRAL SCATTERING MATRIX
      C
0002          DIMENSION SCATER(100,4)
0003          COMPLEX SMATRX(2,2)
0004          COMMON /WKSCT/SCATER,SMATRX
0005          SRSIGM=SQRT(SCATER(I,2))
0006          SMATRX(1,1)=CMPLX(COS(2.*SCATER(I,3)),0.)*SRSIGM
0007          SMATRX(1,2)=CMPLX(SIN(2.*SCATER(I,3)),0.)*SRSIGM
0008          SMATRX(2,1)=SMATRX(1,2)
0009          SMATRX(2,2)=CMPLX(-COS(2.*SCATER(I,3)),0.)*SRSIGM
0010          RETURN
0011          END
```

```
FORTRAN IV      STORAGE MAP

NAME     OFFSET   ATTRIBUTES

I        000014   INTEGER*2 PARAMETER VARIABLE
SRSIGM   000036   REAL*4     VARIABLE
SQRT     000000   REAL*4     PROCEDURE
CMPLX    000000   COMPLEX*8  PROCEDURE
COS      000000   REAL*4     PROCEDURE
SIN      000000   REAL*4     PROCEDURE

COMMON BLOCK /WKSCT/     LENGTH 003140

SCATER   000000   REAL*4     ARRAY (100,4) VECTORED
SMATRX   003100   COMPLEX*8  ARRAY (2,2) VECTORED
```

```
FORTRAN IV      V01C-03F+  THU 28-OCT-82 00:08:02                PAGE 001

      C
0001          SUBROUTINE TRIHED(I)
      C
      C    TRIHEDRAL SCATTERING MATRIX
      C
0002          DIMENSION SCATER(100,4)
0003          COMPLEX SMATRX(2,2)
0004          COMMON /WKSCT/SCATER,SMATRX
0005          SRSIGM=SQRT(SCATER(I,2))
0006          SMATRX(1,1)=CMPLX(-1.,0.)*SRSIGM
0007          SMATRX(1,2)=CMPLX(0.,0.)*SRSIGM
0008          SMATRX(2,1)=SMATRX(1,2)
0009          SMATRX(2,2)=SMATRX(1,1)
0010          RETURN
0011          END
```

```
FORTRAN IV      STORAGE MAP

NAME     OFFSET  ATTRIBUTES

I        000014  INTEGER*2 PARAMETER VARIABLE
SRSIGM   000036  REAL*4     VARIABLE
SQRT     000000  REAL*4     PROCEDURE
CMPLX    000000  COMPLEX*8  PROCEDURE


COMMON BLOCK /WKSCT/    LENGTH 003140

SCATER   000000  REAL*4     ARRAY (100,4) VECTORED
SMATRX   003100  COMPLEX*8  ARRAY (2,2) VECTORED
```

```
FORTRAN IV      V01C-03F+  THU 28-OCT-82 00:08:07                PAGE 001

      C
0001          SUBROUTINE DIPOLE(I)
      C
      C   DIPOLE SCATTERING MATRIX
      C
0002          DIMENSION SCATER(100,4)
0003          COMPLEX SMATRX(2,2)
0004          COMMON /WKSCT/SCATER,SMATRX
0005          SRSIGM=SQRT(SCATER(I,2))
0006          SMATRX(1,1)=CMPLX(-COS(2.*SCATER(I,3)),0.)*SRSIGM
0007          SMATRX(1,2)=CMPLX(-COS(SCATER(I,3))*SIN(SCATER(I,3)),0.)*
             1    SRSIGM
0008          SMATRX(2,1)=SMATRX(1,2)
0009          SMATRX(2,2)=CMPLX(-SIN(2.*SCATER(I,3)),0.)*SRSIGM
0010          RETURN
0011          END
```

```
FORTRAN IV      STORAGE MAP

NAME     OFFSET   ATTRIBUTES

I        000014   INTEGER*2 PARAMETER VARIABLE
SRSIGM   000036   REAL*4     VARIABLE
SQRT     000000   REAL*4     PROCEDURE
CMPLX    000000   COMPLEX*8  PROCEDURE
COS      000000   REAL*4     PROCEDURE
SIN      000000   REAL*4     PROCEDURE

COMMON BLOCK /WKSCT/    LENGTH 003140

SCATER   000000   REAL*4     ARRAY (100,4) VECTORED
SMATRX   003100   COMPLEX*8  ARRAY (2,2) VECTORED
```

```
FORTRAN IV      V01C-03F+  THU 28-OCT-82 00:17:32                PAGE 001

0001          SUBROUTINE HEADER
      C
      C  PLOTTING HEADER DATA PRINTOUT
      C
0002          COMPLEX SMATRX(2,2)
0003          DIMENSION SCATER(100,4)
0004          INTEGER IFILE(8),NXMIT(2),NPRES,KY,KN
0005          COMMON /WORKF/IFSTFQ,IUP,LSTEP,NSTEP,DF,CF,FBW
0006          COMMON /WKSCT/ SCATER,SMATRX
0007          COMMON /HEAD/AISOL,NSCAT,GAINA,NOISE,
             1   RANGE,DBLOSS,NXMIT,IFILE,SD,BIG,
             1   SNRH,SNRV,SNRHI,SNRHQ,SNRVI,SNRVQ,SNR
0008          COMMON /SIGNAL/PTPWR,RANGE4,CR,ANTG2,SLOSS,PI4C
0009          DATA KY,KN/' Y',' N'/
      C
      C
0010          NPRES=KN
0011          IF(NOISE.EQ.0)GOTO 5
0013          NPRES=KY
      C
0014    5     CALL PLOT(-1,0,780)
0015          CALL V14CSZ(4)
0016          TYPE 9
0017          TYPE 10,IFILE,NSCAT,NSTEP,GAINA,AISOL,NXMIT,PTPWR,
             1   CR,DBLOSS,BIG
0018          TYPE 11,NPRES,(SD/1.E-6),SNRH,SNRV,RANGE
0019    9     FORMAT(1H+,40X,'RF GUIDANCE TECHNOLOGY POLARIZATION SIMULATION')
0020    10    FORMAT(9X,'DATA FILE NAME:',8A2,
             1   1X,'NUM. SCATTERERS:',I3,
             1   3X,'FREQ STEPS:',I3,
             1   3X,'ANT GAIN(DB):',F7.2,
             1   3X,'ANT ISOLATION(DB)',F7.2/
             1   9X,'XMIT:',2A2,
             1   3X,'XMIT PWR/CHNL(WATTS):'F6.2,
             1   3X,'COMP RATIO:',F6.2,
             1   3X,'SYSTEM LOSS(DB):',F5.2,
             1   3X,'FFT SCALER:',1PE14.6)
0021    11    FORMAT(9X,'NOISE:',A2,
             1   3X,'NOISE SD(UVOLTS):',F7.5,
             1   3X,'H AVG SNR(DB):',F6.2,
             1   2X,'V AVG SNR(DB):',F6.2,
             1   3X,'RANGE TO TARGET CELL(METERS):',F8.2)
0022          DO 20 I=1,1000
0023    20    CONTINUE
0024          CALL V14CSZ(1)
0025          RETURN
0026          END
```

```
FORTRAN IV      STORAGE MAP

NAME     OFFSET  ATTRIBUTES

NPRES    000646  INTEGER*2 VARIABLE
KY       000014  INTEGER*2 VARIABLE
KN       000016  INTEGER*2 VARIABLE
PLOT     000000  REAL*4    PROCEDURE
V14CSZ   000000  REAL*4    PROCEDURE
I        000650  INTEGER*2 VARIABLE


COMMON BLOCK /WORKF/    LENGTH 000024

IFSTFQ   000000  INTEGER*2 VARIABLE
IUP      000002  INTEGER*2 VARIABLE
LSTEP    000004  INTEGER*2 VARIABLE
NSTEP    000006  INTEGER*2 VARIABLE
DF       000010  REAL*4    VARIABLE
CF       000014  REAL*4    VARIABLE
FMM      000020  REAL*4    VARIABLE

COMMON BLOCK /WKSCT/    LENGTH 003140

SCATER   000000  REAL*4    ARRAY (100,4) VECTORED
SMATRX   003100  COMPLEX*8 ARRAY (2,2) VECTORED

COMMON BLOCK /HEAD/     LENGTH 000114

AISOL    000000  REAL*4    VARIABLE
NSCAT    000004  INTEGER*2 VARIABLE
GAINA    000006  REAL*4    VARIABLE
NOISE    000012  INTEGER*2 VARIABLE
RANGE    000014  REAL*4    VARIABLE
UMLOSS   000020  REAL*4    VARIABLE
XMIT     000024  INTEGER*2 ARRAY (2)
IFILE    000030  INTEGER*2 ARRAY (8)
SU       000050  REAL*4    VARIABLE
BIG      000054  REAL*4    VARIABLE
SNRH     000060  REAL*4    VARIABLE
SNRV     000064  REAL*4    VARIABLE
SNRHI    000070  REAL*4    VARIABLE
SNRHU    000074  REAL*4    VARIABLE
SNRVI    000100  REAL*4    VARIABLE
SNRVU    000104  REAL*4    VARIABLE
SNR      000110  REAL*4    VARIABLE

COMMON BLOCK /SIGNAL/   LENGTH 000030

PTPRR    000000  REAL*4    VARIABLE
RANGE4   000004  REAL*4    VARIABLE
CR       000010  REAL*4    VARIABLE
ANTG2    000014  REAL*4    VARIABLE
SLOSS    000020  REAL*4    VARIABLE
PI4C     000024  REAL*4    VARIABLE
```

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